# O

## Mullard technical handbook

# **Book three**

Components, materials and assemblies

# **Part five**

Loudspeakers, television assemblies and modules

August 1978



## LOUDSPEAKERS, TELEVISION ASSEMBLIES AND MODULES

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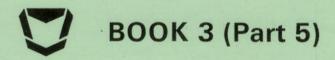
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#### Book 3 comprises the following parts -

- Part 1 Capacitors, resistors
  - Part 2 Magnetic materials and components, piezoelectric ceramics
  - Part 3 Vinkor inductor cores
  - Part 4 RM inductor cores
  - Part 5 Loudspeakers, television assemblies and modules
  - Part 6 Circuit blocks, input and output devices, peripheral devices,



# COMPONENTS MATERIALS AND ASSEMBLIES

Loudspeakers, television assemblies and modules

MULLARD LTD., MULLARD HOUSE, TORRINGTON PLACE, LONDON, WC1E 7HD

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#### DATA HANDBOOK SYSTEM

The Mullard data handbook system is made up of three sets of books, each comprising several parts; plus the Signetics technical handbook.

The three sets of books, easily identifiable by the colours on their covers, are as follows:

Book 1	(blue)	Semiconductor devices and
		integrated circuits
Book 2	(orange)	Valves and tubes
Book 3	(green)	Passive components, materials,

Each part is completely reviewed annually; revised and reprinted where necessary. Revisions to previous data are indicated by an arrow in the margin.

The data contained in these books are as accurate and up to date as it is reasonably possible to make them at the time of going to press. It must however be understood that no guarantee can be given here regarding the availability of the various devices or that their specifications may not be changed before the next edition is published.

The devices on which full data are given in these books are those around which we would recommend equipment to be designed. Where appropriate, other types no longer recommended for new equipment designs, but generally available for equipment production are listed separately with abridged data. Data sheets for these types may be obtained on request. Older devices on which data may still be obtained on request are also included in the index of the appropriate part of each book.

Requests for information on the data handbook system (including Signetics data) and for individual data sheets should be made to

Technical Publications Department Mullard Limited New Road Mitcham Surrey CR4 4XY

Telex: 22194

Information regarding price and availability of devices must be obtained from our authorised agents or from our representatives.

# **SELECTION GUIDE**

## **SELECTION GUIDE**

Section A LOUDSPEAKERS

High power (≥ 10 watts) to DIN45500 requirements for high fidelity speakers.

Cone diameter (inches)	Type No.	Shape of flange	Impedance versions (Ω)	Maximum power (W)	Type of use
1	AD0140/T	round	4, 8	20, 40	Tweeter
1	AD0162/T	round	8, 15	20, 4	Tweeter
1	AD0163/T	round	8, 15	20, 4	Tweeter
2	AD0210/Sq	round	4, 8	60	Squawker
2	AD0211/Sq	round	4, 8	60	Squawker
4	AD4050/W	round	4,8	15	Woofer
5	AD5060/Sq	octagonal	4,8	40	Squawker
5	AD5061/Sq	octagonal	4, 8	10	Squawker
7	AD7060/W	octagonal	4,8	30	Woofer
7	AD7066/W	octagonal	4,8	40	Woofer
8	AD8000	octagonal	-	-	Passive resonator
8	AD8061/W	octagonal	4,8	30	Woofer
8	AD8066/W	octagonal	4, 8	40	Woofer
8	AD8067/W	octagonal	4,8	40	Woofer
10	AD1065/W	round	4, 8	30	Woofer
10	AD10100/W	round	4, 8	40	Woofer

#### High power (≥ 10 watts) full range loudspeakers

5	AD5061/M	octagonal	4,8	10	7 litre enclosures
7	AD7062/M	octagonal	4,8	30	7 litre enclosures
7	AD7063/M	octagonal	4, 8	10	25 litre enclosures
81/2	9710/M8	round	8	20	Studio monitors,etc
10	AD1065/M	round	4, 8, 15	10	20 litre enclosures
12	AD1265/M	round	4, 8, 15	20	P.A. systems
12	AD12100/HP	round	4, 8	50	Guitar amplifier, electronic organs.
12	AD12100/M	round	4, 8, 15	25	50 litre enclosures

#### Medium power (2 - 10 watts) dual cone loudspeakers

Cone diameter (inches)	Type No.	Shape of flange	Impedance versions (Ω)	Maximum power
4 × 6	AD4681/M	oval	4, 8, 25	6
4 x 6	AD4691/M	oval	4, 8, 15, 25	4
5 x 7	AD5780/M	oval	4, 8, 15, 25	6
5 x 7	AD5790/M	oval	4, 8, 15	4
7	AD7080/M	octagonal	4, 8, 15	6
7	AD7091/M	octagonal	4,8	3

#### Medium power (2 - 10 watts) loudspeakers

Medidin power	2 - 10 watts/ loudspe	andis		
3 x 5	AD3591/X	oval	4, 8, 15, 25	3
3 x 5	AD3595/X	oval	4, 8, 15, 25	3
3 x 8	AD3880/X	oval	4, 8, 15	4
3 x 8	AD3890/X	oval	4, 8, 15, 25	4
4	AD4080/X	round	4, 8, 15, 25	3
4	AD4480/X	square	4, 8, 15, 25	3
4	AD4085/X	round	4, 8, 15	3
4	AD4485/X	square	4, 8, 15	3
4	AD4090/X	round	8, 15	2
4	AD4481/X4	square	4	8
4 x 6	AD4681/X	oval	4, 8, 15, 25	6
4 x 6	AD4691/X	oval	4, 8, 15, 25	4
4 x 8	AD4890/X	oval	4, 8, 15, 25	10
5 x 7	AD5780/X	oval	4, 8, 15, 25	6
5 x 7	AD5790/X	oval	4, 8	4
7	AD7080/X	octagonal	4, 8	6
7	AD7091/X	octagonal	4, 8	3

#### Low power (1 - 3 watts) plastic frame loudspeakers

The state of the s				
2½	AD2071/Z	round	4, 8, 15, 25	1
3	AD3071/Y	round	4, 8, 15, 25	2
3	AD3371/Y	square	4, 8, 15, 25	2
4	AD4072/X	round	4, 8, 15, 25	3
4	AD4472/X	square	4, 8, 15, 25	3

#### Crossover networks

Crossover frequency (Hz)	Type No.	Impedance versions (Ω)	Maximum power (W)	Tweeter impedance for high sensitivity tweeters $(\Omega)$
1500	ADF1500/4	4	80	8
1800	ADF 1500/8	8	80	15
2400	ADF2400/4	4	20	8
2400	ADF2400/8	8	20	15
650 and 2800	ADF700/2600/4	4	80	8
700 and 2600	ADF700/2600/8	8	80	15

Section B
TELEVISION TUNERS

Channel	Supply	voltage (V)	Power gain	Noise factor	Type No.
coverage	transistors	tuning diodes	(dB)	(dB)	
V.H.F.	+12	+0.3 to +28	20	7	ELC1042
V.H.F.	+12	+0.3 to +25	20	7	ELC1042/05
U.H.F. 21 – 69	+12	+0.3 to +25	22	7	ELC1043/05
U.H.F. 21 – 69	+12	+0.3 to +25	12	7	ELC1043/06
U.H.F./V.H.F, E2 to C E5 to E12 E21 to E69	+12	+0.5 to +28	29	6.5	ELC2000
U.H.F. E21 to E69	+12	+1 to +28	18	7.5	U321/321LO
U.H.F. E21 to E69	+12	+1 to +28	19	7.5	U322/322LO
V.H.F. NZ1 to E4 1A to 1C E5 to E12 1D to 1J	+12	+1 to +28	20	6.5	V311
V.H.F. E2 to C Morocco 4 to E12	+12	+1 to +28	20	6.5	V314
V.H.F. E2 to S1 S2 to (min) S17	+12	+1 to +28	20	6.5	V315/315LO

#### Section C

#### MONOCHROME TELEVISION ASSEMBLIES

(for use with 110° picture tubes with a neck diameter of 28 mm)

Description	Type No.
Deflection coil assembly	AT1040/15
Deflection unit	AT1074
Line output transformer	AT2048/11
Line linearity control unit (adjustable)	AT4042/02
Line linearity control unit (fixed)	AT4042/14

#### Industrial monochrome assemblies for monitors, V.D.U. etc.

Deflection unit	AT1071/01
Deflection unit	AT1074
Line output transformer	AT2102/01
Line output transformer	AT2140/10
Line linearity control unit (adjustable)	AT4034/01
Line linearity control unit (adjustable)	AT4036
Line linearity control unit (adjustable)	AT4042/08
Line driver transformer	AT4043/59

# Section D COLOUR TELEVISION ASSEMBLIES 1100 self converging system (20AX)

. Description	Type No.	
Deflection coil assemblies	AT1080, 1083/01, 1085	
Line output transformers	AT2080/10, 2080/15	
Line output transformers (diode split)	AT2076/35, 2076/55	
Multipole unit	AT1081	
Line linearity control unit	AT4042/38	
Line driver transformer	AT4043/29	
Twist compensation unit	AT4043/34	
Bridge coil	AT4043/38	
Balancing coil (E - W)	AT4044/20	
Balancing coil	AT4044/26	
Four pole adjusting coil	AT4044/27	
Switched mode power supply transformer	AT2095	

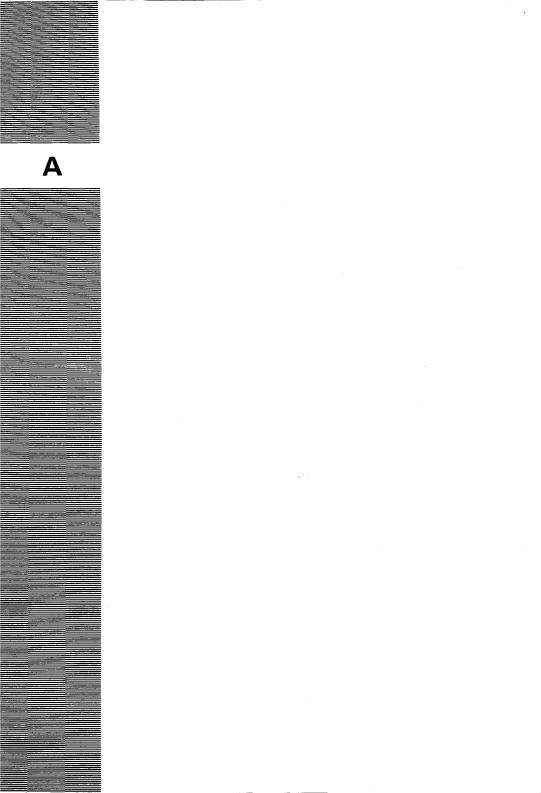
#### Delay lines and crystals

Chrominance delay line	DL50
Chrominance delay line	DL51
Chrominance delay line	DL60
Chrominance delay line	DL700
Luminance delay lines	VS340/1, 400/1, 470/1, 550/1, 600/1
4.4 MHz crystal	4322 152 01100
8.8 MHz crystal	4322 143 03120

#### Section E MODULES

Description	Type No.	
Voltage multiplying module	BG100	
Voltage multiplying module	LP1174 Series	
Voltage multiplying module	LP1194 Series	
Voltage multiplying module	LP1196 Series	





## INTRODUCTORY NOTES

#### INTRODUCTION

A correctly chosen loudspeaker is essential to obtain acoustic results from electro-acoustic equipment.

The following factors should be considered when choosing a loudspeaker:

- Shape, size and attachment with reference to the available space:
- Quality and sensitivity, a compromise between fidelity of reproduction and price:
- The frequency response characteristic in relation to the kind of application:
- Impedance and power handling capacity, which should be adapted to the output stage of the equipment:
- Appearance and finish.

To assist customers in making their choice our loudspeakers have been divided into three main groups:

- High power (≥ 10 watt): (hi-fi/full range)

- Medium power (2 - 10 watt)

- Low power (≤ 2 watt)

#### High power types

#### High power types: tweeters, woofers, squawkers

These speakers have been specially designed for use in hi-fi equipment, where a high power-handling capacity, a very wide frequency range and a negligible distortion level are required. They all conform to the high fidelity standards of IEC268 and DIN45500. Examples of application: sealed hi-fi enclosures with cross-over networks.

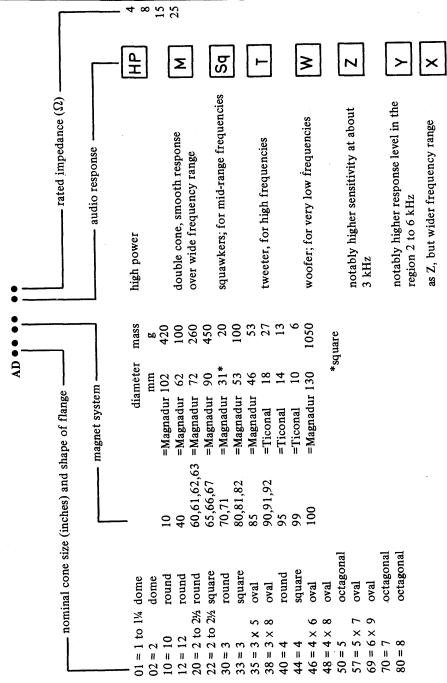
#### High power full range types

These types offer more than the medium and low power types and some of them meet the hi-fi standards of IEC268 and DIN45500. Examples of application: discotheques, public address systems, monitoring and hi-fi equipment in open or sealed enclosures.

#### Medium and low power types

The medium and low power speakers form an extensive group offering a diversity in characteristics, size and price for all kinds of radio and television sets, music centres, tape recorders, sound columns, etc.

Most of the medium and low power speakers contain a ferrite magnet (Magnadur). For television sets and other applications where the external magnetic field should be as small as possible, there are loudspeakers having a metal (Ticonal) magnet in a pot system.



# INTRODUCTORY NOTES

#### RESPONSE CURVES

For the medium and low power range one curve (a), showing the sound pressure as a function of the frequency is given in the Data sheets.

For the high power range the curves a, b and c are given, and for the squawkers and tweeters a directional response curve (d).

Measuring conditions concerning mounting of the loudspeaker:

			sound pressure		distortion
		curve a	curve b	curve d	curve c
range	measured in	anechoic room	half free field/ anechoic room	anechoic room	anechoic room
Medium /low power		unmounted			
High power full range		unmounted	baffle or enclosure		baffle or enclosure
High power	tweeters	unmounted	baffle	unmounted	baffle
range	squawkers	unmounted	baffle or enclosure	unmounted	baffle or enclosure
	woofers	unmounted	enclosure		enclosure

#### TERMS AND DEFINITIONS

- "Unmounted": The loudspeaker is placed in a clamping set-up which does not influence its radiation characteristics.
- "Mounted in enclosure": The loudspeaker with the gasket outside the enclosure of dimensions specified on the data sheet (flush mounted or front mounted as specified).
- "Baffle": The loudspeaker is fitted to a baffle, dimensions of which are specified on the data sheet (flush mounted or front mounted).
- "Half free field": The acoustical conditions on the forward side approach those of free space.
- "Anechoic room": The acoustical conditions approach those of free space. (IEC publication 268, part 5, section 1).
- "Operating power": is the sine-wave power input to the loudspeaker which corresponds with a sound level of 96 dB with respect to  $2 \times 10^{-4} \, \mu$ bar at a microphone distance of 1 m, or 86 dB on a distance of 3 m respectively. This sound level is the average level over the rated frequency range of the loudspeaker.



#### TEST METHODS AND MEASUREMENTS

The atmospheric conditions for measurement are:

Temperature

15 to 35 °C Relative humidity: 45 to 75 %

Pressure:

860 to 1060 mbar

#### Impedance

The impedance is the modulus of the lowest value of the electrical impedance in the frequency range above the bass resonance frequency of the loudspeaker as determined by the method specified in para. 3 below.

#### 1. 1 Measuring apparatus

- Audio frequency sinewave signal generator with a constant output voltage over the range from 0 to 20 000 Hz.
- Linear amplifier with an output impedance not greater than 1/3 of the rated loudspeaker impedance and a power output of approx. 0.1 × the power handling capacity of the loudspeaker.
- A 1Ω resistor connected in series with the loudspeaker.
- An electronic voltmeter shunted across the 1  $\Omega$  resistor.

#### 1.2 Conditions

- The loudspeaker is unmounted.
- The power input to the loudspeaker will not exceed 0.1 x the power handling capacity as determined in para. 4 below.

#### 1.3 Measuring result

Rated impedance is stated on the data sheets. The measured impedance will not be lower than 20% of the rated impedance.

#### Voice coil resistance

The voice coil resistance is the (d.c.) resistance of the voice coil.

#### 2. 1 Measuring apparatus

Low current d.c. Ohm-meter.

#### 2. 2 Conditions

The d.c. power input to the loudspeaker does not exceed 0.1 x the power handling capacity.

#### 2. 3 Measuring circuit

The rated resistance is given on the data sheets, tolerance ± 10%



#### INTRODUCTORY NOTES

#### 3 Resonance frequency

The resonance frequency is that frequency where the modulus of the electrical impedance has its first principal maximum in an ascending scale, the electrical input being such as to have no significant effect on the resonant frequency.

#### 3. 1 Measuring apparatus

Same as for "Impedance". See para. 1.

#### 3. 2 Conditions

- The loudspeaker is measured unmounted.
- The resonance frequency is determined after applying to the loudspeaker for a duration of 5 s a test signal equal to that required to test the power handling capacity.

#### 3. 3 Measuring result

The resonance frequency is that frequency at which the voltmeter indicates the first minimum deflection as the frequency is swept slowly from 0 Hz, the output voltage of the amplifier being such that the voltmeter reads for the resonance frequency:

40 to 60 mV for loudspeakers with a rated impedance less than 20  $\Omega$ :

15 to 25 mV for loudspeakers with a rated impedance between 20  $\Omega$  and 100  $\Omega$ :

4 to  $\,$  6 mV for loudspeakers with a rated impedance greater than 100  $\Omega.$ 

The rated resonance frequency is stated on the data sheets.

#### 4 Power handling capacity

The power handling capacity is the nominal power which the loudspeaker will satisfactorily handle as checked by an accelerated life test.

#### 4.1 Test apparatus

- Generator supplying test signal in accordance with IEC268, para. 9.3.
- Power amplifier with an output impedance not greater than 1/3 of the rated impedance of the loudspeaker.
- Voltmeter indicating the r.m.s. value of the voltage.

#### 4.2 Conditions

- A test voltage is applied to the loudspeaker for an uninterrupted period of 100 hrs. The r.m.s. value of this voltage corresponds with the specified power handling capacity of the loudspeaker.
- The test voltage has a frequency distribution corresponding with that of the output of a filter as specified in IEC Publication 268, part 5 para. 9.3 when fed from a white noise source.
- If the loudspeaker is designed to operate in a restricted frequency range, the corresponding network (filter) which is connected to the loudspeaker during the test, is specified on the data sheet. The test voltage is measured at the input terminals of the network.
- The method of mounting is as specified on the data sheet.

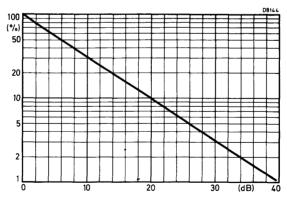
#### 4. 3 Test result

To pass this test the loudspeaker has to function properly at the end of the test period. Deviation from the specified resonance frequency is allowed.

#### 5 Total non-linear distortion

This is the ratio between the r.m.s. value of the harmonic content of the sound pressure to the value of the total sound pressure over the frequency range of the loudspeaker.

The difference in dB between fundamentals and harmonic contents, can be converted into a distortion percentage with the aid of the following nomogram.



Difference in dB converted into % distortion

# INTRODUCTORY NOTES

#### 5.1 Conditions

- The loudspeaker is mounted as specified on the data sheet.
- The power input to the loudspeaker is the operating power.
- The microphone distance is as specified on the data sheet. (See also definition of "Operating power").

#### 5.2 Measuring result

The distortion curve with its limit of high power loudspeakers is given on the data sheet.

#### 6 Sweep voltage

The sweep voltage test imposes on the loudspeaker a sinusoidal test signal of specified constant amplitude. The frequency of this signal is swept through the specified frequency range.

#### 6. 1 Test apparatus

- Audio frequency sinewave signal generator with a constant output voltage over the range from 0 to 20 000 Hz.
- Linear amplifier with an output power appropriate to the loudspeaker under test and an output impedance not greater than  $1/3 \times$  the rated loudspeaker impedance. For power see 6.2.
- An electronic voltmeter with high input impedance.

#### 6.2 Conditions

- The loudspeaker is tested unmounted.
- The input voltage is
  - a) for the medium and low power range such that the power input to the loudspeaker is 0.5 × the specified power handling capacity.
  - b) for the high power range as specified on the data sheets.
- If the loudspeaker is designed to operate in a restricted frequency range, the corresponding network (filter) which is connected to the loudspeaker during the test, is specified on the data sheet. The test voltage is measured at the input terminals of the network.

#### 6. 3 Test result

To pass this test the loudspeaker has to function properly during the test.

#### 7 Flux density

This is the magnetic flux density measured in the air gap.

#### 7.1 Measuring apparatus

- Differential search coil
- Galvanometer

#### 7. 2 Conditions

- The distance between the centres of the two coils is equal to the air gap height minus 1 mm.
- The two coils are put into the air gap symmetrical with respect to the poleplate.

#### 7. 3 Measuring result

The minimum flux density as measured on production samples is stated on the data sheet.

#### 8 Frequency response

The frequency response is the graph representing the sound pressure as a function of frequency applying to the loudspeaker a constant sine-wave signal V.

#### 8. 1 Measuring apparatus

#### The apparatus is set as follows:

- Writing speed	125 mm/s	
- Paper speed	3 mm/s	
- Range potentiometer	50 dB	
- Lower limiting frequency	10 Hz	
- Rectifier response	r.m.s.	
- Writing width	100 mm	
- Compressor speed	$300  \mathrm{dB/s}$	

#### 8.2 Conditions

- Sine-wave signal  $V = \sqrt{W. Z_r}$
- for anechoic room measurements W = 50 mW, unless otherwise stated on the data sheets.

V = test voltage

 $Z_r$  = rated impedance as specified on the data sheet

- Microphone position: in axis of loudspeaker on a distance of 50 cm for anechoic room measurements
- Curve a is measured in a anechoic room; loudspeaker unmounted
- Curve b is measured in a half free field; loudspeaker mounted as specified on the data sheet
- Curve d is measured in a anechoic room; loudspeaker unmounted,



# INTRODUCTORY NOTES

#### 8. 3 Measuring result

A description of the sensitivity and the frequency response curve(s), together with the limits for curve a are given on the data sheet.

#### 9 Direction of magnetisation

The magnet is so magnetised that the centrepole is <u>south</u> for systems with a ring magnet, and <u>north</u> for systems with a slug magnet.

#### 10 Polarity

The cone of the loudspeaker will move outward when a d.c. voltage is applied to the terminals so that the red marked terminal is positive.

The voltage applied does not exceed the "sweep voltage".



## 1 inch HIGH POWER DOME TWEETER LOUDSPEAKER

#### APPLICATION

For the reproduction of audio frequencies from  $1600~\rm{Hz}$  to  $22\,000~\rm{Hz}$  in multi-way high-fidelity loudspeaker systems. Minimum recommended cross-over frequency 1600 Hz with  $12~\rm{dB/octave}$  slope.

TECHNICAL DATA	version		
	T4	Т8	
Rated impedance	4	8	Ω
Voice coil resistance	3, 4	6, 3	Ω
Rated frequency range	1600 to	20 000	Hz
Resonance frequency	1200 I		Hz
Power handling capacity, measured with filter: $12~\mu F - 0$ , $35~mH$ $5~\mu F - 0$ , $2~mH$ $8~\mu F - 0$ , $5~mH$ $3$ , $2~\mu F - 0$ , $35~mH$ loudspeaker unmounted	20 40		W W W
Operating power		4	W
Sweep voltage (500 to 20 000 Hz)	3	4,5	v
Energy in air gap	5	9	mJ
Flux density	0,	9	T
Air-gap height	2,	5	mm
Voice coil height	2, 4	3, 2	mm
Core diameter	2	5	mm
Magnet material diameter mass	6	Magnadur           61         n           0, 1         k	
Mass of loudspeaker	0, 2	5	kg

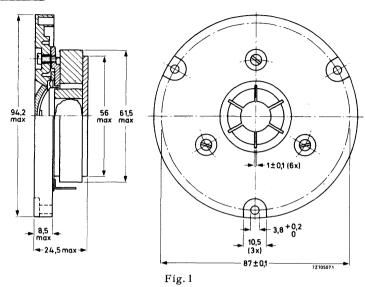
The loudspeaker has a polycarbonate dome and a voice coil of aluminium wire.

Connection to the loudspeaker is by means of 3.2 mm (0.12 inch) tag connectors or by soldering.



## AD0140/T.

#### Dimensions (mm)



One tag is indicated by a red mark for in-phase connection

Baffle hole diameter 75 mm.

Face of loudspeaker should lie in line with plane of baffle.

#### FREQUENCY RESPONSE CURVES

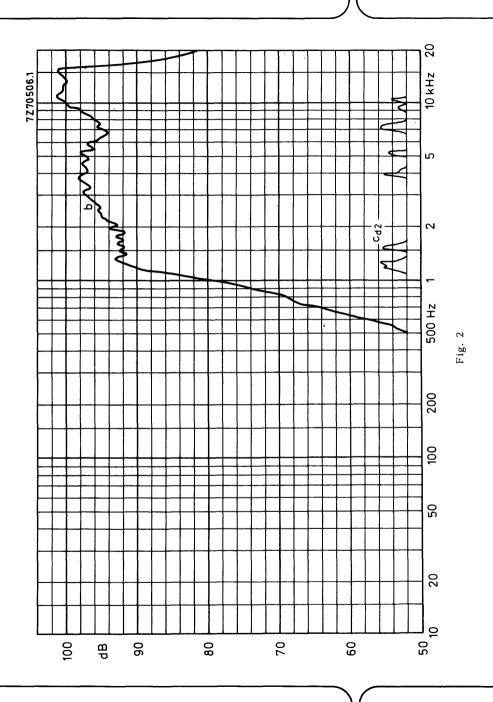
Curve b: Sound pressure measured in half free field, input at operating power.

**Mullard** 

Loudspeaker mounted on baffle, dimensions 50 x 50 mm.

Curve c: 2nd and 3rd harmonic distortion, measured at the operating power of 4 W in

anechoic room. Loudspeaker unmounted.





# 1 INCH HIGH POWER DOME TWEETER LOUDSPEAKER

#### APPLICATION

For use in direct and indirect radiating systems for reproduction of audio frequencies from 2000 Hz to 22 000 Hz with very low distortion in multi-way high fidelity loudspeaker systems in accordance with DIN45500. Minimum recommended cross-over frequency 1600 Hz. The loudspeaker has a very high sensitivity.

#### **TECHNICAL DATA**

	version		
	T8	T 15	
Rated impedance	8	15	Ω
Voice coil resistance	6,3	12,5	Ω
Rated frequency range	2000 to 22 000		H2
Resonance frequency	1000		Hz
Power handling capacities a/b (see Fig. 1) at 2000 Hz C = $8 \mu F$ L = 0,5 mH C = 3,3 $\mu F$ L = 1 mH at 4000 Hz C = 3,2 $\mu F$ L = 0,35 mH C = 1,5 $\mu F$ L = 0,8 mH	20/4 50/6	20/4 50/6	W W W
Operating power		2	W
Sweep voltage frequency range: $500$ - $20000$ Hz high pass filter: $8\mu\text{F}$ - $0.5$ mH	4,5	5,5	v
Energy in air gap		75	
Flux density		1,2	T
Air gap height		2,5	mm
Voice coil height	2,4	3,4	mm
Core diameter		25	mm
Magnet material diameter mass		72 0,24	mm kg
Mass of loudspeaker		0,5	kg

The loudspeaker has a polycarbonate dome and a diffusor integrated in the cover.

Connection to the loudspeaker by means of 2, 8 mm (0, 11 inch) Fastons or soldering.

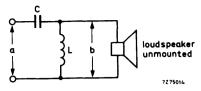


Fig. 1. Measuring circuit.

a = system power handling capacity

b = loudspeaker power handling capacity.

#### Dimensions (mm)

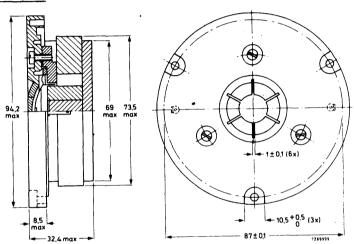


Fig. 2.

One tag is indicated by a red mark for in-phase connection. Face of loudspeaker should not lie behind plane of baffle.

#### FREQUENCY RESPONSE CURVES (see Fig. 3)

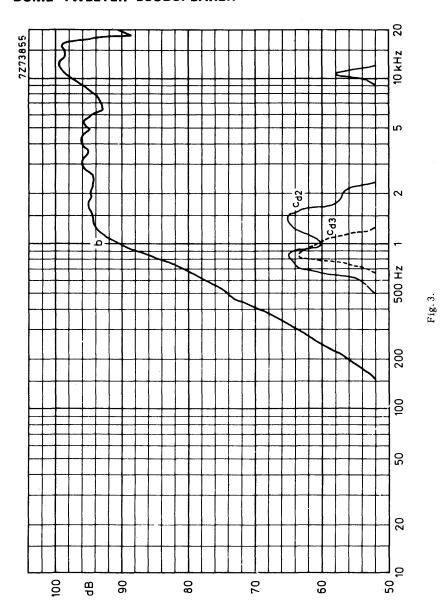
Curve b: Sound pressure measured in anechoic room, loudspeaker unmounted.

Above 1000 Hz, over the width of one octave, the sound pressure may be a maximum of 2 dB lower than indicated.

Curve c: 2nd and 3rd harmonic distortion, measured at the operating power of 2 W in anechoic room, loudspeaker unmounted.

# 1 INCH HIGH POWER DOME TWEETER LOUDSPEAKER

## AD0162 /T.



- Mullard



#### 1 INCH HIGH POWER DOME TWEETER LOUDSPEAKERS

#### **APPLICATION**

...........

For use in direct and indirect radiating systems for reproduction of frequencies from 2000 Hz to 22 000 Hz with very low distortion in multi-way high-fidelity loudspeaker systems in accordance with DIN 45500. Minimum recommended crossover frequency 2000 Hz. The loudspeaker has a very high sensitivity.

TECHNICAL DATA	version		
	T8	T15	
Rated impedance	8	15	Ω
Voice coil resistance	6,3	12,5	Ω
Rated frequency range	200	0 to 22 000	Hz
Resonance frequency	1300		Hz
Power handling capacities, a/b (see Fig.1), loudspeaker unmounted,			
at 2000 Hz; C = 8 µF; L = 0,5 mH at 2000 Hz; C = 3,3 µF; L = 1 mH	20/4	20/4	W W
at 4000 Hz; C = 3,2 µF; L = 0,35 mH at 4000 Hz; C = 1,5 µF; L = 0,8 mH	50/6	50/6	W W
Operating power		2	W
Sweep voltage, frequency range: 500 to 20 000 Hz high pass filter: $8 \mu F - 0.5 \text{ mH}$ 3,3 $\mu F - 1 \text{ mH}$	4,5	5,5	v v
Energy in air gap		75	mJ
Flux density		1,2	T
Air-gap height		2,5	mm
Voice coil height	2,4	3,4	mm
Core diameter		25	mm
Magnet material diameter mass	M	Magnadur 72 m 0,24 k	
Mass of loudspeaker		0,5	kg

The loudspeaker has an impregnated textile dome and a diffuser integrated in the cover. Connection to the loudspeaker by means of 2,8 mm (0,11 inch) Fastons or soldering.



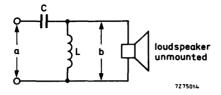


Fig.1 Measuring circuit.

a = system power handling capacity.

b = loudspeaker power handling capacity.

#### Dimensions (mm)

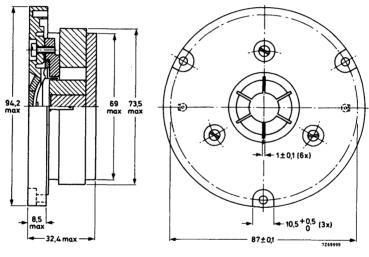


Fig.2

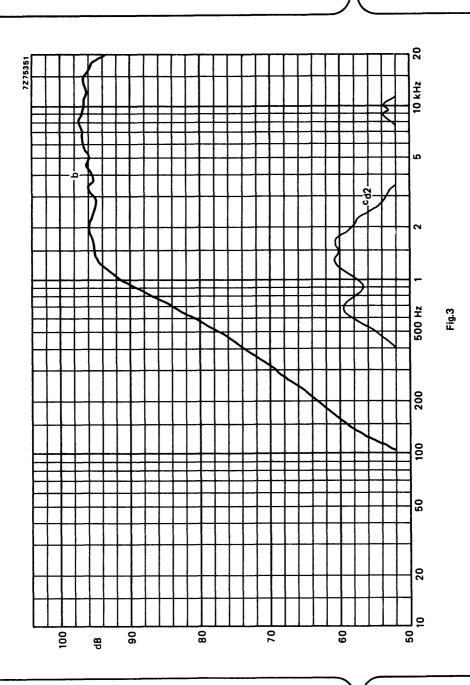
One tag is indicated by a red mark for in-phase connection. Face of loudspeaker should not lie behind plane of baffle.

#### FREQUENCY RESPONSE CURVES (see Fig.3)

May 1977

Curve b: Sound pressure measured in anechoic room, loudspeaker unmounted.

Curve c: 2nd and 3rd harmonic distortion, measured at the operating power of 2 W in anechoic room, loudspeaker unmounted.





#### APPLICATION

For the reproduction of audio frequencies from 500 to 5000 Hz with very low distortion in multi-way high-fidelity loudspeaker systems according to DIN45500. The loudspeaker has an excellent spherical radiation pattern.

TECHNICAL DATA	version		
	Sq4	Sq8	
Rated impedance	4	8 (3	
Voice coil resistance	3, 4	6,6 Ω	
Resonance frequency	370	Hz	Z
Rated frequency range	550 to 5000		Z
Power handling capacity, measured with filter 36 μF - 1,2 mH 18 μF - 2,4 mH	60	W 60	
loudspeaker unmounted		00	
Power handling capacity of speaker only	20	w	
Operating power	5	w	
Sweep voltage (100 to 10 000 Hz, filter 36 $\mu F$ = 1, 2 mH $_{18}~\mu F$ = 2, 4 mH)	4,5	V 6, 3 V	
Energy in air gap	. 0	mJ	J
Flux density	0,8	T	
Air-gap height	5	mn	m
Voice coil height	3, 3	3,6 mr	m
Core diameter	50	mn	m
Magnet material	Magna	der	
diameter mass	102 0, 42		
Mass of loudspeaker	1	kg	,

The loudspeaker has a paper dome, textile rim and a sealed pot; no acoustic isolation required.

Connection to the loudspeaker is by means of 5,1 mm (0, 2 inch) Fastons or soldering.

### Dimensions (mm)

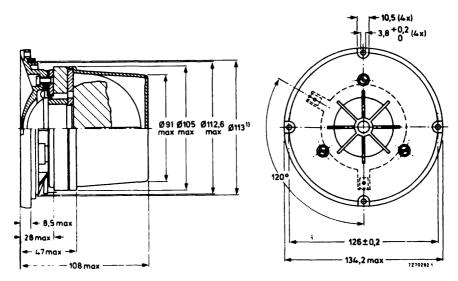


Fig. 1

One tag is indicated by a red mark for in-phase connection.

### FREQUENCY RESPONSE CURVES

See Fig. 2 Input power 50 mW

Curve b: Sound pressure measured in anechoic room, loudspeaker mounted on DIN baffle at operating power.

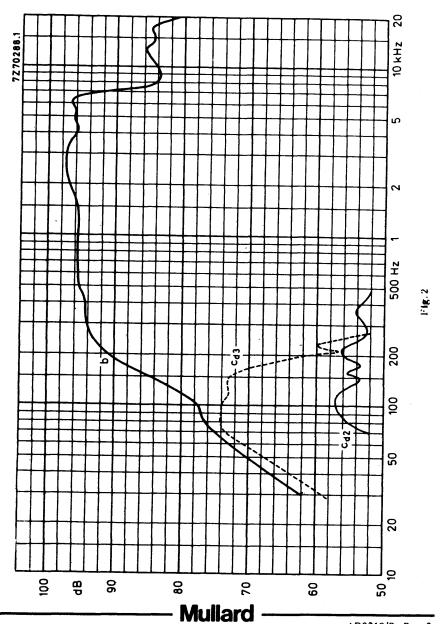
Curve c: 2nd and 3rd harmonic distortion, measured at the operating power of 5 W in anechoic room. Loudspeaker front mounted on DIN baffle.



<sup>1)</sup> Baffle hole diameter 110 mm

# 2 inch HIGH POWER DOME SQUAWKER LOUDSPEAKER

AD0210/Sq.



AD0210/Sq Page 3

			_	

## 2 INCH HIGH POWER DOME SQUAWKER LOUDSPEAKER

### APPLICATION

For the reproduction of audio frequencies from 500 to 5000 Hz with very low distortion in multi-way high-fidelity loudspeaker systems according to DIN 45500. The loudspeaker has an excellent spherical radiation pattern.

TECHNICAL DATA		version				
	Sq4	Sq	8			
Rated impedance	4		8 Ω			
Voice coil resistance:	3,4	6,	6 Ω			
Resonance frequency	340	37	0 Hz			
Rated frequency range	550	to 500	0 Hz			
Power handling capacity, loudspeaker unmounted, measured with filter 50 $\mu F - 1.6$ mH 24 $\mu F - 3.2$ mH	60	6	w o w			
Power handling capacity of speaker only		20	W			
Operating power		5	W			
Sweep voltage (100 to 10 000 Hz, filter 50 $\mu$ F $-$ 1,6 mH 24 $\mu$ F $-$ 3,2 mH)	4,5	6,	V 3 V			
Energy in air gap	;	250	mJ			
Flux density		0,8	Т			
Air-gap height		5	mm			
Voice coil height	3,3	3,	6 mm			
Core diameter		50	mm			
Magnet material	Mag	madur				
diameter mass		102 ,42	mm kg			
Mass of loudspeaker		1	kg			

The loudspeaker has a textile dome and surround, and a sealed pot; no acoustic isolation required. Connection to the loudspeaker is by means of 5,1 mm (0,2 inch) Fastons or soldering.



Dimensions (mm) - 10,5 (4x) Ø91 Ø105 Ø112,6 Ø113\* max max max 1200 8,5 max 28 max -126±0,2 - 47 max → 134,2 max – 108 max 7270292.1

Fig. 1

One tag is indicated by a red mark for in-phase connection.

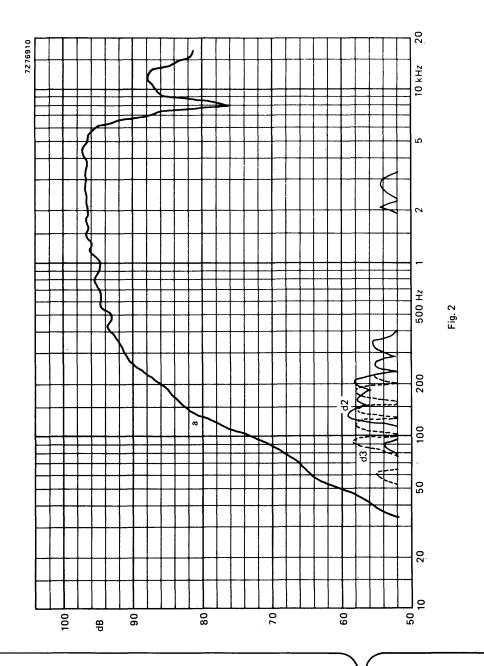
### FREQUENCY RESPONSE CURVES (see Fig. 2)

Curve a: Sound pressure measured in anechoic room, loudspeaker mounted on IEC baffle at operating power.

Curves d2 and d3: 2nd and 3rd harmonic distortion, measured at the operating power in anechoic room. Loudspeaker front mounted on IEC baffle.

**Mullard** 

<sup>\*</sup> Baffle hole diameter 110 mm.





## 21/2 INCH LOW POWER LOUDSPEAKER

### **APPLICATION**

For portable receivers and intercoms.

TECHNICAL DATA			versi	on	
	<b>Z4</b>	Z8	Z15	Z25	
Rated impedance	4	8	15	25	9
Voice coil resistance	3,5	7,1	13,7	22,8	5
Rated frequency range		1	80 to 40	000	ı
Resonance frequency			360		ŀ
Power handling capacity, loudspeaker unmounted, measured without filter			1		١
Operating power (sound level 90 dB, 0,5 in)			0,25		
Sweep voltage (frequency range: 240 to 15000 Hz)	1	1,4	1,9	2,5	١
Energy in air gap			12,7		r
Flux density			0,74		7
Air-gap height			2,5		r
Voice coil height	2,7	2,2	3,0	3,6	r
Core diameter			10		
Magnet material			Magnad	ur	
diameter			31		п
mass			0,02		k
Mass of loudspeaker			0,064		k

The loudspeaker has a plastic frame, and a paper cone and surround. Connection to the loudspeaker by means of 2,8 mm (0,11 inch) Fastons or soldering.



Dimensions (mm)

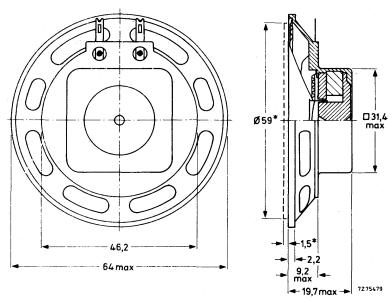


Fig.1.

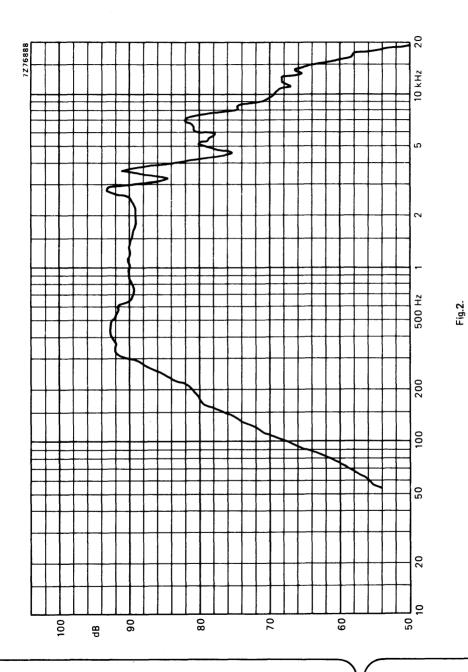
\*Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

One tag is indicated by + sign for in-phase connection.

FREQUENCY RESPONSE CURVE (see Fig. 2)

Sound pressure measured in anechoic room, loudspeaker mounted on IEC baffle.

Mullard





**Mullard** 



## 3 INCH LOW POWER LOUDSPEAKER

### **APPLICATION**

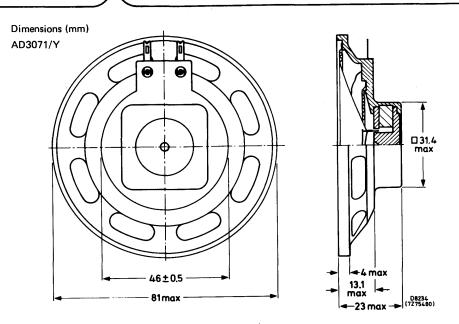
For portable receivers and intercoms.

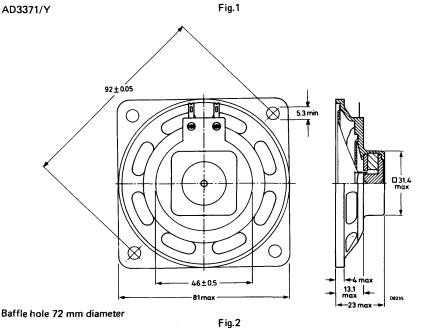
### **TECHNICAL DATA**

	version				
	Y4	Y8	Y15	Y25	
Rated impedance	4	8	15	25	$\Omega$
Voice coil resistance	3.5	7.1	13.7	22.8	Ω
Rated frequency range	1	00 to 6	000		Hz
Resonance frequency		250			Hz
Power handling capacity, loudspeaker unmounted, measured without filter			2		w
Operating power (sound level 90 dB, 0.5 in)			0.225		w
Sweep voltage (frequency range 170 to 15 000 Hz)	1.4	2.0	2.7	3.5	٧
Energy in air gap			12.7		mJ
Flux density			0.74		Т
Air-gap height			2.5		mm
Voice coil height	2.7	2.2	3.0	3.6	mm
Core diameter			10		mm
Magnet material	Magnadur				
diameter			31		mm
mass			0.02		kg
Mass of loudspeaker			0.069		kg

The loudspeaker has a plastic frame, and a paper cone and surround. Connection to the loudspeaker by means of 2.8 mm (0.11 inch) Fastons or soldering.

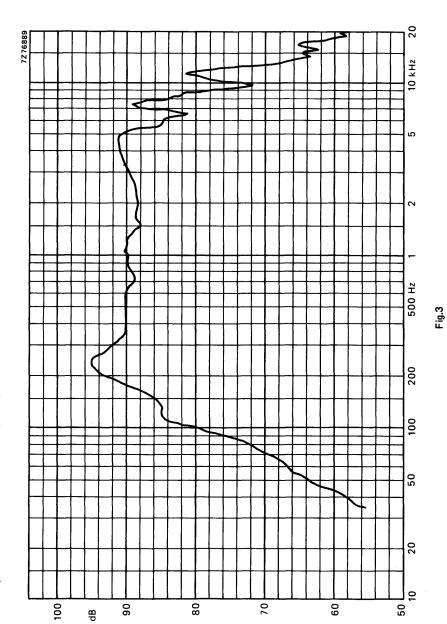






1 19.

FREQUENCY RESPONSE CURVE Sound pressure measured in anechoic room, loudspeaker mounted on IEC baffle.





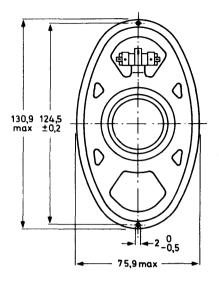
### **APPLICATION**

For use in portable radios, tape recorders and, due to absence of stray magnetic field, this loudspeaker can also be used in television sets. High sensitivity.

TECHNICAL DATA		ve	ersion		
	X4	X8	X15	X25	
Rated impedance	4	8	15	25	
Voice coil resistance	3,4	7,2	13,8	22,6	
Rated frequency range			85 to 120	000	
Resonance frequency			180		
Power handling capacity, measured without filter, loudspeaker unmounted			3		
Sweep voltage	2,4	3,5	4,7	6,1	
Energy in air gap			39		
Flux density			8,0		
Air-gap height			3		
Voice coil height	2	1,8	2,55	2,8	
Core diameter			18		
Magnet material diameter mass	Ticonal	Ticonal	Ticonal 18 0,027	Ticonal	
Mass of loudspeaker			0,13		

The loudspeaker has a paper cone and surround.

Connection to the loudspeaker by means of 2.8 mm (0.11 inch) tag connectors or by soldering.



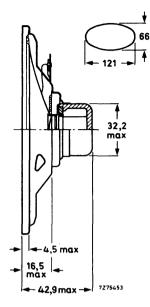


Fig. 1.

One tag is indicated by a red mark for in-phase connection.

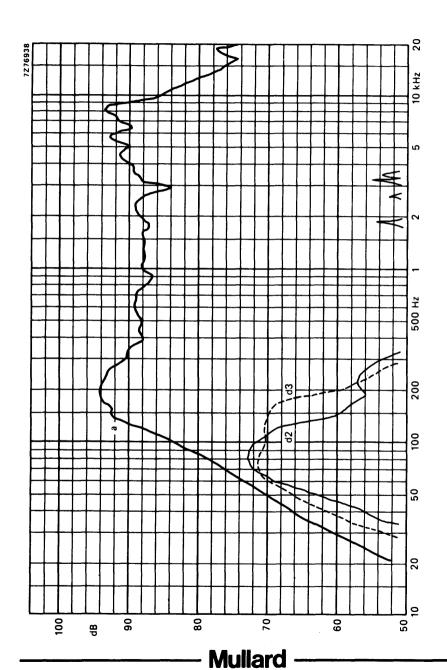
### FREQUENCY RESPONSE CURVE (see Fig. 2)

Curve a: Sound pressure.

Curves d2 and d3: 2nd and 3rd harmonic distortion.

The curves are measured in anechoic room, loudspeaker mounted on IEC baffle.

## AD3591/X



ig. 2.

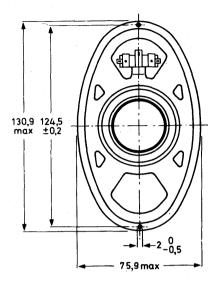
## 3 x 5 INCH MEDIUM POWER LOUDSPEAKER

### APPLICATION

For use in portable radios, tape recorders and, due to absence of stray magnetic field, this loudspeaker can also be used in television sets. High sensitivity,

TECHNICAL DATA	version				
	X4	Х8	X15	X25	
Rated impedance	4	8	15	25	Ω
Voice coil resistance	3,4	7,1	13,5	22,5	Ω
Rated frequency range		90 to	15000		Hz
Resonance frequency	180	170	170	170	Hz
Power handling capacity, measured without filter, loudspeaker unmounted			3		w
Operating power (sound level 90 dB, 1 m)		1,5			W
Sweep voltage, frequency range: 80 to 20 000 Hz	2,5	3,5	4,7	6,2	V
Energy in air gap		20	,5		mJ
Flux density		0,	77		T
Air-gap height		2	2,5		mm
Voice coil height	3,5	4,2	2,7	3,3	mm
Core diameter		14	,5		mm
Magnet material diameter mass		Ticor 14 0,0	,5		mm kg
Mass of loudspeaker		0,	13		kg

The loudspeaker has a paper surround and a foam plastic gasket on the flange. Connection to the loudspeaker by means of 2,8 mm (0,11 inch) Fastons, or soldering.



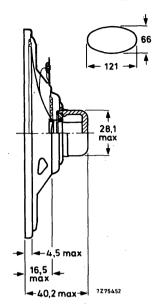


Fig. 1.

One tag is indicated by a red mark for in-phase connection.

### FREQUENCY RESPONSE CURVES (see Fig. 2)

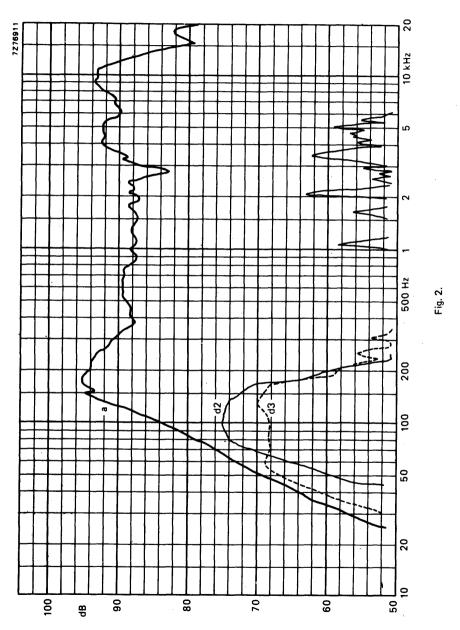
Curve a: Sound pressure.

Curves d2 and d3: 2nd and 3rd harmonic distortion, measured at the operating power of 1,5 W.

The curves are measured in anechoic room, loudspeaker mounted on IEC baffle.



## AD3595/X



Mullard -

### APPLICATION

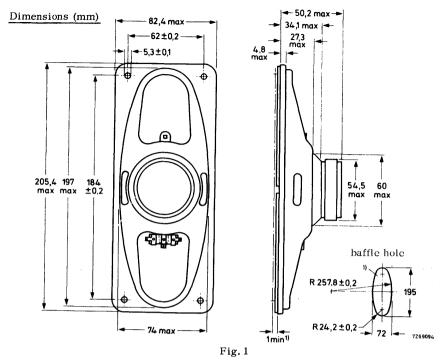
For use in portable radios and tape recorders

### TECHNICAL DATA

	version			
	X4	X8	X 15	
Rated impedance	4	8	15	Ω
Voice coil resistance	3,4	7, 1	13, 8	Ω
Resonance frequency	120	120	120	Hz
Power handling capacity, measured without filter, loudspeaker unmounted	4	4	4	W
Sweep voltage	2,8	5,5	5,5	v
Energy in airgap	55	55	55	mJ
Flux density	1	1	1	T
Airgap height	3	3	3	mm
Voice coil height	2,4	3, 1	2,5	mm
Core diameter	18	18	18	mm
Magnet material		Magnadur		
diameter	53	53	53	mm
weight	0, 1	0, 1	0,1	kg
Weight of loudspeaker	0,3	0,3	0,3	kg

The loudspeaker has a paper cone and a treated paper surround.

Connection to the loudspeaker by means of 2.8 mm (0.11 inch) tag connectors or by soldering.



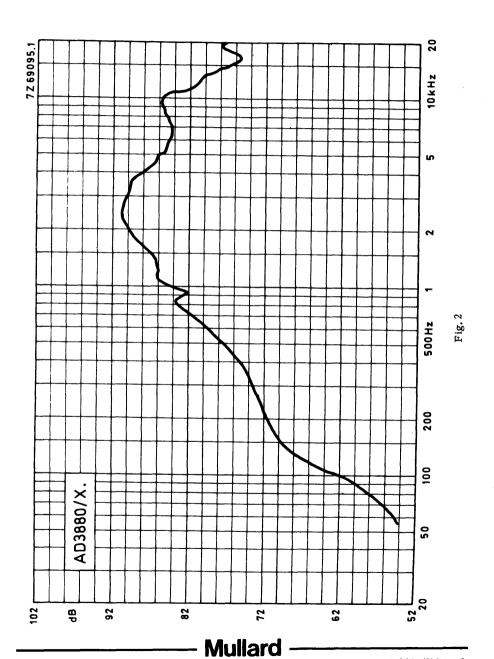
l) Baffle hole and clearance depth required for cone movement at specified power handling capacity.

One tag is indicated by a red mark for in-phase connection.

### FREQUENCY RESPONSE CURVE

See Fig. 2. Input power 50 mW

Sound pressure measured in anechoic room, loudspeaker unmounted. Above 1000 Hz the sound pressure may be, over the width of one octave, maximum 2 dB lower than indicated.



## 3 x 8 INCH OVAL MEDIUM POWER LOUDSPEAKERS

AD3890/X Series

#### APPLICATION

For use in portable radios and tape recorders.

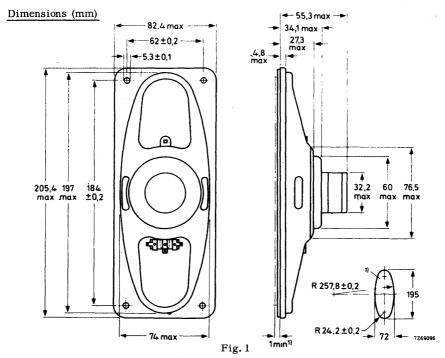
The absence of stray field due to ticonal sinterpot magnet system, makes this loudspeaker also suitable for use in television sets.

### TECHNICAL DATA

	version					
	X4	X8	X 15	X25		
Rated impedance	4	8	15	25	Ω	
Voice coil resistance	3,4	7, 1	13,5	22,7	Ω	
Resonance frequency	120	120	120	120	Hz	
Power handling capacity, measured without filter, loudspeaker unmounted	4	4	4	4	w	
Sweep voltage	2,8	4	5,5	7, 1	v	
Energy in airgrap	39	39	39	39	mJ	
Flux density	0.8	0,8	0,8	0,8	Т	
Airgap height	3	3	3	3	mm	
Voice coil height	2,4	2,8	2,5	2,8	mm	
Core diameter	18	18	18	18	mm	
Magnet material diameter weight	Ticonal 18 0,027	Ticonal 18 0,027	Ticonal 18 0,027	Ticonal 18 0,027	mm kg	
Weight of loudspeaker	0,21	0,21	0,21	0,21	kg	

The loudspeaker has a paper cone and a treated paper surround.

Connection to the loudspeaker by means of 2.8 mm (0.11 inch) tag connectors or by soldering.



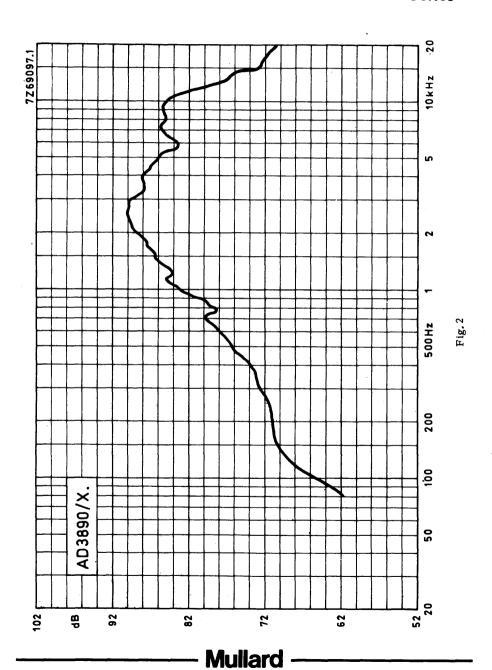
1) Baffle hole and clearance depth required for cone movement at specified power handling Capacity.

One tag is indicated by a red mark for in-phase connection.

### FREQUENCY RESPONSE CURVE

See Fig. 2. Input power 50 mW

Sound pressure measured in anechoic room, loudspeaker unmounted. Above 1000 Hz the sound pressure may be, over the width of one octave, maximum 2 dB lower than indicated.



## 4 INCH HIGH POWER WOOFER LOUDSPEAKER

#### APPLICATION

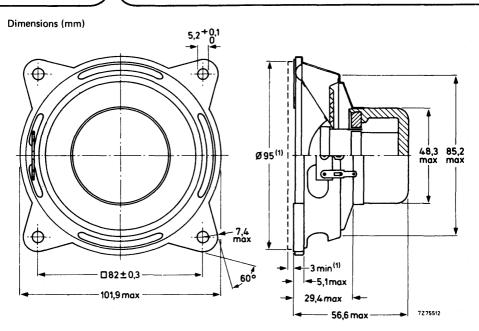
The absence of stray field due to Ticonal sinterpot magnet system makes this loudspeaker very suitable for use in television sets. It can be used in sealed acoustic enclosures and in bass reflex enclosures of maximum 7 litres.

### **TECHNICAL DATA**

		version		
	W4		W8	_
Rated impedance	4		8	Ω
Voice coil resistance	3.2		7	$\Omega$
Rated frequency range		35 to 2000		Hz
Resonance frequency		60		Hz
Power handling capacity measured without filter, mounted in 7 1 bass reflex enclosure		15		w
Operating power		8		W
Sweep voltage, frequency range; 30 to 6000 Hz	5.5		7.75	٧
Energy in air gap		100		mJ
Flux density		0.85		T
Air-gap height		5		mm
Voice coil height		6		mm
Core diameter		25		mm
Magnet material diameter mass		Ticonal 25 0.06		mm kg
Mass of loudspeaker		0.42		kg

The loudspeaker has a rubber surround and a sealing strip at the rear of the basket. Connection to the loudspeaker by means of 2.8 mm (0.11 inch) Fastons, or soldering.





(1) Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

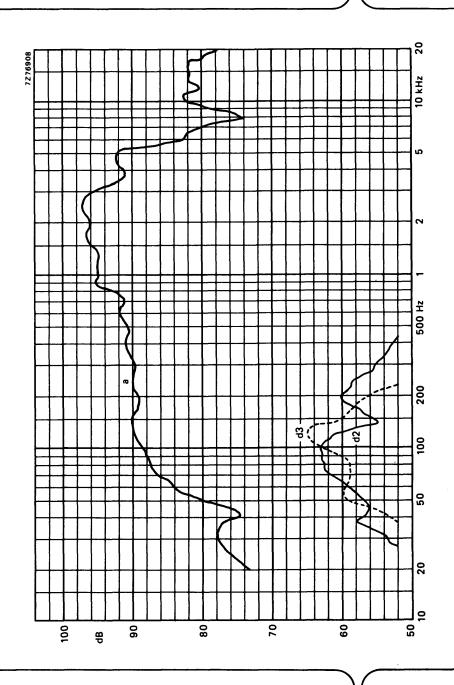
One tag is indicated by a red mark for in-phase connection.

### FREQUENCY RESPONSE CURVES (see Fig. 2)

Curve a: Sound pressure measured in anechoic room, loudspeaker mounted in 7 l bass reflex enclosure. Curves d2 and d3: 2nd and 3rd harmonic distortion, measured at the operating power.

**Mullard** 

AD4050/W





## 4 INCH MEDIUM POWER LOUDSPEAKERS

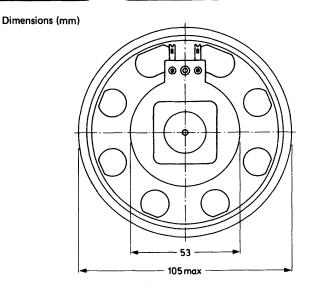
### **APPLICATION**

For portable receivers and intercoms.

TECHNICAL DATA	version				
	X4	X8	X15	X25	
Rated impedance	4	8	15	25	Ω
Voice coil resistance	3,5	7,1	13,7	22,8	Ω
Rated frequency range		80 to	15 000		Hz
Resonance frequency		1	70		Hz
Power handling capacity, loudspeaker unmounted, measured without filter			3		W
Operating power (sound level 90 dB, 0,5 in)		0,	18		W
Sweep voltage (frequency range 100 to 20 000 Hz)	1,4	2	2,7	3,5	V
Energy in air gap		12	2,7		mJ
Flux density		0,	74		T
Air-gap height		2	2,5		mm
Voice coil height	2,7	2,2	3,0	3,6	mm
Core diameter		1	10		mm
Magnet material		Mag	ıadur		
diameter			31		mm
mass		0,0	02		kg
Mass of loudspeaker,					
round flange version		0,0			kg
square flange version		0,0	5/		kg

The loudspeakers have a plastic frame, and a paper cone and surround. Connection to the loudspeakers is by means of 2,8 mm (0,11 inch) Fastons or soldering.





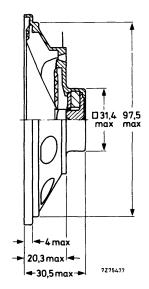
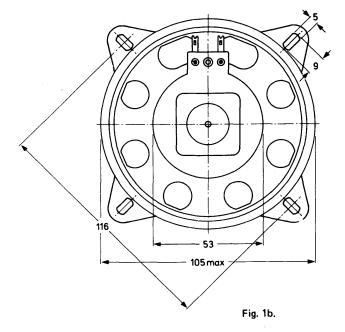
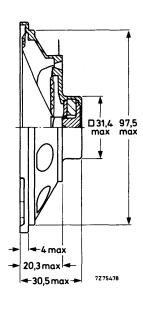
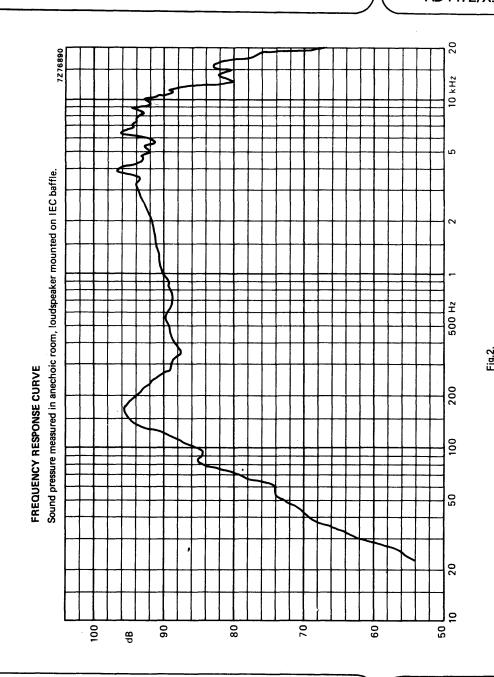


Fig. 1a.





One tag is indicated by a red mark for in-phase connection.





Mullard

September 1977

### 4 INCH MEDIUM POWER LOUDSPEAKERS

AD4080/X AD4480/X Series

#### APPLICATION

For portable receivers, small tape recorders and intercoms.

#### **TECHNICAL DATA**

		ve	rsion		
	X4	X8	X 15	X25	
Rated impedance	4	8	15	25	Ω
Voice coil resistance	3, 4	7, 1	13, 8	22, 6	Ω
Resonance frequency	165	165	165	165	Hz
Power handling capacity, measured without filter loudspeaker unmounted	3	3	3	3	w
Sweep voltage	2, 45	3, 5	4, 75	6, 1	v
Energy in airgap	55	55	55	55	mJ
Flux density	1	1	1	1	T
Airgap height	3	3	3	3	mm
Voice coil height	2, 4	3, 1	2, 55	2, 8	mm
Core diameter	18	18	18	18	mm
Magnet material	[	Magn	adur	•	
diameter weight	53 0, 1	53 0, 1	53 0, 1	53 0, 1	mm kg
Weight of loudspeaker	0, 25	0, 25	0, 25	0, 25	kg

The loudspeaker has a paper cone and surround.

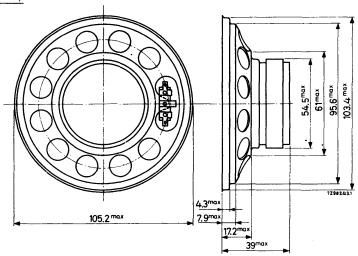
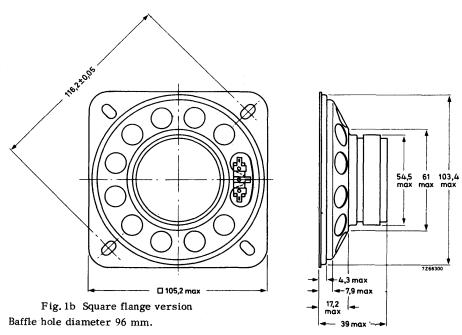


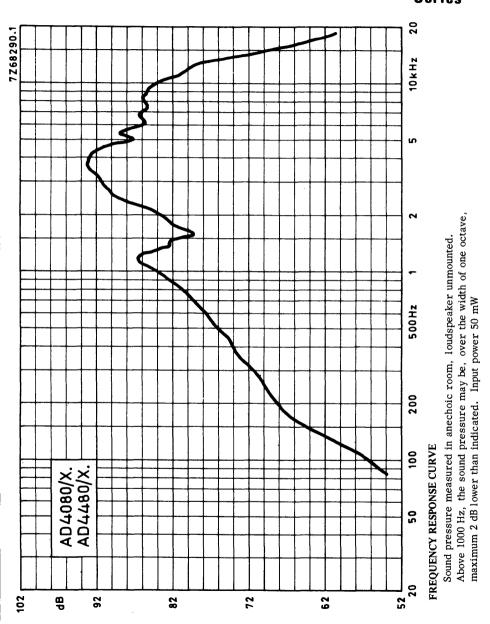
Fig. la Round flange version



One tag is indicated by a red mark for in-phase connection.

# 4 INCH MEDIUM POWER LOUDSPEAKERS

### AD4080/X AD4480/X Series



#### APPLICATION

For audio equipment in general. Frequency response up to 12 kHz, high sensitivity in bass region.

TECHNICAL DATA	version			
	X4	<b>X8</b>	X15	
Rated impedance	4	8	15	Ω
Voice coil resistance	3,4	7,1	13,5	Ω
Rated frequency range	80	to 14 00	00	Hz
Resonance frequency		150		Hz
Power handling capacity, measured without filter, loudspeaker unmounted		3		w
Operating power (sound level 90 dB, 1 m)		0,7		W
Sweep voltage (75 to 20 000 Hz)	2,5	3,5	4,7	V
Energy in air gap		38		mJ
Flux density		1,1		Т
Air-gap height		2,5		mm
Voice coil height	3,5	4,1	2,7	mm
Core diameter		14		mm
Magnet material diameter mass		Magnadu 46 0,053	ī	mm kg
Mass of loudspeaker		0,16		kg

The loudspeaker has a paper rim. Connections to the loudspeaker can be made by means of Fastons or by soldering.

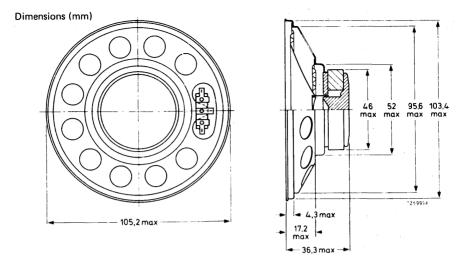


Fig. 1a Round flange type AD4085/X.

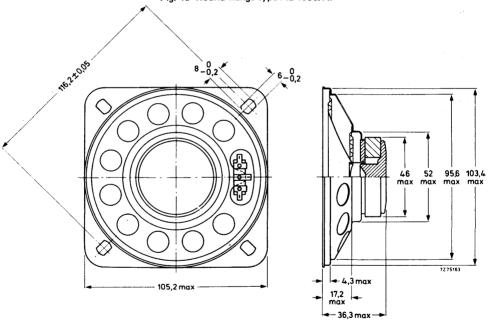


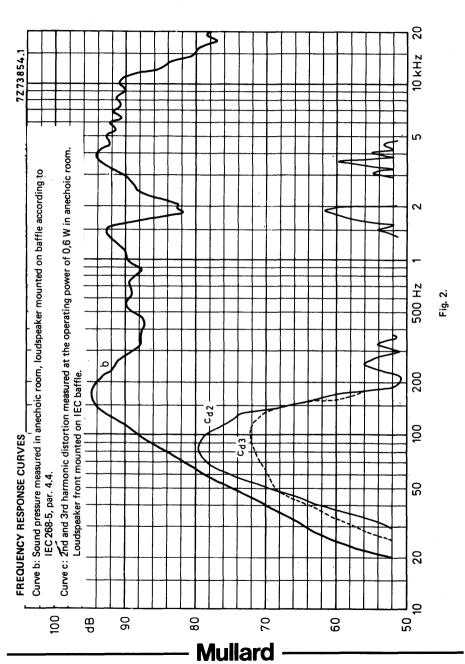
Fig. 1b Square flange type AD4485/X.

Baffle hole diameter 96 mm.

One tag is indicated by a red mark for in-phase connection.



# 4 INCH MEDIUM POWER LOUDSPEAKER



# 4 INCH LOW POWER LOUDSPEAKERS

AD4090/X. Series

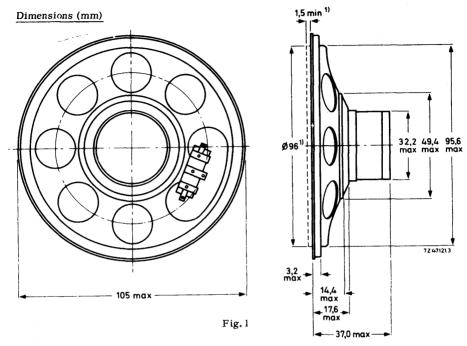
#### **APPLICATION**

For portable receivers.

#### **TECHNICAL DATA**

	ve	rsion	
	X8	X 15	
Rated impedance	8	15	Ω
Voice coil resistance	7, 2	13, 8	Ω
Resonance frequency	190	190	Hz
Power handling capacity, measured without filter loudspeaker unmounted	2	2	w
Sweep voltage	2,8	3, 9	v
Energy in airgap	39	39	ınJ
Flux density	0,8	0,8	Т
Airgap height	3	3	mm
Voice coil height	1,8	2, 55	mm
Core diameter	18	18	mm
Magnet material diameter weight	Ticonal 18 0,027	Ticonal 18 0,027	mm kg
Weight of loudspeaker	0, 125	0, 125	kg

The loudspeaker has a paper cone and surround.



 Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

One tag is indicated by a red mark for in-phase connection.

#### FREQUENCY RESPONSE CURVE

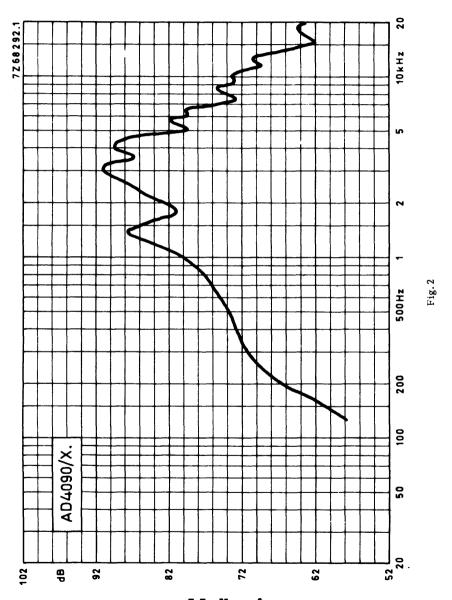
Fig. 2 Input power 50 mW

Sound pressure measured in anechoic room, loudspeaker unmounted.

Above 1000 Hz the sound pressure may be, over the width of one octave, maximum 2 dB lower than indicated.

### 4 INCH LOW POWER LOUDSPEAKERS

### AD4090/X. Series



**Mullard** 

AD4090/X Page 3

### 4 INCH MEDIUM POWER LOUDSPEAKER

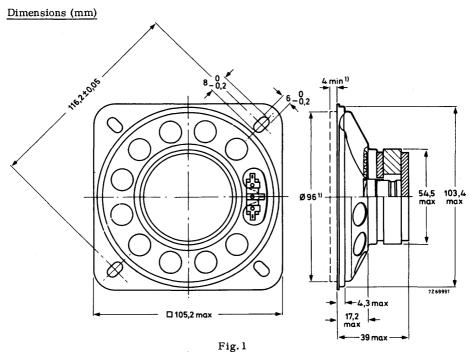
#### **APPLICATION**

With its excellent power handling capacity very suitable for car radios.

#### TECHNICAL DATA

Rated impedance	4	Ω
Voice coil resistance	3, 4	Ω
Rated frequency range	90 to 14000	Hz
Resonance frequency	140	Hz
Power handling capacity, measured without filter loudspeaker unmounted	8	W
Operating power (sound level 90 dB, 1 m)	0,8	W
Sweep voltage (80 to 20000 Hz)	3,5	V
Energy in air gap	50	mJ
Flux density	0, 95	T
Air gap height	3	mm
Voice coil height	4, 4	mm
Core diameter	18	mm
Magnet material diameter mass	Magnadur 54 0, 1	mm kg
Mass of loudspeaker	0,25	kg

The loudspeaker has a paper cone and a textile surround.



1) Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

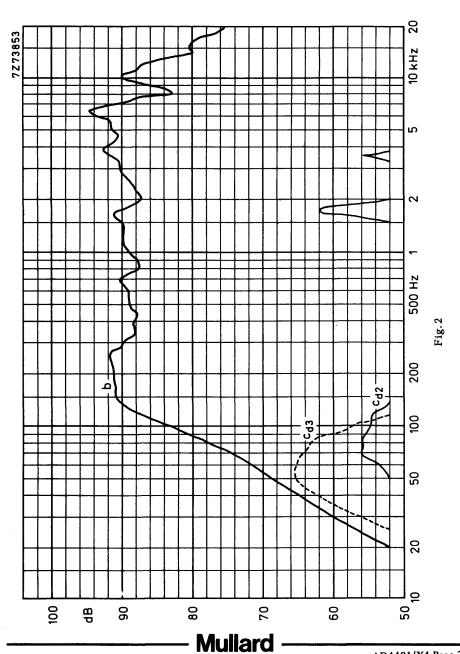
One tag is indicated by a red mark for in-phase connection.

#### FREQUENCY RESPONSE CURVES (see Fig. 2)

Curve b: Sound pressure measured in anechoic room, loudspeaker mounted on baffle according to IEC268-5, par. 4.4.

Curve c: 2nd and 3rd harmonic distortion, measured at the operating power of 0,8 W. Loudspeaker front mounted on IEC baffle.

### 4 INCH MEDIUM POWER LOUDSPEAKER





### 4 x 6 INCH OVAL MEDIUM POWER LOUDSPEAKERS

### AD4681/M Series

#### APPLICATION

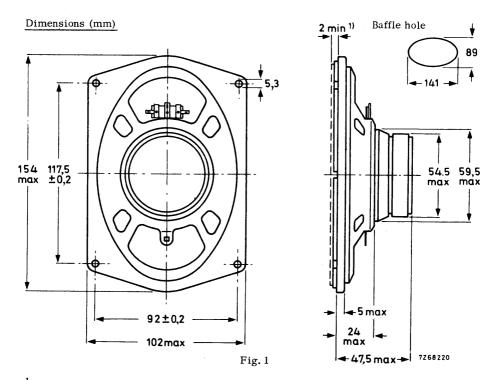
A full range loudspeaker for car and domestic radios, tape recorders and portable record players.

This speaker has an extended frequency response up to 20 kHz.

#### TECHNICAL DATA

		version		
	M4	M8	M25	
Rated impedance	4	8	25	Ω
Voice coil resistance	3,4	7,1	22,7	Ω
Resonance frequency	135	135	135	Hz
Power handling capacity, measured without filter loudspeaker unmounted	6	6	6	<b>W</b> ,
Sweep voltage	2,8	4	7,1	$\mathbf{v}$
Energy in airgap	55	55	55	mJ
Flux density	1	1	1	$\mathbf{T}^{\cdot}$
Airgap height	3	3	3	mm
Voice coil height	3	3,9	4	mm
Core diameter	18	18	18	mm
Magnet material		   Magnadur		
diameter weight	53 0, 1	53 0,1	53 0, 1	mm kg
Weight of loudspeaker	0,26	0,26	0,26	kg

The loudspeaker has a paper cone and surround and a foam plastic gasket on the flange.

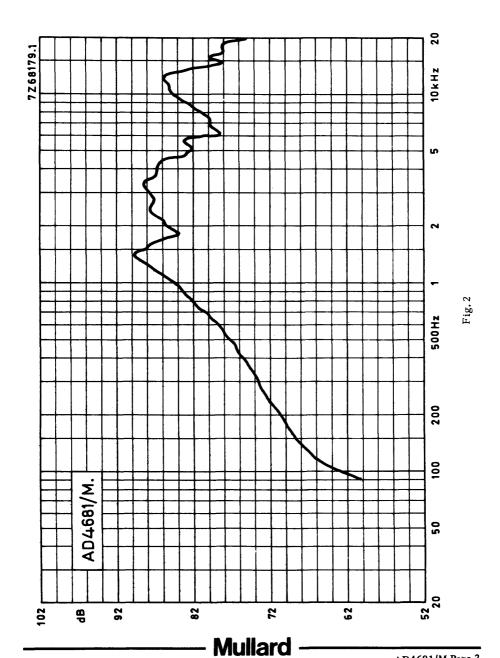


 Baffle hole and clearance depth required for cone movement at specified power handling capacity.

One tag is indicated by a red mark for in-phase connection.

#### FREQUENCY RESPONSE CURVE

Fig. 2 Sound pressure measured in anechoic room, loudspeaker unmounted. Above 1000 Hz the sensitivity may be, over the width of one octave, maximum 2 dB lower than indicated. Input power 50 mW.



#### 4 × 6 inch OVAL MEDIUM POWER LOUDSPEAKER

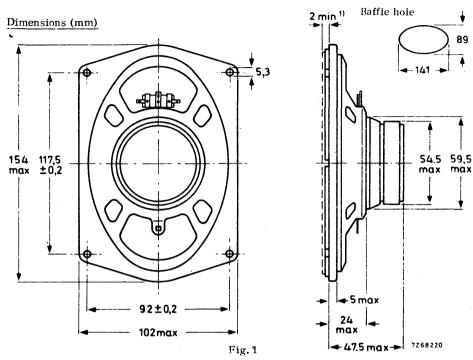
#### **APPLICATION**

For car and domestic radios, tape recorders and portables. Frequency range up to 12 kHz. High sensitivity at 3000 Hz.

TECHNICAL DATA		ver	sion		
	X4	X8	X15	X25	_
Rated impedance	4	8	15	25	Ω
Voice coil resistance	3, 1	7,1	13,5	22, 7	Ω
Resonance frequency	140	140	140	140	Hz
Power handling capacity, measured without filter loudspeaker unmounted	6	6	6	6	W
Sweep voltage	3, 5	4,9	6,7	8, 7	v
Energy in airgap	55	55	55	55	mJ
Flux density	1	1	1	1	T
Airgap height	3	3	3	3	mm
Voice coil height	3	. 3, 9	3, 2	4	mm
Core diameter	18	18	18	18	mm
Magnet material	!	Magi	nadur,		
diameter weight	53 0, 1	53 0, 1	53 0, 1	53 0, 1	mm kg
Weight of loudspeaker	0, 26	0,26	0, 26	0, 26	kg

The loudspeaker has a paper cone and surround and a foam plastic gasket on the flange.





1) Baffle hole and clearance depth required for cone movement at specified power handling capacity.

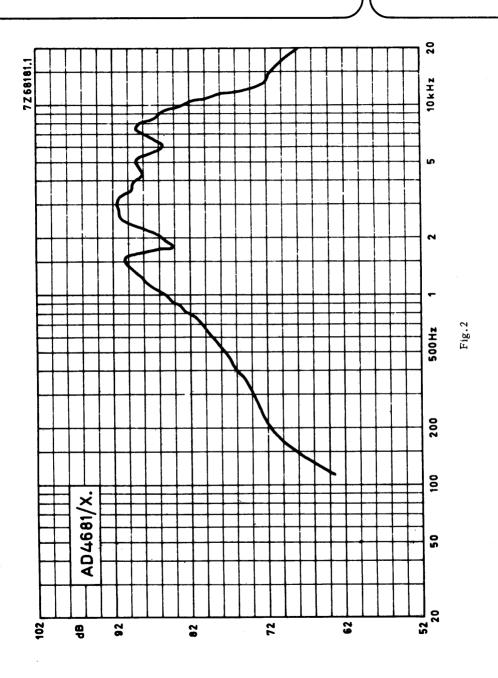
One tag is indicated by a red mark for in-phase connection.

#### FREQUENCY RESPONSE CURVE

Fig. 2. Input power 50 mW Sound pressure measured in anechoic room. loudspeaker unmounted. Above 1000 Hz the sensitivity may be, over the width of one octave, maximum 2 dB lower than indicated.

**Mullard** 

2





### 4 x 6 INCH OVAL MEDIUM POWER LOUDSPEAKERS

#### APPLICATION

A full range loudspeaker with an extended frequency response up to 20 kHz. Due to absence of stray ticonal sinterpot magnetic field, this loudspeaker can be used for black and white as well as colour television sets.

#### **TECHNICAL DATA**

			vers	sion	
	M4	M8	M15	M25	
Rated impedance	4	8	15	25	Ω
Voice coil resistance	3,4	7,1	13,5	22,7	Ω
Resonance frequency	135	135	135	135	Hz
Power handling capacity, measured without filter loudspeaker unmounted	4	4	4	4	W
Sweep voltage	2,8	4	5,5	7,1	v
Energy in airgap	39	39	39	39	mJ
Flux density	0,8	0,8	0,8	0,8	т
Airgap height	3	3	3	3	mm
Voice coil height	3	3,9	3,2	4	mm
Core diameter	18	18	18	18	mm
Magnet material diameter weight	Ticonal 18 0,027	Ticonal 18 0,027	Ticonal 18 0,027	Ticonal 18 0,027	mm kg
Weight of loudspeaker	0, 16	0, 16	0, 16	0, 16	kg

The loudspeaker has a paper cone and surround and a foam plastic gasket on the flange.

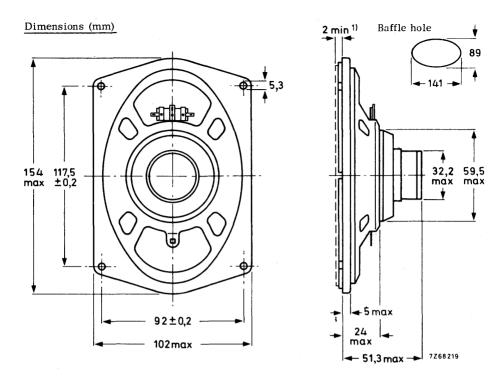


Fig. 1

 $^{1}$ ) Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

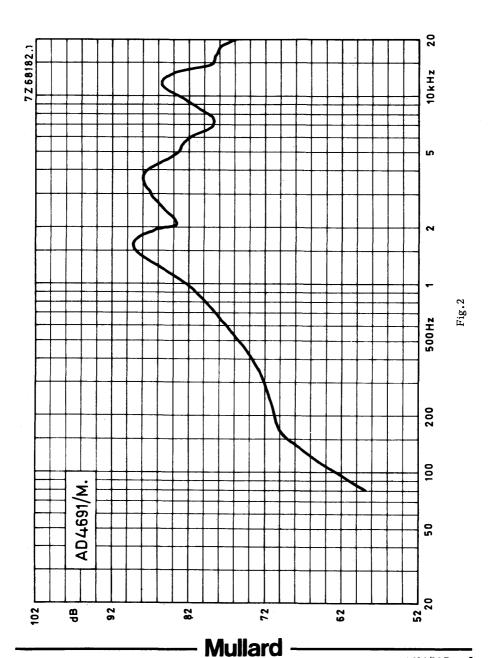
One tag is indicated by a red mark for in-phase connection.

#### FREOUENCY RESPONSE CURVE

Fig. 2 Sound pressure measured in anechoic room, loudspeaker unmounted. Above 1000 Hz the sensitivity may be, over the width of one octave, maximum 2 dB lower than indicated. Input power 50 mW



# 4 x 6 INCH OVAL MEDIUM POWER LOUDSPEAKERS



### 4 x 6 INCH OVAL MEDIUM POWER LOUDSPEAKER

### AD4691/X

#### APPLICATION

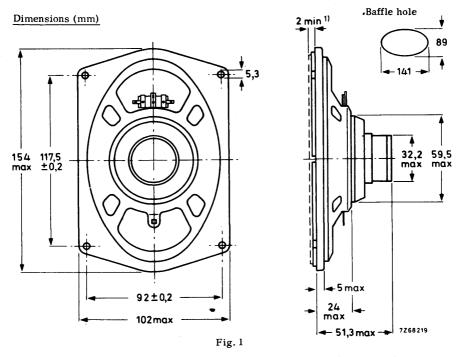
Due to absence of stray magnetic ticonal sinterpot field, the loudspeaker can be used in black and white as well as colour television sets.

High sensitivity at 3000 Hz. Frequency response up to 12 kHz.

**TECHNICAL DATA** 

		vers	ion		
	X4	X8	X15	X25	
Rated impedance	4	8	15	25	Ω
Voice coil resistance	3,4	7,1	13, 5	22, 7	Ω
Resonance frequency	140	140	140	140	Hz
Power handling capacity, measured without filter, loudspeaker unmounted	4	4	4	4	W
Sweep voltage	2,8	4	5, 5	7, 1	v
Energy in airgap	39	39	39	39	mJ
Flux density	0,8	0,8	0,8	0,8	T
Airgap height	3	3	3	3	mm
Voice coil height	3	3, 9	3, 2	4	mm
Core diameter	18	18	18	18	mm
Magnet material diameter weight	Ticonal 18 0,027	Ticonal 18 0,027	Ticonal 18 0,027	Ticonal 18 0,027	mm kg
Weight of loudspeaker	0, 16	0, 16	0, 16	0, 16	kg

The loudspeaker has a paper cone and surround and a foam plastic gasket on the flange.



 Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

One tag is indicated by a red mark for in-phase connection.

#### FREQUENCY RESPONSE CURVE

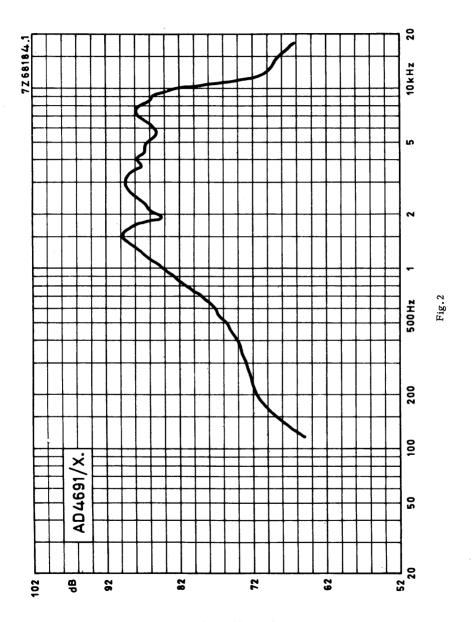
Fig. 2. Input power 50 mW

Sound pressure measured in anechoic room, loudspeaker unmounted.

Above 1000 Hz sensitivity may be, over the width of one octave, maximum 2 dB lower than indicated.

### 4 x 6 INCH OVAL MEDIUM POWER LOUDSPEAKER

### **AD4691/X**



# 4 x 8 INCH OVAL MEDIUM POWER LOUDSPEAKER

### AD4890/X

#### APPLICATION

For colour television sets. Low stray field, low resonance frequency, high sensitivity in bass region.

#### TECHNICAL DATA

		ver	sion		
	_X4	X8	X15	X25	
Rated impedance	4	8	15	25	
Voice coil resistance	3.4	7.1	13,5	22,7	
Rated frequency range	55 to 13 000				
Resonance frequency	110				
Power handling capacity, measured without filter, loudspeaker unmounted			10		
Operating power (sound level 90 dB, 1 m)	0,7				
Sweep voltage (55 to 20 000 Hz)	4	5,7	7,8	10,	
Energy in air gap			39		
Flux density		(	0,8		
Air gap height			3		
Voice coil height	4,5	3, 9	3, 2	4	
Core diameter			18		
Magnet material diameter mass			onal 18 027		
Mass of loudspeaker	0, 23				

The loudspeaker has a paper cone and surround and a foam plastic gasket on the flange.

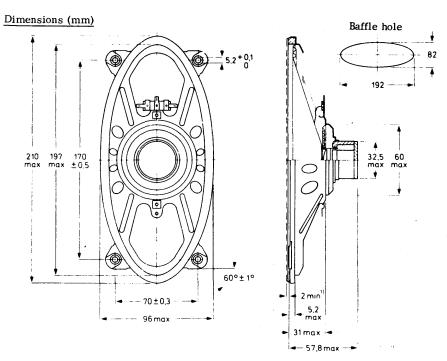


Fig. 1

### FREQUENCY RESPONSE CURVES (see Fig. 2)

Curve b: Sound pressure measured in anechoic room, loudspeaker mounted on baffle according to 1EC268-5. par. 4.4.

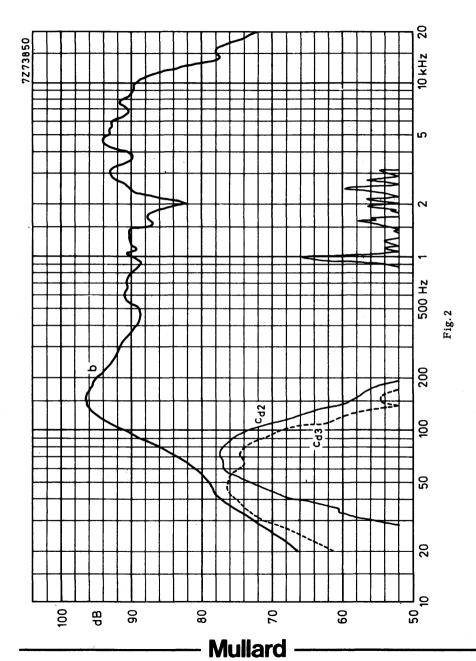
Curve c: 2nd and 3rd harmonic distortion, measured at the operating power of 0,7 W in anechoic room. Loudspeaker front mounted on IEC baffle.

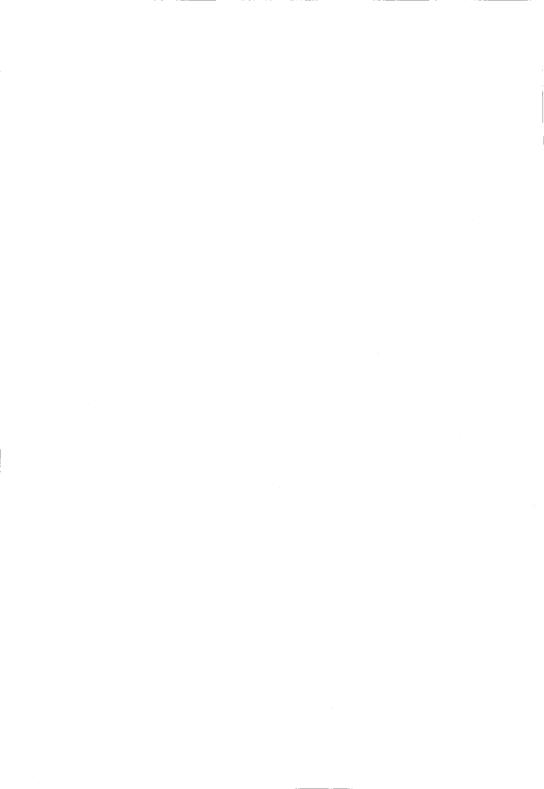


<sup>1)</sup> Clearance depth required for cone movement at the specified power handling capacity.

One tag is indicated by a red mark for in-phase connection.

4 x 8 INCH OVAL MEDIUM POWER LOUDSPEAKER





## 5 inch HIGH POWER SQUAWKER LOUDSPEAKER

#### APPLICATION

For the reproduction of audio frequencies from 500 to 4500 Hz with very low distortion in multi-way high-fidelity loudspeaker systems in accordance with DIN45500. The loudspeaker has an excellent spherical radiation pattern. Rated frequency range 500 to 5000 Hz.

#### TECHNICAL DATA

	version		
	Sq4	Sq8	_
Rated impedance	4	8	Ω
Voice coil resistance	3,4	6, 4	Ω
Resonance frequency	210	210	Hz
Power handling capacity measured with filter: 72 $\mu F$ - 2,1 mH (4Ω) $$36~\mu F$ - 4.5 mH (8Ω) loudspeaker unmounted	40 -	- 40	w w
Operating power	4	4	W
Sweep voltage frequency range: 400 - 5000 Hz filter high pass : 72 $\mu F$ - 2, 1 mH (4 $\Omega$ ) 36 $\mu F$ - 4, 5 mH (8 $\Omega$ )	3,5	_ 5	V V
Energy in air gap	140	140	mJ
Flux density	0, 93	0, 93	T
Air-gap height	5	5	mm
Voice coil height	6, 8	6, 8	mm
Core diameter	25	25	mm
Magnet material diameter	<b>Mag</b> n 72	adur 72	mm
mass	0, 23	0, 23	kg
Mass of loudspeaker	0.8	0.8	kø

The loudspeaker has a paper cone, a rubber surround and a sealed pot; no acoustic isolation required.

Connection to the loudspeaker by means of 6.3 mm (0.25 inch) tag connectors or by soldering.



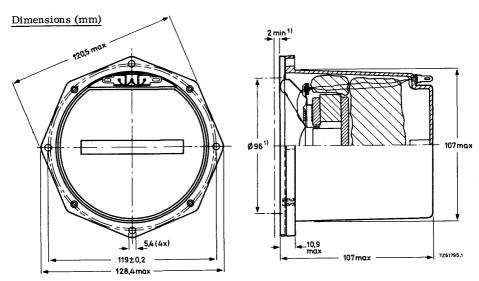


Fig. 1

 Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

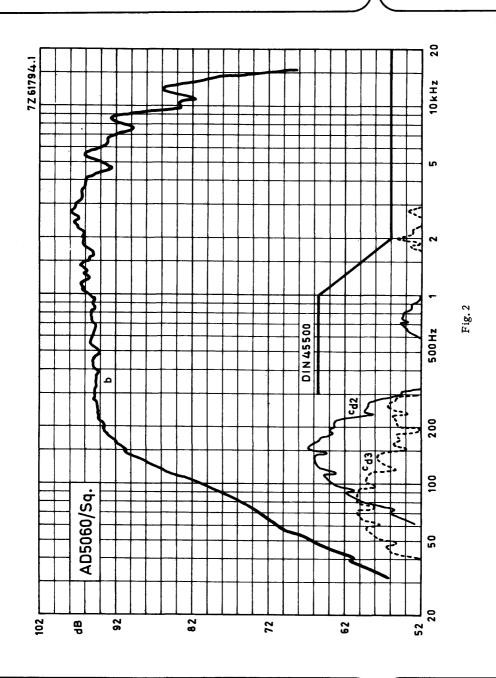
One tag is indicated by a red mark for in-phase connection.

#### FREQUENCY RESPONSE CURVES (see Fig. 2)

Curve b: Sound pressure measured in half free field at operating power of  $4\ W$  in anechoic room, loudspeaker mounted on IEC baffle.

Curve c: 2nd and 3rd harmonic distortion, measured at the operating power of 4 W in anechoic room.

2





## 5 inch HIGH POWER FULL RANGE LOUDSPEAKER

#### APPLICATION

A full range loudspeaker for small sealed enclosures of maximum 7 litres and also suitable for use in bookshelves enclosures.

Extended frequency response 75 - 20 kHz in 7 litres enclosures.

#### TECHNICAL DATA

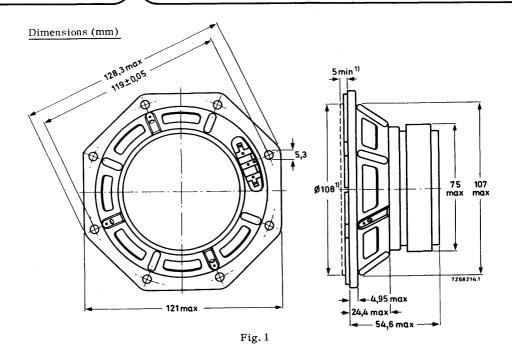
	ver	version		
	M4	M8		
Rated impedance	4	8	Ω	
Voice coil resistance	3,4	7	Ω	
Resonance frequency	85	85	Hz	
Power handling capacity measured without filter, loudspeaker unmounted	10	10	w	
Operating power	2	2	w	
Sweep voltage	3, 2	4.5	v	
Energy in airgap	127	127	mJ	
Flux density	0, 87	0, 87	Т	
Airgap height	5	5	mm	
Voice coil height	6,5	6,5	mm	
Core diameter	25	25	mm	
Magnet material diameter weight	<b>Mag</b> r 72 0, 26	i nadur   72   0, 26	mm kg	
Weight of loudspeaker	0, 665	0, 665	kg	

The loudspeaker has a paper cone, a textile surround and a foam plastic gasket on the flange.

Connection to the loudspeaker by means of 2.8 mm (0.11 inch) tag connectors or by soldering.



## AD5061/M.



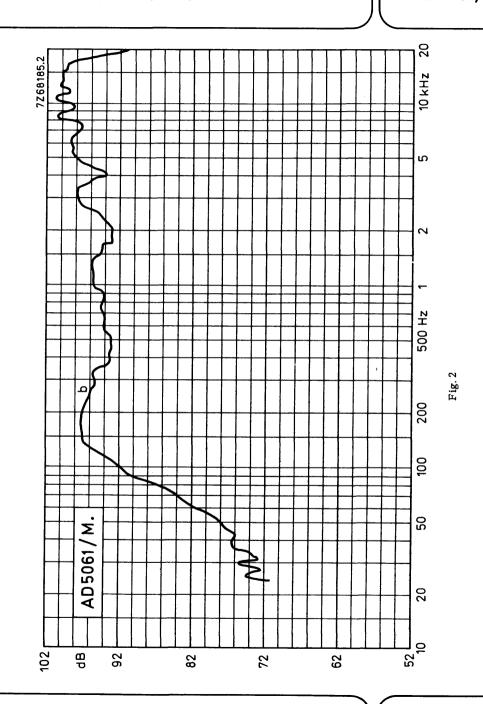
1) Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

One tag is indicated by a red mark for in-phase connection.

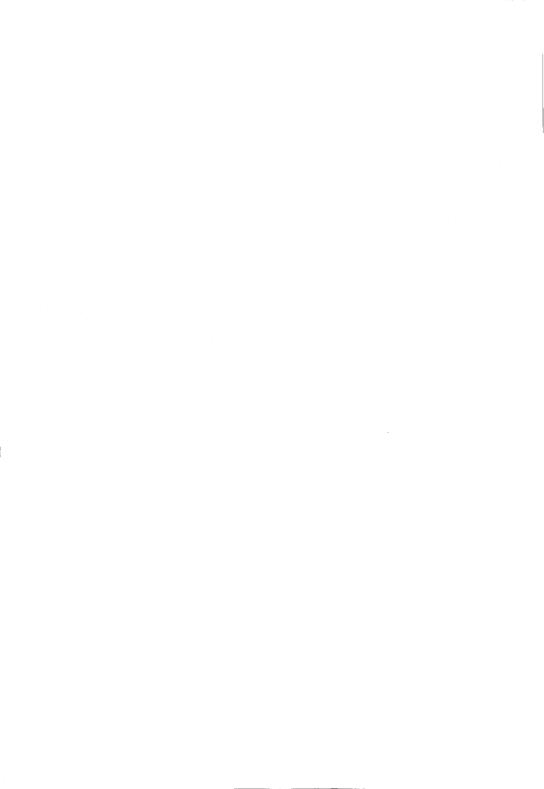
#### FREQUENCY RESPONSE CURVE

Curve b: Sound pressure measured in anechoic room at input power of 2 W. Loudspeaker mounted on IEC baffle.

AD5061/M.







## 5 INCH HIGH POWER SQUAWKER LOUDSPEAKER

## AD5061/Sq

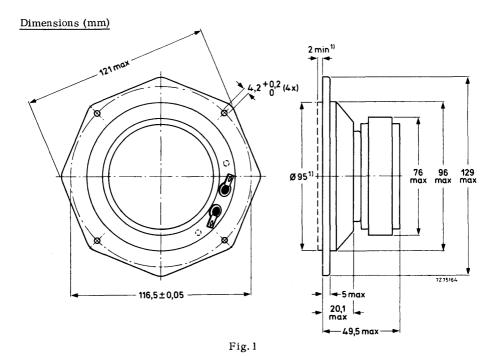
#### APPLICATION

For the reproduction of audio frequencies from 1300 to 5000 Hz with very low distortion in multi-way high-fidelity loudspeaker systems in accordance with DIN45500. The loudspeaker has an excellent spherical radiation pattern.

TECHNICAL DATA	vers	ion	
	Sq4	<b>S</b> q8	
Rated impedance	4	8	Ω
Voice coil resistance	3, 4	7	Ω
Resonance frequency	680		Hz
Rated frequency range	1300 to 5000		Hz
Power handling capacity, measured with filter: 24 $\mu F=0,4$ mH $$12\mu F=0,8$ mH loudspeaker unmounted	10	10	w w
Operating power		2	W
Sweep voltage frequency range: $300-5000~Hz$ high pass filter: $24~\mu F - 0.4~mH$ $12~\mu F - 0.8~mH$	3, 5	5	V V
Energy in air gap	14	0	mJ
Flux density	0,9	3	T
Air-gap height		5	mm
Voice coil height	6,	8	mm
Core diameter	2	5	mm
Magnet material diameter mass	<b>Magna</b> 7 0, 2	2	mm kg
Mass of loudspeaker	0,	8	kg

The loudspeaker has a sealed frame and a textile rim.

Connection to the loudspeaker is by means of 2,8 mm (0,11 inch) Fastons or soldering.



1) Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

One tag is indicated by a red mark for in-phase connection.

#### FREQUENCY RESPONSE CURVES (see Fig. 2)

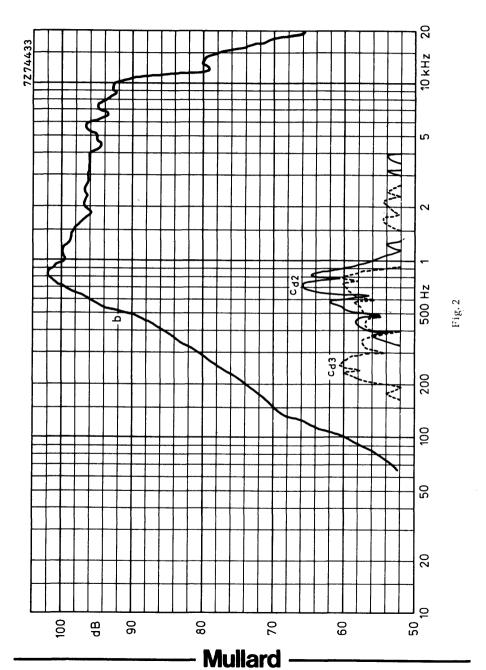
Curve b: Sound pressure measured in half free field at operating power of 2 W in anechoic room, loudspeaker mounted on IEC baffle.

Curve c: 2nd and 3rd harmonic distortion, measured at the operating power of 2 W in anechoic room.



## 5 INCH HIGH POWER SQUAWKER LOUDSPEAKER

## AD5061/Sq



			,

### 5 × 7 inch OVAL MEDIUM POWER LOUDSPEAKER

#### APPLICATION

A full range loudspeaker for car and domestic radios, tape recorders and portable record players.

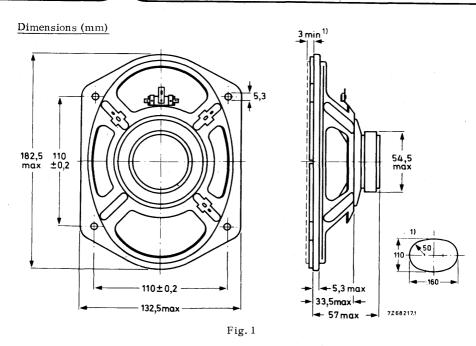
Due to its dual-cone construction, this loudspeaker has an extended frequency response up to  $20\ \text{kHz}$ .

TECHNICAL DATA		vers	sion		
	M4	M8	M15	M25	
Rated impedance	4	8	15	25	Ω
Voice coil resistance	3,4	7, 1	13,5	22,7	Ω
Resonance frequency	100	100	100	100	Hz
Power handling capacity, measured without filter loudspeaker unmounted	6	6	6	6	W
Sweep voltage	2,8	4	5,5	8,7	<b>V</b> .
Energy in air gap	53	53	53	53	mJ
Flux density	0, 98	0, 98	0, 98	0, 98	T
Air-gap height	3	3	3	3	mm
Voice coil height	3	3, 9	3, 2	4	mm
Core diameter	18	18	18	18	mm
Magnet material		Magr	adur		
diameter mass	53 0, 1	53 0,1	53 0, 1	53 0, 1	mm kg
Mass of loudspeaker	0, 32	0, 32	0, 32	0, 32	kg

The loudspeaker has a paper cone and a foam plastic gasket on the flange.

Connection to the loudspeaker by means of 0.28 mm (0.11 inch) tag connectors or by soldering.



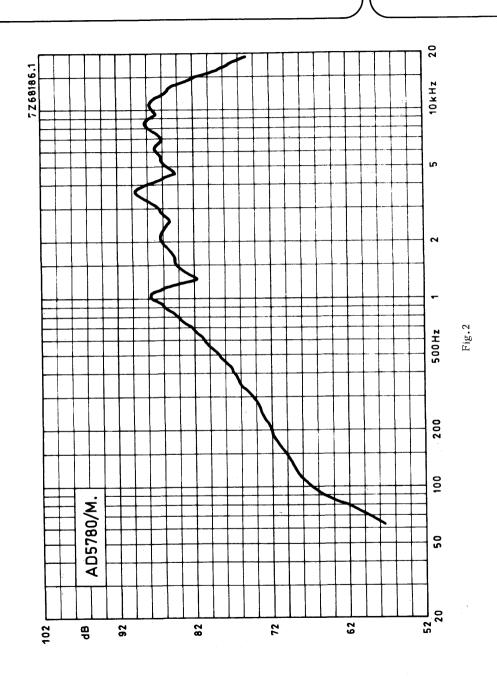


 Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

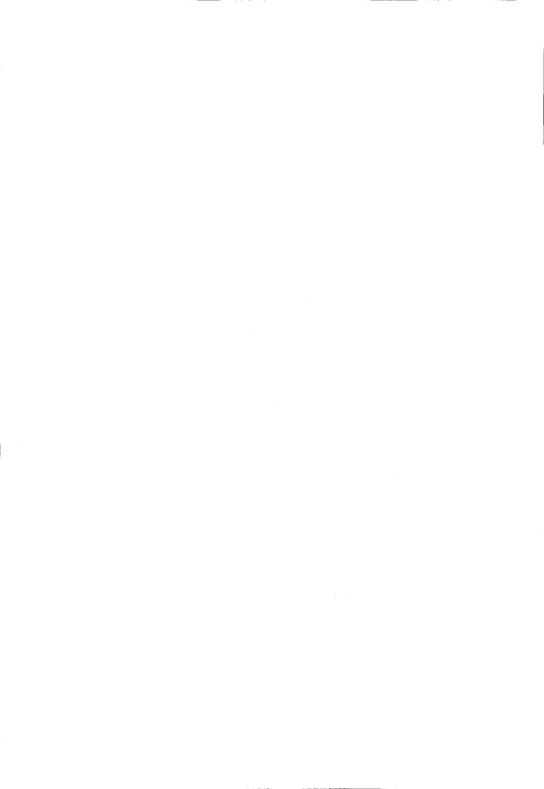
One tag is indicated by a red mark for in-phase connection.

#### FREQUENCY RESPONSE CURVE

Fig. 2. Input power 50 mW
Sound pressure measured in anechoic room, loudspeaker unmounted.
Above 1000 Hz the sensitivity may be, over the width of one octave, maximum 2 dB lower than indicated.







### 5 × 7 inch OVAL MEDIUM POWER LOUDSPEAKER

#### APPLICATION

For car and domestic radios, tape recorders and portable record players. High sensitivity at 4000 Hz. Frequency range up to  $10\ \text{kHz}$ .

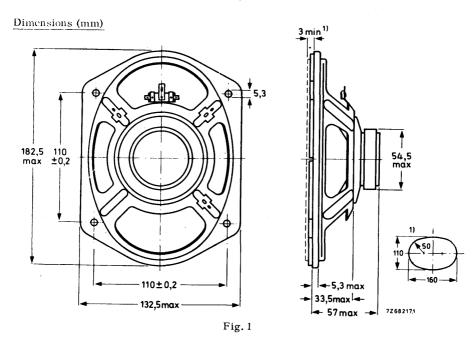
TECHNICAL DATA		vers	ion		
	X4	X8	X15	X25	
Rated impedance	4	8	15	25	Ω
Voice coil resistance	3,4	7, 1	13,5	22,7	Ω
Resonance frequency	115	115	115	115	Hz
Power handling capacity, measured without filter, loudspeaker unmounted	6	6	6	6	W
Sweep voltage	3, 4	3, 5	4,8	6, 1	V
Energy in airgap	55	55	55	55	mJ
Flux density	0,98	0, 98	0, 98	0, 98	T
Airgap height	3.	3	3	3	mm
Voice coil height	3	3, 9	3, 2	4	mm
Core diameter	18	18	18	18	mm
Magnet material		Magn	adur	•	
diameter	53	53	53	53	mm
weight	0, 1	0, 1	0, 1	0, 1	kg
Weight of loudspeaker	0, 32	0, 32	0,32	0, 32	kg

The loudspeaker has a paper cone and surround and a foam plastic gasket on the flange.

Connection to the loudspeaker by means of 2.8 mm (0.11 inch) tag connectors or by soldering.



## AD5780/X.



1) Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

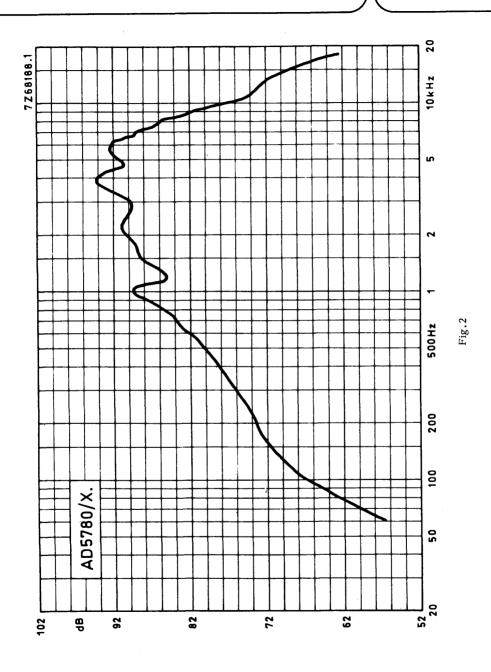
One tag is indicated by a red mark for in-phase connection.

#### FREQUENCY RESPONSE CURVE

Fig. 2. Input power 50 mW

Sound pressure measured in anechoic room, loudspeaker unmounted.

Above 1000 Hz the sensitivity may be, over the width of one octave, maximum 2 dB lower than indicated.





# 5 x 7 INCH OVAL MEDIUM POWER LOUDSPEAKERS

#### APPLICATION

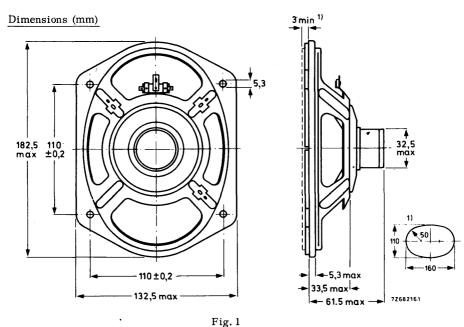
Due to absence of stray magnetic ticonal sinterpot field, the loudspeaker can be used in black and white as well as colour television sets. High sensitivity at 3000 Hz.

#### TECHNICAL DATA

		version		
	M4	M8	M15	
Rated impedance	4	8	15	Ω
Voice coil resistance	3, 4	7, 1	13,5	Ω
Resonance frequency	100	100	100	Hz
Power handling capacity, measured without filter, loudspeaker unmounted	4	. 4	4	W
Sweep voltage	2,8	4	5,5	v
Energy in airgap	39	39	39	mJ
Flux density	0,8	0,8	0, 8	T
Airgap height	3	3	3	mm
Voice coil height	3	3,9	3, 2	mm
Core diameter	18	18	18	mm
Magnet material diameter weight	Ticonal 18 0,027	Ticonal 18 0,027	Ticonal 18 0,027	mm kg
Weight of loudspeaker	0, 22	0, 22	0, 22	kg

The loudspeaker has a paper cone and surround and a foam plastic gasket on the flange.

Connection to the loudspeaker by means of 2.8 mm (0.11 inch) tag connectors or by soldering.



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Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

One tag is indicated by a red mark for in-phase connection.

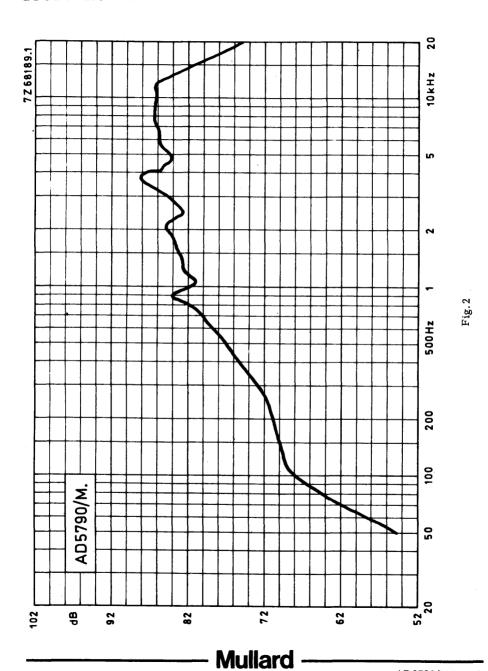
#### FREQUENCY RESPONSE CURVE

Fig. 2 Sound pressure measured in anechoic room, loudspeaker unmounted.

Above 1000 Hz the sensitivity may be, over the width of one octave, maximum 2 dB lower than indicated.

Input power 50 mW





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## 5 × 7 inch OVAL MEDIUM POWER LOUDSPEAKER

#### APPLICATION

Due to absence of stray magnetic Ticonal sinterpot field, the loudspeaker can be used in black and white as well as colour television sets. High sensitivity at  $3000~\rm{Hz}$ .

#### TECHNICAL DATA

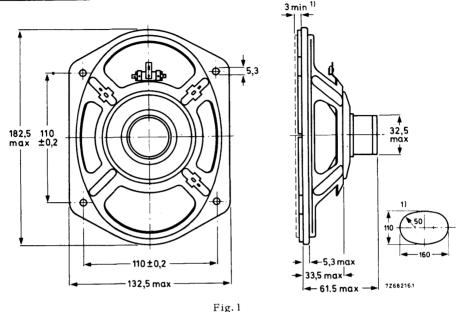
	version		
	X4	X8	
Rated impedance	4	8	Ω
Voice coil resistance	3, 4	7, 1	Ω
Resonance frequency	11	5	Hz
Power handling capacity, measured without filter, loudspeaker unmounted		4	w
Operating power	0,	7	w
Sweep voltage	2, 45	4	v
Energy in air gap	3	9	mJ
Flux density	0,	8	T
Air-gap height	:	3	mm
Voice coil height	3	3, 9	mm
Core diameter	18		mm
Magnet material diameter mass	18	Ticonal 18 0,027	
Mass of loudspeaker	0, 22	2	kg

The loudspeaker has a paper cone, a treated paper surround and a foam plastic gasket on the flange.

Connection to the loudspeaker by means of 2.8 mm (0.11 inch) tag connectors or by soldering.



#### Dimensions (mm)



1) Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

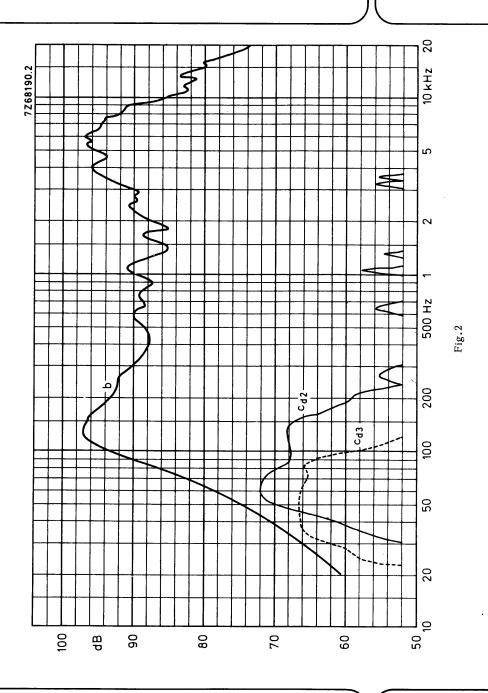
One tag is indicated by a red mark for in-phase connection.

### FREQUENCY RESPONSE CURVES (see Fig. 2)

Curve b: Sound pressure measured in anechoic room at input power of 2, 2 W. Loudspeaker mounted on IEC baffle.

**Mullard** 

Curve c: 2nd and 3rd harmonic distortion, measured at input power of 2, 2 W in anechoic room. Loudspeaker mounted on IEC baffle.







# 7 inch HIGH POWER WOOFER LOUDSPEAKER

## AD7060/W

#### **APPLICATION**

For high fidelity reproduction in sealed acoustic enclosures.

Maximum enclosure volume 7 litres; maximum recommended cross-over frequency 3000 Hz.

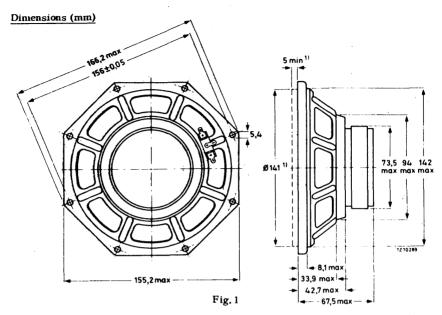
Rated frequency range 40 to 3000 Hz.

#### **TECHNICAL DATA**

	version			
	W 4	W8		
Rated impedance	4	8	Ω	
Voice coil resistance	3,4	6, 8	Ω	
Resonance frequency	45	45	Hz	
Power handling capacity, measured without filter, mounted in 7 & sealed enclosure	30	30	W	
Operating power	6, 3	6, 3	W	
Sweep voltage frequency range 35 - 5000 Hz	3,8	5, 3	v	
Energy in airgap	135	140	mJ	
Flux density	0, 87	0, 93	Т	
Airgap height	5	5	mm	
Voice coil height	9,7	9,7	mm	
Core diameter	25	25	mm	
Magnet material	Mag	! Magnadur		
diameter weight	72 0, 26	72 0, 26	mm kg	
Weight of loudspeaker	0.68	0.68	kø	

The loudspeaker has a rubber surround.

Connection to the loudspeaker by means of 6, 3 mm (0, 25 inch) Fastons or soldering.



1) Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

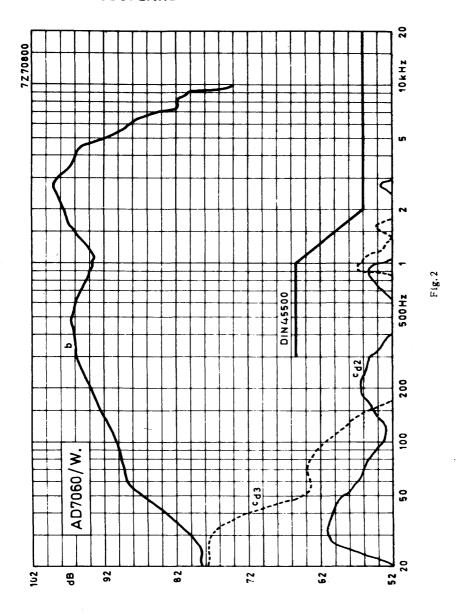
One tag is indicated by a red mark for in-phase connection.

#### FREQUENCY RESPONSE CURVES (see Fig. 2)

Curve b: Sound pressure measured in anechoic room, input at an operating power of 6, 3 W.

Loudspeaker mounted in sealed 80 l enclosure, filled with 1 kg of glass wool.

Curve c: 2 nd and 3 rd harmonic distortion, measured at the operating power of 6, 3 W in anechoic room, loudspeaker mounted in sealed 80 l enclosure with 1 kg of glass wool.





# 7 INCH HIGH POWER FULL RANGE LOUDSPEAKER

## AD7062/M

#### APPLICATION

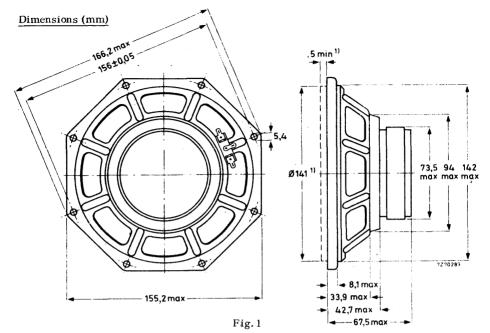
For high fidelity reproduction in sealed acoustic enclosures. Maximum enclosure volume 7 litres. High power handling capacity with very low distortion.

#### TECHNICAL DATA

	version		
	M4	M8	
Rated impedance	4	8	Ω
Voice coil resistance	4, 3	8	Ω
Resonance frequency	45	45	Hz
Power handling capacity, measured without filter, mounted in 71 sealed enclosure	30	30	W
Operating power	5	5	W
Sweep voltage	3, 8	5,3	V
Energy in air gap	135	140	mJ
Flux density	0,87	0,93	T
Air-gap height	5	5	mm
Voice coil height	11	11	mm
Core diameter	25	25	mm
Magnet material diameter mass	<b>Ma</b> g 72 0, 26	nadur   72   0, 26	mm kg
Mass of loudspeaker	0,68	0,68	kg

The loudspeaker has a rubber surround and a double cone.

Connection to the loudspeaker by means of 6,3 mm (0,25 inch) Fastons or soldering.



1) Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

One tag is indicated by a red mark for in-phase connection.

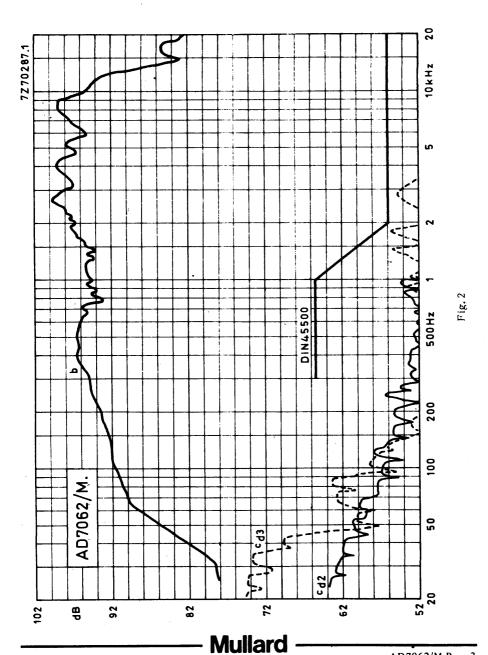
#### FREQUENCY RESPONSE CURVE

See Fig. 2

Curve b: Sound pressure measured in anechoic room at operating power. Loudspeaker mounted in sealed 801 enclosure, filled with 1 kg of glass wool.

Curve c: 2<sup>nd</sup> and 3<sup>rd</sup> harmonic distortion, measured at the operating power of 5 W in anechoic room, loudspeaker mounted in sealed 80 l enclosure, filled with 1 kg of glass wool.

## 7 INCH HIGH POWER FULL RANGE LOUDSPEAKER



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#### **APPLICATION**

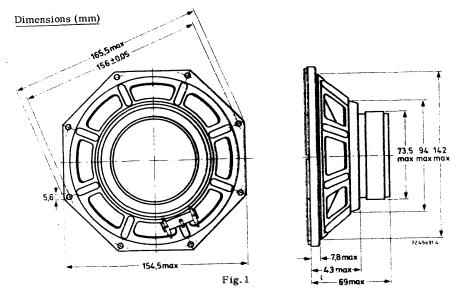
For high fidelity reproduction in sealed acoustic enclosures. Maximum enclosure volume 25 litres.

#### **TECHNICAL DATA**

	version		
	M4	M8	
Rated impedance	4	8	Ω
Voice coil resistance	3, 4	7	Ω
Resonance frequency	55	55	Hz
Power handling capacity, measured without filter, loudspeaker unmounted	10	10	w
Operating power	2, 2	2, 2	W
Sweep voltage	4,5	6, 3	$\dot{\mathbf{v}}$
Energy in airgap	127	127	mJ
Flux density	0, 87	0, 87	Т
Airgap height	5	5	mm
Voice coil height	6,8	6,8	mm
Core diameter	25	25	mm
Magnet material	l Magnadur		
diameter weight	72 0, 26	72 0, 26	mm kg
Weight of loudspeaker	0,745	0, 745	kg

The loudspeaker has a textile surround and a double cone.

Connection to the loudspeaker by means of 6,3 mm (0,25 inch) Fastons or soldering.



Baffle hole diameter 141 mm
One tag is indicated by a red mark for in-phase connection.

#### FREQUENCY RESPONSE CURVES

Curve b: Sound pressure measured in anechoic room at input power of 2, 2 W. Loudspeaker mounted on IEC baffle.

Curve c: 2nd and 3rd harmonic distortion, measured at input power of 2.2 W in anechoic room. Loudspeaker mounted on IEC baffle.

## 7inch HIGH POWER FULL RANGE LOUDSPEAKER

## AD7063/M

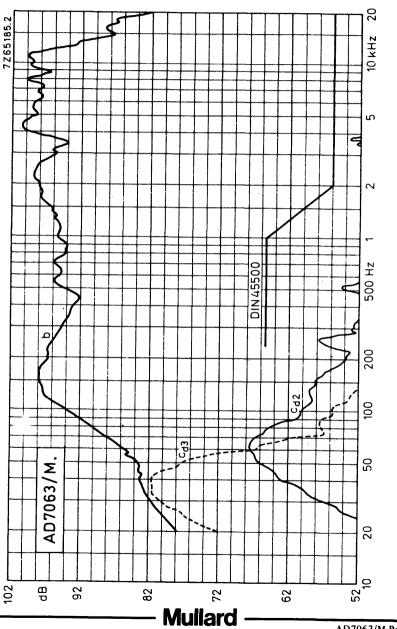


Fig. 2

## 7 inch HIGH POWER WOOFER LOUDSPEAKER

#### APPLICATION

For high fidelity reproduction in sealed acoustic enclosures in accordance with DIN45500. Maximum enclosure volume 7 1.

Maximum recommended cross-over frequency 2000 Hz. High power handling capacity with very low distortion.

#### TECHNICAL DATA

	version		
	W4	W8	
Rated impedance	4	8	Ω
Voice coil resistance	4, 3	8	Ω
Resonance frequency	45	45	Hz
Power handling capacity, measured without filter mounted in 7 1 sealed enclosure	40	40	W
Operating power	4	4	W
Sweep voltage	3, 8	5,3	V
Energy in air gap	. 225	207	mJ
Flux density	1, 1	1, 2	T
Air-gap height	5	5	mm
Voice coil height	11	11	mm
Core diameter	25	25	mm
Magnet material diameter mass	<b>Mag</b> i 90 0, 45	nadur   90   0,45	mm kg
Mass of loudspeaker	1.15	1. 15	kø

The loudspeaker has a paper cone and rubber surround.

Connection to the loudspeaker by means of 6.3 mm (0.25 inch) tag connectors or by soldering.



## AD7066/W.

#### Dimensions (mm)

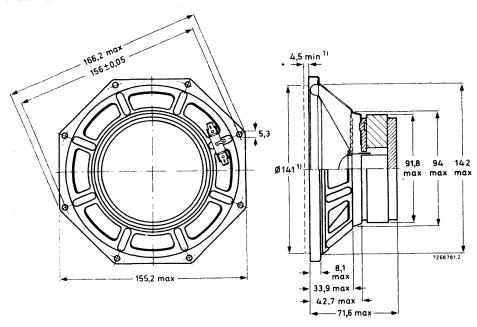


Fig.1

1) Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

One tag is indicated by a red mark for in-phase connection.

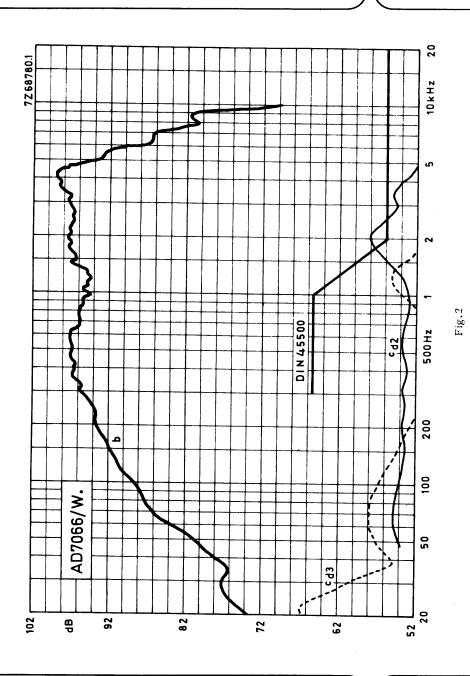
#### FREQUENCY RESPONSE CURVES

#### Fig. 2

Curve b: Sound pressure measured in anechoic room, loudspeaker mounted in sealed 80 l enclosure, filled with 1 kg of glass wool.

**Mullard** 

Curve c: Total non-linear distortion, measured at the operating power of 4 W in anechoic room, loudspeaker mounted in sealed 801 enclosure, filled with 1 kg of glass wool. Loudspeaker front mounted on baffle, dimensions 640 x 540 mm.





### 7 inch OCTAGONAL MEDIUM POWER LOUDSPEAKER

#### APPLICATION

For car and domestic radios, acoustic enclosures and public address systems. Frequency range up to  $15\ kHz$ .

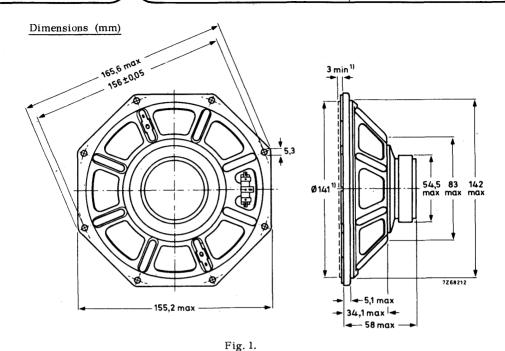
#### TECHNICAL DATA

	version			
	M4	M8	M15	_
Rated impedance	4	8	15	Ω
Voice coil resistance	3,4	7, 1	13,5	Ω
Resonance frequency	105	105	105	Hz
Power handling capacity, measured without filter loudspeaker unmounted	6	6	6	W
Sweep voltage	2,8	4	6, 7	$\mathbf{v}$
Energy in airgap	55	55	53	mJ
Flux density	0, 98	0, 98	0, 98	T
Airgap height	3	3	3	mm
Voice coil height	3	3, 9	3, 2	mm
Core diameter	18	18	18	mm
Magnet material		Magnadur		
diameter	- 53	53	53	mm
weight	0,1	0, 1	0, 1	kg
Weight of loudspeaker	0, 29	0, 29	0, 29	kg

The loudspeaker has a dual paper cone and a paper surround and has a foam plastic gasket on the flange.

Connection to the loudspeaker by means of 2.8 mm (0.11 inch) tag connectors or by soldering.





1) Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

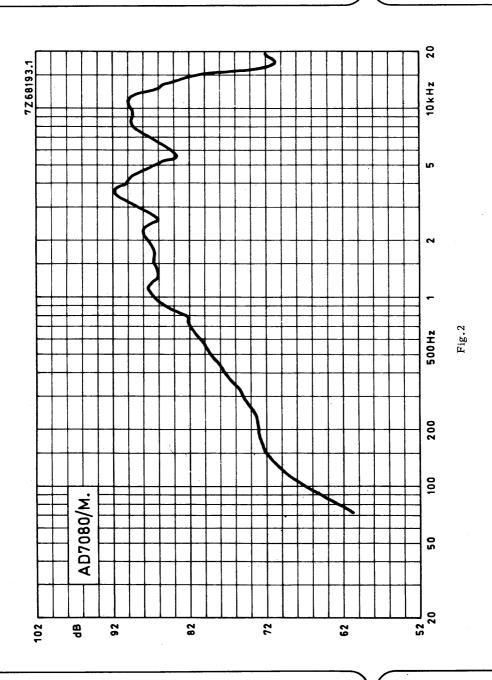
One tag is indicated by a red mark for in-phase connection.

#### FREQUENCY RESPONSE CURVE

Fig. 2. Input power 50 mW

Sound pressure measured in anechoic room, loudspeaker unmounted.

Above 1000 Hz the sensitivity may be over the width of one octave, maximum 2 dB lower than indicated.





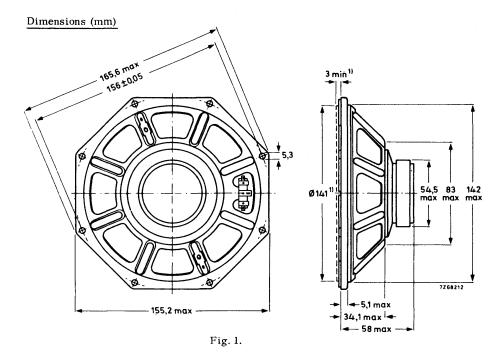
#### APPLICATION

For car and domestic radios and accoustic enclosures. High sensitivity at 4000 Hz.

TECHNICAL DATA	version		
	X4	X8	
Rated impedance	4	8	Ω
Voice coil resistance	3,4	7, 1	Ω
Resonance frequency	115	115	Hz
Power handling capacity, measured without filter loudspeaker unmounted	6	6	W
Sweep voltage	3,5	4,9	v
Energy airgap	55	55	mJ
Flux density	0, 98	0, 98	T
Airgap height	3	3	mm
Voice coil height	3	3,9	mm
Core diameter	18	18	mm
Magnet material	Magna	adur	
diameter	53	53	mm
weight	0, 1	0, 1	kg
Weight of loudspeaker	0,29	0,29	kg

The loudspeaker has a paper cone and surround and a foam plastic gasket on the flange.

Connection to the loudspeaker by means of 2.8 mm (0.11 inch) tag connectors or by soldering.



 Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

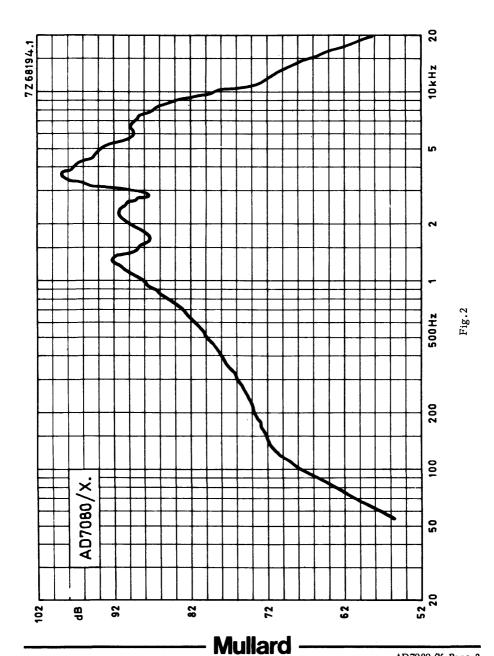
One tag is indicated by a red mark for on-phase connection.

#### FREQUENCY RESPONSE CURVE

Fig. 2. Input power 50 mW

Sound pressure measured in anechoic room, loudspeaker is unmounted.

Above 1000 Hz the sensitivity may be, over the width of one octave, maximum 2 dB lower than indicated.



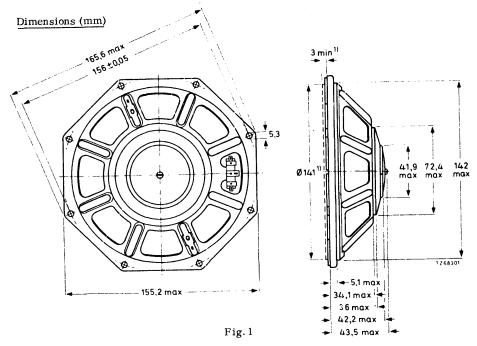
#### APPLICATION

For television sets and record players.

TECHNICAL DATA	ver	version		
	M4	M8		
Rated impedance	4	8	Ω	
Voice coil resistance	3,4	7, 1	Ω	
Resonance frequency	105	105	Hz	
Power handling capacity, measured without filter loudspeaker unmounted	3	. 3	W	
Sweep voltage	2, 45	3,5.	v	
Energy in airgap	39	39	mJ	
Flux density	0,8	0,8	T	
Airgap height	. 3	3	mm	
Voice coil height	2,4	3, 1	mm	
Core diameter	18	18	mm	
Magnet material diameter weight	Ticonal 18 0,027	Ticonal 18 0,027	mm kg	
Weight of loudspeaker	0, 22	0, 22	kg	

The loudspeaker has a dual paper cone, a paper surround and a foam plastic gasket on the flange.

Connection to the loudspeaker by means of 2.8 mm (0.11 inch) tag connectors or by soldering.



 Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

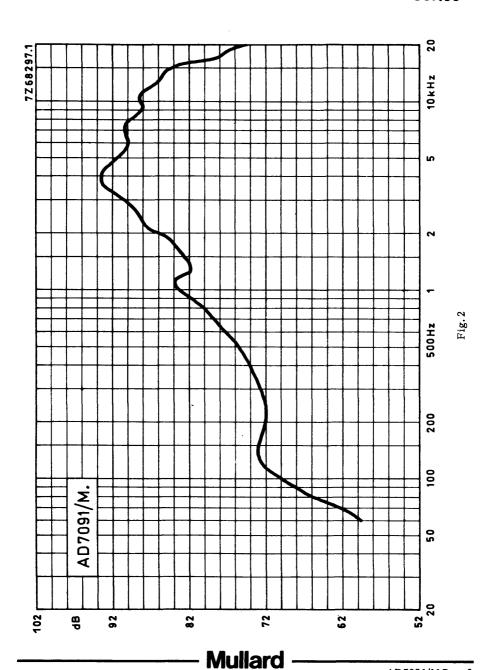
One tag is indicated by a red mark for in-phase connection.

#### FREQUENCY RESPONSE CURVE

Fig. 2 Input power 50 mW

Sound pressure measured in anechoic room, loudspeaker unmounted.

Above 1000 Hz the sound pressure may be, over the width of one octave, maximum 2 dB lower than indicated.



## 7 INCH ROUND MEDIUM POWER LOUDSPEAKER

## AD7091/X

#### APPLICATION -

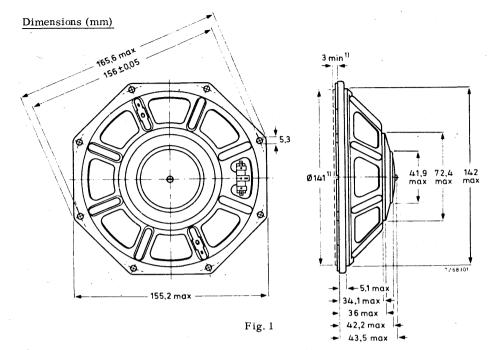
For television sets and record players.

#### TECHNICAL DATA

	ver	version	
	X4	X8	
Rated impedance	4	. 8	Ω
Voice coil resistance	3,4	7, 1	Ω
Resonance frequency	115	115	Hz
Power handling capacity, measured without filter loudspeaker unmounted	3	3	W
Sweep voltage	2, 45	3, 5	v
Energy in airgap	39	39	mJ
Flux density	0,8	0,8	T
Airgap height	3	3	mm
Voice coil height	2, 4	3, 1	mm
Core diameter	18	18	mm
Magnet material diameter weight	Ticonal 18 0,027	Ticonal 18 0,027	mm kg
Weight of loudspeaker	0, 22	0, 22	kg

The loudspeaker has a paper cone and surround and a foam plastic gasket on the flange.

Connection to the loudspeaker by means of 2.8 mm (0.11 inch) tag connectors or by soldering.



 $^{\mathrm{1}}$ ). Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

One tag is indicated by a red mark for in-phase connection.

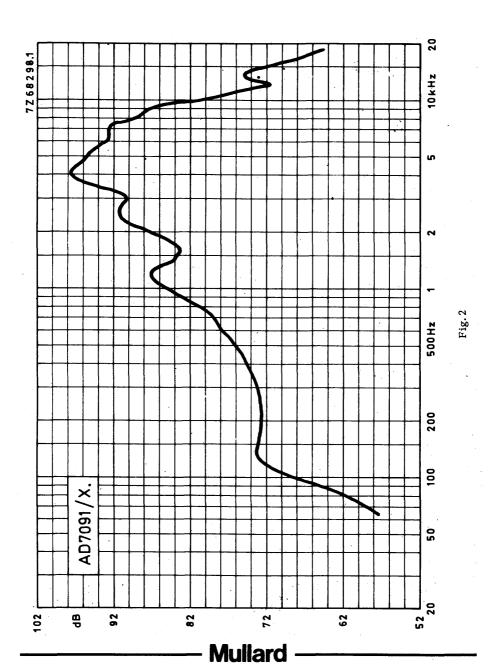
#### FREQUENCY RESPONSE CURVE

Fig. 2 Input power 50 mW

Sound pressure measured in anechoic room, loudspeaker unmounted.

Above 1000 Hz the sound pressure may be, over the width of one octave, maximum 2 dB lower than indicated.

# 7 INCH ROUND MEDIUM POWER LOUDSPEAKER

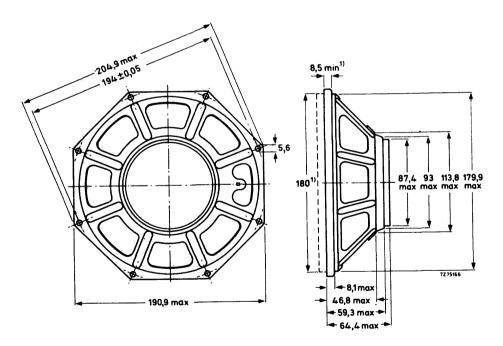


#### APPLICATION

To be used in combination with 8 inch woofer loudspeakers in a sealed enclosure for an improved bass response.

#### **TECHNICAL DATA**

Effective area	2, 5 x 10 <sup>2</sup>	m²
Moving mass:	# - <b>-</b>	
tuned mass	21,5	g
cone mass	9,8	g
total moving mass	31,3	g
Mass of radiator	0, 235	kg



<sup>1)</sup> Baffle hole and clearance depth required for cone movement.

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## 8 inch HIGH POWER WOOFER LOUDSPEAKER

#### APPLICATION

For high fidelity reproduction in sealed acoustic enclosures. Maximum enclosure volume 25 litres. Maximum recommended cross-over frequency 2000 Hz. Rated frequency range 30 to  $5\,000$  Hz.

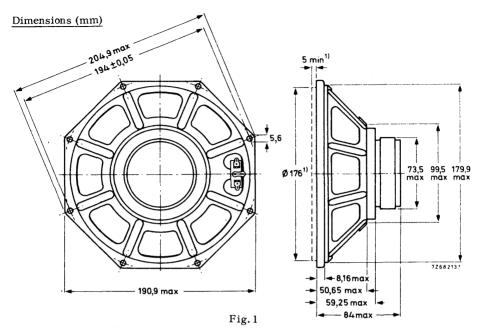
#### **TECHNICAL DATA**

	version		
	W4	w8	
Rated impedance	. 4	8	Ω
Voice coil resistance	4,3	8	Ω
Resonance frequency	42	42	Hz
Power handling capacity, measured without filter, mounted in 25 l sealed enclosure	30	30	w
Operating power	3,4	3,4	W
Sweep voltage	5	7	V
Energy in air gap	135	140	mJ
Flux density	0,87	0,93	T
Air-gap height	5	5	mm
Voice coil height	11	11	mm
Core diameter	25	25	mm
Magnet material	Magnadur		
diameter	72	72	mm
mass	0, 26	0,26	kg
Mass of loudspeaker	0,8	0,8	kg

The loudspeaker has a paper cone and a rubber surround.

Connection to the loudspeaker by means of 6.3 mm (0.25 inch) tag connectors or by soldering.





1) Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

One tag is indicated by a red mark for in-phase connection.

#### FREQUENCY RESPONSE CURVES

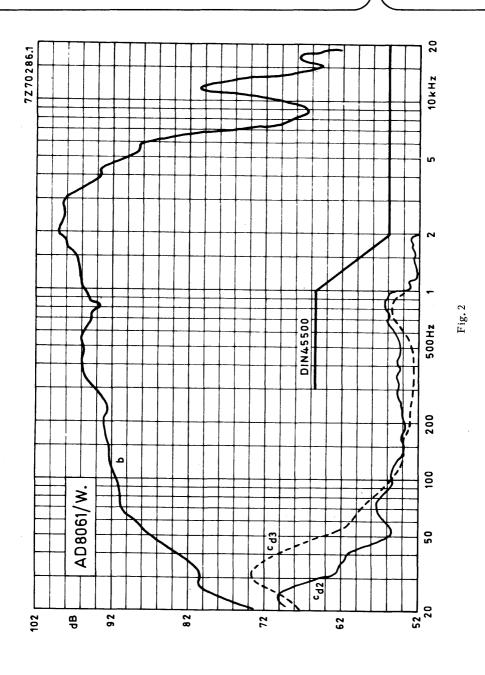
See Fig. 2

Curve b: Sound pressure measured in anechoic room at operating power. Loudspeaker mounted in sealed 80 l enclosure, filled with 1 kg of glass wool.

Curve c: 2<sup>nd</sup> and 3<sup>rd</sup> harmonic distortion, measured at the operating power of 3,4 W in anechoic room, loudspeaker mounted in 801 enclosure, filled with 1 kg of glass wool.

**Mullard** 

2







version

### 8 inch HIGH POWER WOOFER LOUDSPEAKER

#### APPLICATION

For high fidelity reproduction in sealed acoustic enclosures. Maximum enclosure volume 25 litres. Maximum recommended cross-over frequency 2500 Hz. Rated frequency range 30 to  $5\,000$  Hz.

#### **TECHNICAL DATA**

	version		
•	W4	W8	
Rated impedance	4	8	$\Omega$
Voice coil resistance	4, 3	8	$\Omega$
Resonance frequency	39	39	Hz
Power handling capacity, measured without filter, mounted in 251 sealed enclosure	40	40	w
Operating power	2, 5	2,5	W
Sweep voltage	5	7	ν
Energy in air gap	229	203	mJ
Flux density	1,1	1,2	T
Air-gap height	5	5	mm
Voice coil height	11	11	mm
Core diameter	25	25	mm
Magnet material	Magnadui		
diameter	90	90	mm
mass	0,45	0,45	kg
Mass of loudspeaker	1,15	1,15	kg

The loudspeaker has a paper cone and rubber surround.

Connection to the loudspeaker by means of 6.3 mm (0.25 inch) tag connectors or by soldering.



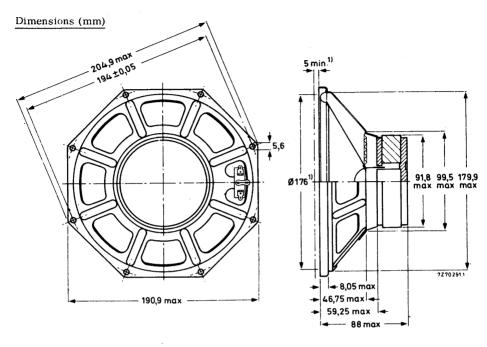


Fig. 1

1) Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

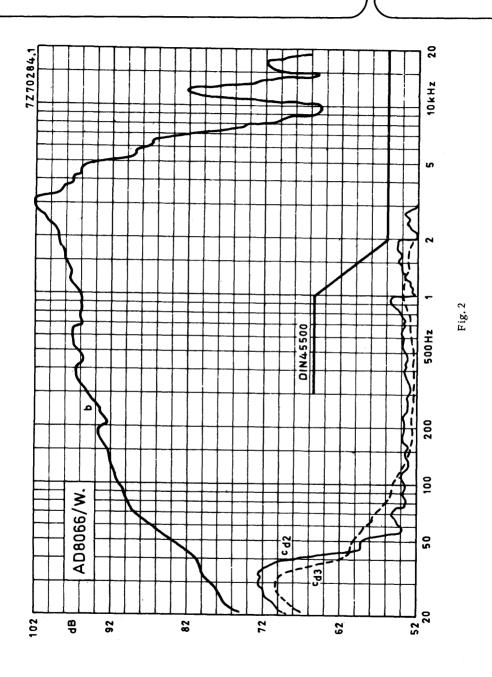
One tag is indicated by a red mark for in-phase connection.

#### FREQUENCY RESPONSE CURVES

Sec Fig. 2

Curve b: Sound pressure measured in anechoic room at operating power. Loudspeaker mounted in sealed 80 l enclosure, filled with 1 kg of glass wool.

Curve c: 2nd and 3rd harmonic distortion, measured at the operating power of 2,5 W in anechoic room, loudspeaker mounted in sealed 80 l enclosure, filled with 1 kg of glass wool.





## 8 inch HIGH POWER WOOFER LOUDSPEAKER

## APPLICATION

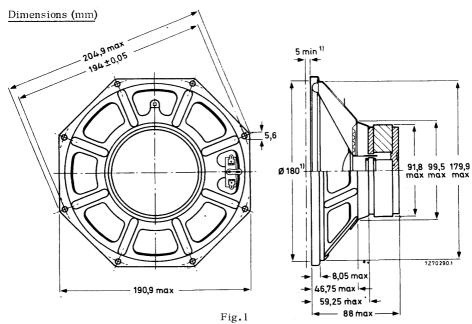
For high fidelity reproduction according to DIN45500 in scaled acoustic enclosures. Maximum enclosure volume 25 litres. Maximum recommended cross-over frequency 3000 Hz.

### **TECHNICAL DATA**

	version		
	W4	W8	
Rated impedance	4	8	Ω
Voice coil resistance	3, 2	6,4	Ω
Resonance frequency	32	32	Hz
Power handling capacity, measured without filter, mounted in 25 l enclosure	4()	40	W
Operating power	6	6	W
Sweep voltage	5	7	V
Energy in air gap	225	225	mJ
Flux density	0,7	0,7	Т
Air-gap height	5	5	mm
Voice coil height	12,7	12,8	mm
Core diameter	34	34	ກາກາ
Magnet material diameter mass	<b>Mag</b> n 90   0, 42	90° 0,42	mm kg
Mass of loudspeaker	1,3	1,3	kg

The loudspeaker has a paper cone and a rubber surround.





 $^{1}$ ) Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

One tag is indicated by a red mark for in-phase connection.

### FREQUENCY RESPONSE CURVES

See Fig. 2

Curve a: Sound pressure measured in anechoic room, loudspeaker unmounted.

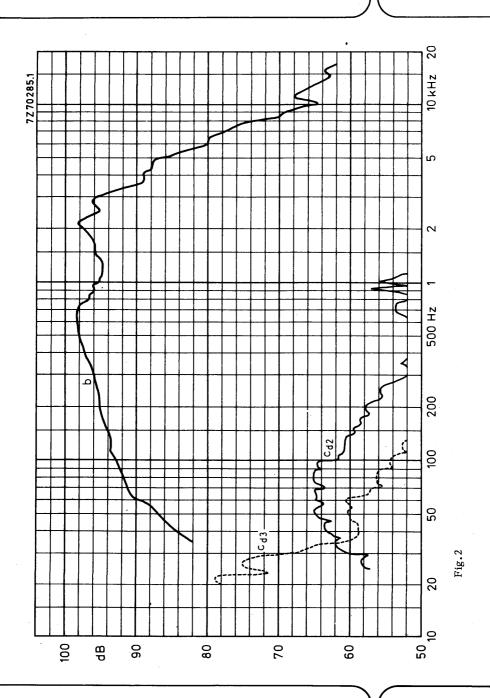
Above 1000 Hz the sound pressure may be, over the width of one octave, maximum 2 dB lower than indicated. Input power 50 mW (0,44 V).

Curve b: Sound pressure measured in half free field at operating power. Loudspeaker mounted in sealed 80 l enclosure, filled with 1 kg of glass wool.

Curve c: 2nd and 3rd harmonic distortion, measured at the operating power of 6 W in anechoic room, loudspeaker mounted in sealed 80 l enclosure, filled with 1 kg of glass wool.

**Mullard** 

Curve e: Maximum distortion according DIN45500, Blatt 7.





# 9710/M8

# 8½INCH HIGH POWER FULL RANGE LOUDSPEAKER

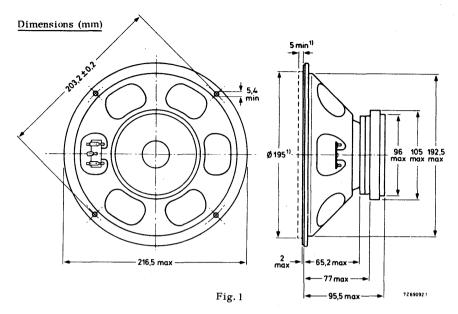
#### APPLICATION

A full range loudspeaker for studio monitoring equipment and domestic bass reflex enclosures for high fidelity reproduction from 45 Hz to 19 kHz.

#### TECHNICAL DATA

Rated impedance	8	Ω
Voice coil resistance	5	Ω
Resonance frequency	50	Hz
Power handling capacity, measured without filter, loudspeaker mounted in sealed enclosure < 30 l loudspeaker mounted in sealed enclosure > 30 l	20 10	W W
Operating power	1, 3	W
Sweep voltage	5,9	V
Energy in airgap	361	mJ
Flux density	0,75	T
Airgap height	11	mm
Voice coil height	7	mm
Core diameter	34	mm
Magnet material diameter weight	Magnadur 105 0, 4	mm kg
Weight of loudspeaker	1, 75	kg

The loudspeaker has a paper surround and a cork gasket on the flange.



 Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

One tag is indicated by a red mark for in-phase connection.

## FREQUENCY RESPONSE CURVES

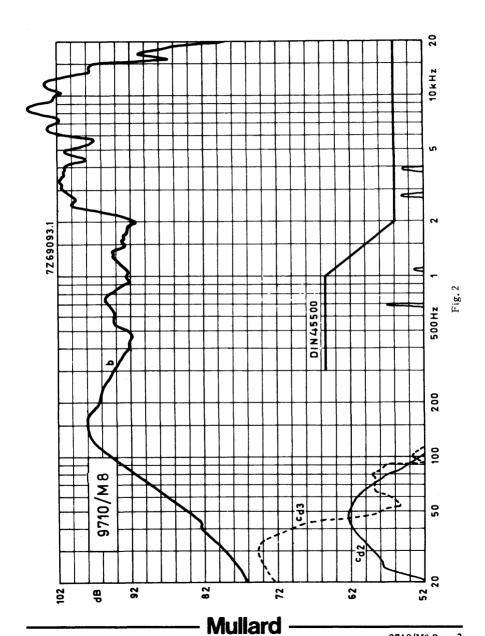
See Fig. 2.

Curve b: Sound pressure measured in anechoic room, loudspeaker mounted in sealed 80 l enclosure. Input power at operating power of 1, 3 W.

Curve c: 2nd and 3rd harmonic distortion, measured at the operating power of 0.7 W in anechoic room, loudspeaker mounted in sealed 80 l enclosure, filled with 1 kg of glass wool.



# 8½INCH HIGH POWER FULL RANGE LOUDSPEAKER





# IO INCH HIGH POWER FULL RANGE LOUDSPEAKERS

# AD1065/ M

#### APPLICATION

A full range loudspeaker with high sensitivity for public address systems in enclosures greater than  $20\ litres$ .

Smooth response from 60 Hz to 18000 Hz.

### TECHNICAL DATA

	version			
	M4	M8	M15	
Rated impedance	4	8	15	Ω
Voice coil resistance	3,4	7	13	Ω
Resonance frequency	55	55	55	Hz
Power handling capacity, measured without filter, loudspeaker unmounted	10	10	10	W
Operating power	1,5	1,5	1,5	W
Sweep voltage	4,5	6, 3	8,7	V
Energy in airgap	225	225	225	mJ
Flux density	1,12	1,12	1,12	T
Airgap height	5	5	5	mm
Voice coil height	6,5	6,5	4, 5	mm
Core diameter	25	25	25	mm
Magnet material diameter weight	90 0, <b>4</b> 5	Magnadur 90 0,45	90 0, <b>4</b> 5	mm kg
Weight of loudspeaker	1,52	1,52	1,52	kg

The loudspeaker has a paper surround and a double cone.

Connection to the loudspeaker by means of 6, 3 mm (0, 25 inch) Fastons or soldering.

## Dimensions (mm)

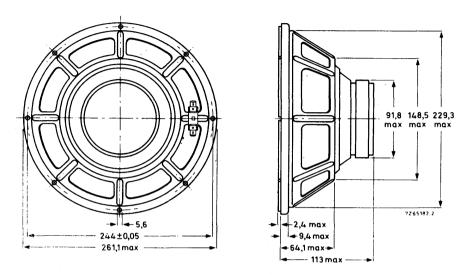


Fig. 1

Baffle hole diameter 227 mm

One tag is indicated by a red mark for in-phase connection.

## FREQUENCY RESPONSE CURVES (see Fig. 2)

Curve b: Sound pressure measured in anechoic room at operating power of 1,5 W.

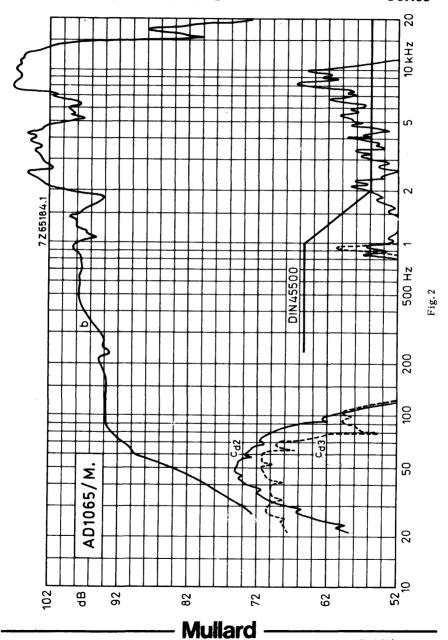
Loudspeaker mounted in sealed 80 l enclosure, filled with 1 kg of glass wool.

Curve c: 2nd and 3rd harmonic distortion, measured at operating power of 1,5 W in anechoic room. Loudspeaker mounted in 80 l enclosure, filled with 1 kg of glass wool.



# IO INCH HIGH POWER FULL RANGE LOUDSPEAKERS

# AD10**65/ M** Series



## 10 inch HIGH POWER WOOFER LOUDSPEAKER

#### APPLICATION

TECHNICAL DATA

Energy in airgap

Flux density

Airgap height

Voice coil height

For high fidelity reproduction in sealed acoustic enclosures in accordance with DIN 45500. Recommended enclosure volume 35 titres. Maximum recommended cross-over frequency 1000 Hz. Rated frequency range 40 to 3000 Hz.

version

W8

280

0,94

13, 5

5

mJ

Т

mm

mm

W4

280

0.94

12, 1

5

Rated impedance	4	8	Ω
Voice coil resistance	3,2	6,8	Ω
Resonance frequency	20	20	Hz
Power handling capacity, measured without filter, mounted in 35 l sealed enclosure	30	30	W
Operating power	5	5	w
Sweep voltage	5	7	V

Core diameter	25	25	mm
Magnet material	Mag	nadur	
diameter	90	90	mm
weight	0,45	0,45	kg
Weight of loudspeaker	1, 8	1,8	kg

The loudspeaker has a paper cone and a rubber surround.



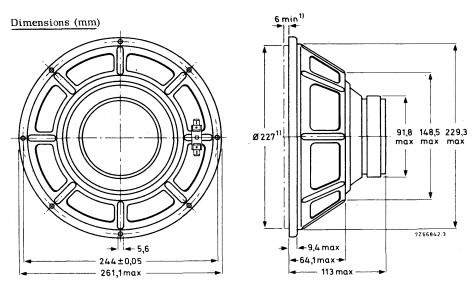


Fig. 1

1) Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

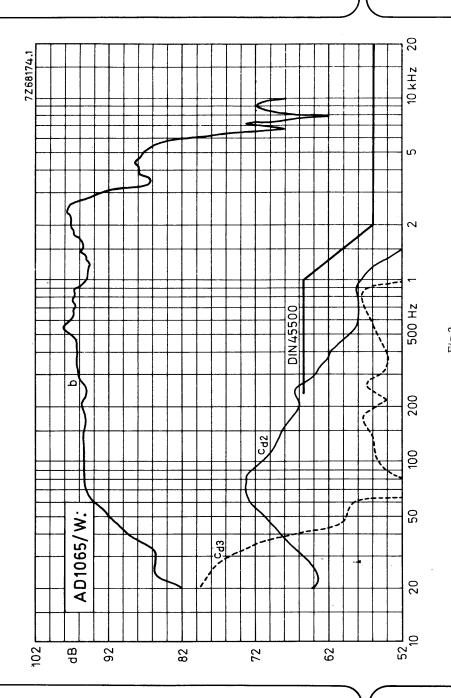
One tag is indicated by a red mark for in-phase connection.

## FREQUENCY RESPONSE CURVES (see Fig. 2)

Curve b : Sound pressure measured in anechoic room at operating power of 5 W. Loudspeaker mounted in sealed 80 l enclosure, filled with 1 kg of glass wool.

**Mullard** 

Curve  $c:2^{\underline{nd}}$  and  $3^{\underline{rd}}$  harmonic distortion, measured at operating power of 5 W in anechoic room. Loudspeaker mounted in 80 l enclosure, filled with 1 kg of glass wool.







## 10 inch HIGH POWER WOOFER LOUDSPEAKER

#### APPLICATION

For high fidelity reproduction in sealed acoustic enclosures in accordance with DIN45500. Recommended enclosure volume 35 litres. Maximum recommended cross-over frequency 800 Hz. Rated frequency range 35 to 800 Hz.

#### TECHNICAL DATA

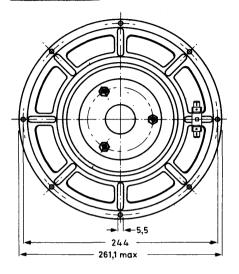
	version		
	W4	W8	
Rated impedance	4	8	Ω
Voice coil resistance	3,4	6,5	Ω
Resonance frequency	25	25	Hz
Power handling capacity, measured without filter mounted in 35 l sealed enclosure	40	40	W
Operating power	2,5	2,5	w
Sweep voltage	5	7	$\mathcal{N}$
Energy in airgap	820	820	mJ
Flux density	1,03	1,03	T
Airgap height	8	8	mm
Voice coil height	15	17, 2	mm
Core diameter	50	50	mm
Magnet material	Mag	nadur	
diameter weight	130 1,05	130 1,05	mm kg
Weight of loudspeaker	3, 0	3, 0	kg

The loudspeaker has a paper cone and a rubber surround.



## AD10100/W.

## Dimensions (mm)



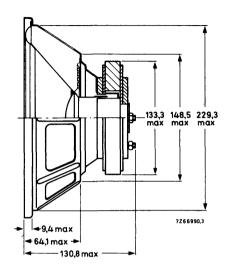


Fig. 1

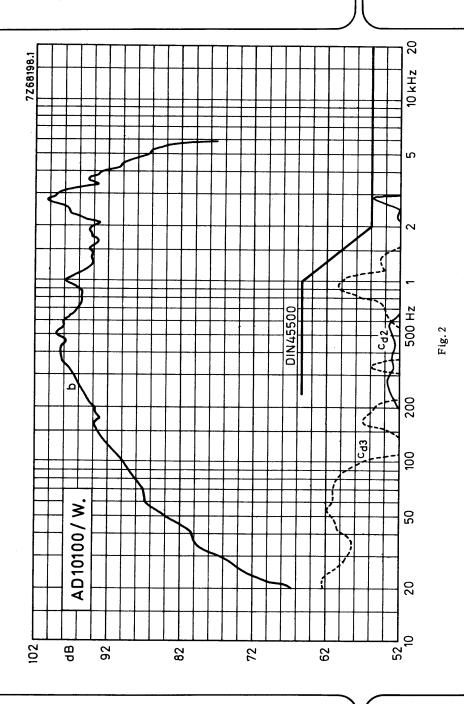
Baffle hole diameter 227 mm

One tag is indicated by a red mark for in-phase connection.

### FREQUENCY RESPONSE CURVES (see Fig. 2)

Curve b: Sound pressure measured in anechoic room at operating power of 2,5 W. Loudspeaker mounted in sealed 80 l enclosure, filled with 1 kg of glass wool.

Curve c: 2nd and 3rd harmonic distortion, measured at operating power of 2,5 W in anechoic room. Loudspeaker mounted in sealed 801 enclosure, filled with 1 kg of glass wool.





# 12 INCH HIGH POWER FULL RANGE LOUDSPEAKERS

## AD1265/M Series

### APPLICATION

Public address systems.

#### TECHNICAL DATA

	version			
	M4	M8	M15	
Rated impedance	4	8	15	75
Voice coil resistance	3,4	7	13	Ω
Resonance frequency	45	45	45	Hz
Power handling capacity, measured without filter, loudspeaker unmounted	20	20	20	W
Operating power	1, 44	1,44	1,44	W
Sweep voltage	6,3	9	12, 2	v
Energy in airgap	225	225	225	mJ
Flux density	1,12	1, 12	1, 12	T
Airgap height	5	5	5	mm
Voice coil height	6,5	6,5	4,5	mm
Core diameter	25	25	25	mm
Magnet material diameter weight	90 0, 45	Magnadur 90 0, 45	90 0, 45	mm kg
Weight of loudspeaker	1,8	1,8	1,8	kg

The loudspeaker has a paper surround and a double cone.

Connection to the loudspeaker by means of 6,3 mm (0,25 inch) Fastons or soldering.

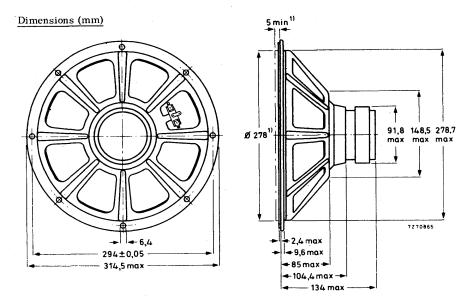


Fig. 1

 Baffle hole and clearance depth required for cone movement at the specified power handling capacity.

One tag is indicated by a red mark for in-phase connection.

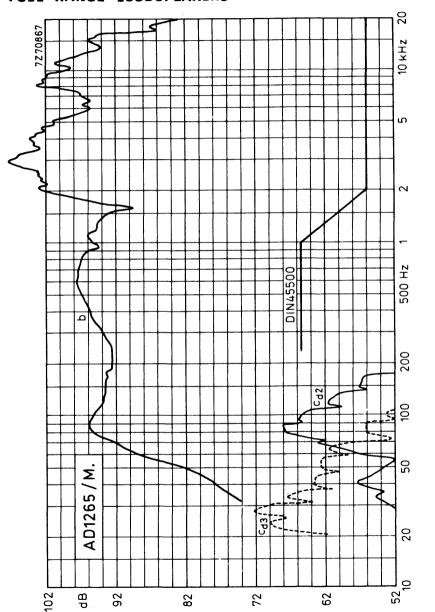
## FREQUENCY RESPONSE CURVES (see Fig. 2)

Curve b: Sound pressure measured in anechoic room at operating power of 1,44 W. Loudspeaker mounted in sealed 801 enclosure, filled with 1 kg of glass wool.

Curve c:  $2\underline{nd}$  and  $3\underline{rd}$  harmonic distortion, measured at operating power of 1,44 W in anechoic room. Loudspeaker mounted in sealed 80 l enclosure, filled with 1 kg of glass wool.

# 12 INCH HIGH POWER FULL RANGE LOUDSPEAKERS

# AD1265/M Series





## 12 inch HIGH POWER FULL RANGE LOUDSPEAKER

### **APPLICATION**

A dual cone loudspeaker for high power applications such as guitar amplifiers and electronic organs.

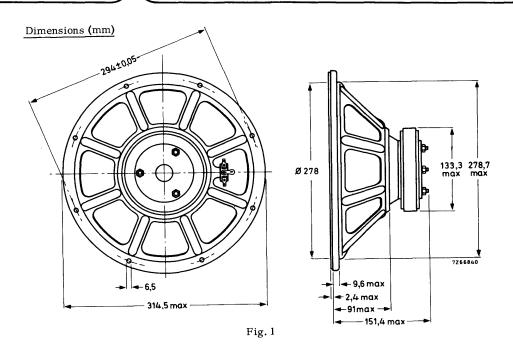
TECHN	NICAL	D/	<b>\T</b> A

I DOM WORLD DATA	ver	version		
	HP4	HP8	•	
Rated impedance	4	8	Ω	
Voice coil resistance	3, 5	7,2	Ω	
Resonance frequency	60	60	Hz	
Power handling capacity, measured without filter loudspeaker unmounted	50	50	W	
Operating power	1	1	w	
Sweep voltage	10	14	v	
Energy in airgap	820	820	mJ	
Flux density	1,03	1,03	Т	
Airgap height	8	8	mm	
Voice coil height	12, 2	12,5	mm	
Core diameter	50	50	mm	
Magnet material	l Magr	adur		
diameter wei <i>g</i> ht	130	130 1	mm kg	
Weight of loudspeaker	3, 27	3, 27	kg	

The loudspeaker has a paper cone, a textile surround and a cork gasket on the flange.



## AD12100/HP.



Baffle hole diameter 278 mm

One tag is indicated by a red mark for in-phase connection.

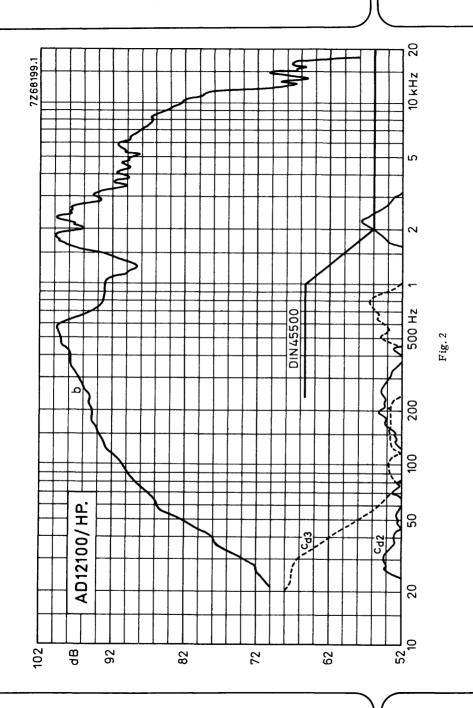
## FREQUENCY RESPONSE CURVES (see Fig. 2)

Curve b: Sound pressure measured in anechoic room at operating power of 1 W.

Loudspeaker mounted in sealed 80 l enclosure, filled with 1 kg of glass wool.

**Mullard** 

Curve c: 2nd and 3rd harmonic distortion, measured at operating power of 1 W in anechoic room. Loudspeaker mounted in 80 l enclosure, filled with 1 kg of glass wool.







## 12 inch HIGH POWER FULL RANGE LOUDSPEAKER

#### APPLICATION

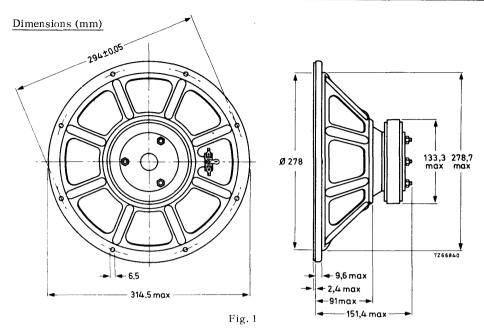
A dual-cone loudspeaker with extremely high sensitivity for power applications such as public address systems, discotheques and domestic enclosures greater than 50 litres, and open baffles.

TECHNICAL DATA	version			
	M4	M8	M15	_
Rated impedance	4	8	15	Ω
Voice coil resistance	3, 2	7	13,2	Ω
Resonance frequency	45	45	45	Hz
Power handling capacity, measured without filter, loudspeaker unmounted	25	25	25	·W
Operating power	0,55	0,55	0,6	w
Sweep voltage	6, 3	9	12, 2	v
Energy in airgap	970	970	970	mJ
Flux density	1, 15	1, 15	1, 15	T
Airgap height	8	8	8	mm
Voice coil height	9, 1	10,3	13, 3	mm
Core diameter	33,4	33,4	33, 4	mm
Magnet material diameter weight	130 1	Magnadur 130 1	130	mm kg
Weight of loudspeaker	3, 3	3, 3	3, 3	kg

The loudspeaker has a paper cone and surround and a cork gasket on the flange.

Connection to the loudspeaker by means of 6.3 mm (0.25 inch) tag connectors or by soldering.





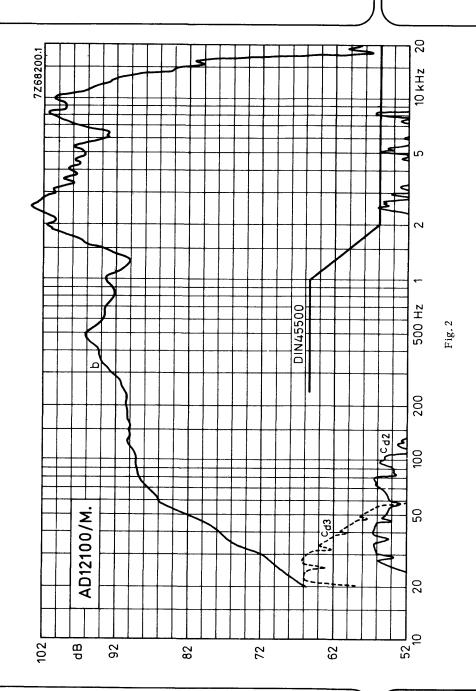
Baffle hole diameter 278 mm

One tag is indicated by a red mark for in-phase connection.

## FREQUENCY RESPONSE CURVES (see Fig. 2)

Curve b: Sound pressure measured in anechoic room at operating power of 0,55 W. Loudspeaker mounted in sealed 80 l enclosure, filled with 1 kg of glass wool.

Curve c: 2nd and 3rd harmonic distortion, measured at operating power of 0,55 W in anechoic room. Loudspeaker mounted in 80 l enclosure, filled with 1 kg of glass wool.







## 2-WAY CROSS-OVER NETWORK

## **APPLICATION**

For use in 2-way loudspeaker systems with high fidelity woofers and dome tweeters. The latter with increased impedance to obtain a higher power handling capacity for the system.

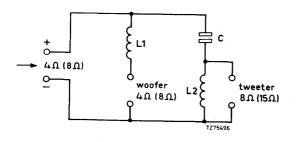
### **TECHNICAL DATA**

The state of the s	
Rated impedance	
type ADF1500/4	4 Ω
type ADF1500/8	$\Omega$ 8
Cross-over frequency	
type ADF1500/4	1500 Hz
type ADF1500/8	1800 Hz
Power handling capacity	80 W
Slope	00 11

low pass high pass

6 dB/octave 12 dB/octave

## Circuit diagram

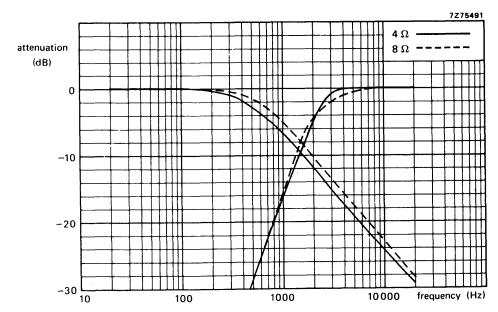


	version		
	4 Ω	8 Ω	
L1 L2	1,2 0,5	2,1 1,2	mH
C	8	3,3	· mH μF



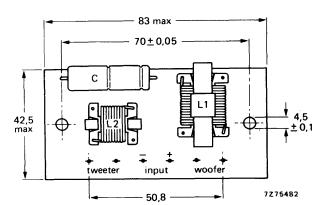
## ADF1500

## Frequency characteristics



## Dimensions (mm) and connections

Total height 35 mm 6 soldering tags for connection



## 2-WAY CROSS-OVER NETWORK

#### **APPLICATION**

For use in 2-way loudspeaker systems with high fidelity or high quality woofers and cone tweeters AD2071/T., AD2090/T. or AD2095/T.

### **TECHNICAL DATA**

Rated impedance type ADF2400/4 type ADF2400/8

Cross-over frequency

Power handling capacity

Slope

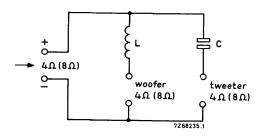
low pass high pass 4 · Ω

8Ω

2400 Hz 20 W

> 6 dB/octave 6 dB/octave

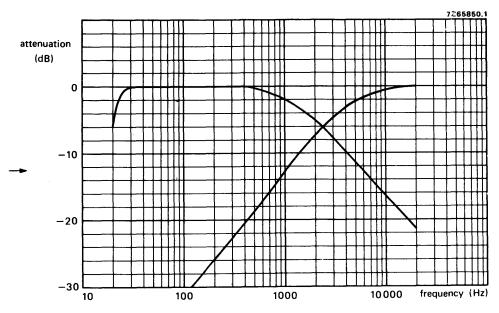
## Circuit diagram



	versi	1	
	4 Ω	Ω8_	
L C	0,5 12	1,2 5	mΗ μF

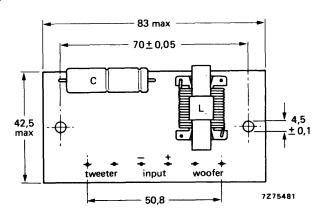


### Frequency characteristics



### Dimensions (mm) and connections

Total height 35 mm 6 soldering tags for connection



### 3-WAY CROSS-OVER NETWORK

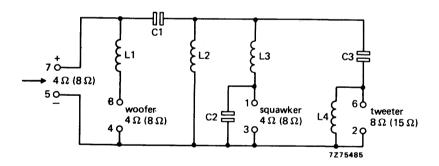
### **APPLICATION**

For use in 3-way loudspeaker systems with high fidelity woofers, squawkers and dome tweeters; the high sensitivity type tweeters AD0162/0163 should have twice the impedance of the woofer and squawker to obtain a higher power handling capacity.

### **TECHNICAL DATA**

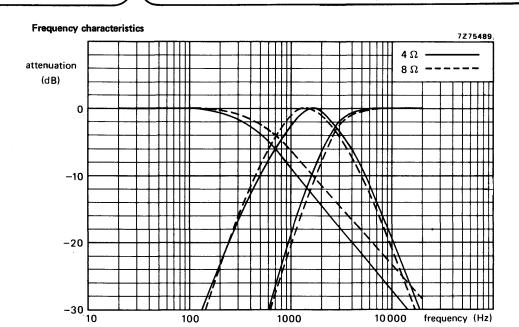
I ECHNICAL DATA	
Rated impedance type ADF700/2600/4 type ADF700/2600/8	4 Ω 8 Ω
Cross-over frequencies type ADF700/2600/4 type ADF700/2600/8	650 and 2800 Hz 700 and 2600 Hz
Power handling capacity	80 W
Slope low pass band pass (mid-range) high pass	6 dB/octave 12 dB/octave 12 dB/octave

### Circuit diagram



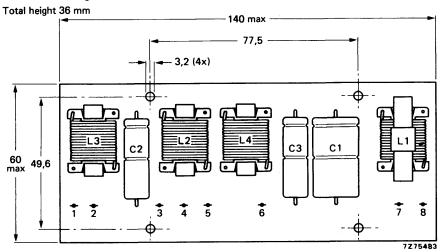
	version		
	4 Ω	$\Omega$ 8	
L1	2,1	3	mH
L2	1,2	2,1	mH
L3	0,35	0,8	mH
L4	0,5	0,8	mH
C1	36	24	μF
C2	8	5	μF
C3	5	3,3	μF





### Dimensions (mm) and connections

See also circuit diagram for connection to the 8 soldering tags.



**Mullard** 

## **TELEVISION TUNERS**

В

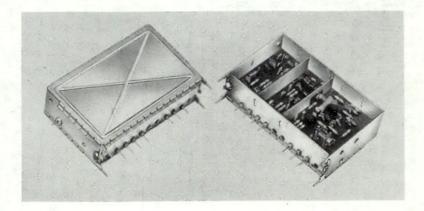


# V.H.F. TELEVISION TUNER with diode tuning

### QUICK REFERENCE DATA

Designed to cover the present 405 line v.h.f. and 625 line (wired distribution) channels. It also covers the CCIR v.h.f. channels required for export receivers.

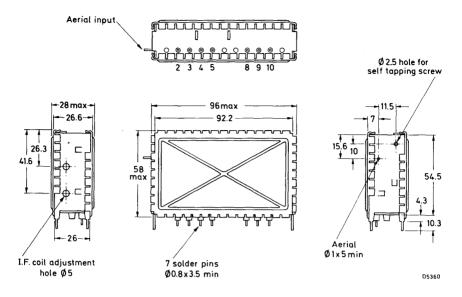
Systems	U.K. and CCIR	
Channel coverage		
band I	1 to 5 and A to C	
band III	6 to 14, D to I and CCIR	E12
Supply voltages		
transistors and switching diodes	+12	V
tuning diodes	+0.3 to +28	V
Noise factor, typical		
band I	7	dB
band III	7	dB
Power gain, typical		
band I	20	dB
band III	22	dB

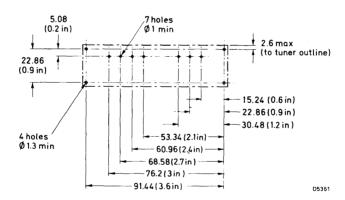


### CONSTRUCTION

The components are mounted on a printed wiring board, suitable screening is added and the whole assembly is placed in a metal box. The external connection terminations are brought out of the box, by feed-through capacitors. The i.f. coil is situated at one end of the tuner and can be adjusted through a hole provided in the end of the box.

### DIMENSIONS (millimetres) AND CONNECTIONS





Piercing diagram viewed from component side of board

Pin	Connection	Pin	Connection
2	A.G.C., positive	8	+12V, oscillator + mixer supply
3	+12V switching voltage, band III	9	Test point
4	+12V, r.f. transistor supply	10	I.F. output
5	+0.3 to +28V tuning diode supply		

### **ELC1042**

# V.H.F. TELEVISION TUNER with diode tuning

### CIRCUIT DESCRIPTION

The tuner is of the three transistor type, comprising an r.f. stage, a mixer and an oscillator. The tuning of the r.f./mixer bandpass filter and the oscillator stage is accomplished by variable capacity diodes.

Switching from band I to band III is achieved by applying a 12V supply to five diodes, via pin 3 of the tuner.

### ELECTRICAL DATA

Unless otherwise specified, all characteristics apply at an ambient temperature of  $20\pm5$  C and a relative humidity of 75% max. All values are typical unless otherwise specified.

No.	Conditions	Value
Channel coverage	41.5MHz sound carrier	channel 1
band I (see note 1)	63.25MHz sound carrier	channel 5
Channel coverage	175.25MHz vision carrier	channel D
band III (see note 1)	224.25MHz vision carrier	CCIR E12
Input impedances	asymmetrical	75Ω
Intermediate	405 line system	34.65MHz
frequencies, vision	625 line system	39.5 MHz
Intermediate	405 line system	38.15MHz
frequencies, sound 625 line system		33.5 MHz
Supply voltages	transistors tuning diodes (see note 2) switching diodes (band III only)	+12V ± 10% +0.3V to +28V +12V
	r.f. amplifier at nominal gain	3.4mA
	oscillator plus mixer at nominal gain	6.8mA
Supply currents tuning diodes		35μΑ
	switching diodes	12mA
	band I at nominal gain	2.5V
	band I at 40dB gain reduction	4.9V
A.G.C. voltage (see figs. 3 and 4)	band III at nominal gain	2.5V
(see ngs. Janu 4)	band III at 40dB gain reduction	4.3V

### ELECTRICAL DATA (contd.)

	Conditions	Value
	band I at 40dB gain reduction	max. 0.8mA
A.G.C. current	ent band III at 40dB gain reduction	
A.G.C. range	both bands	min. 40dB
V.S.W.R.	both bands at nominal gain, or with a.g.c. except channel 1	max. 3.5
R.F. bandwidth	band I channels 2 to 5 and A to C band III all channels	10 to 12MHz 10 to 25MHz
Relative levels of	any channel except channels I and CCIR E12	max, 3dB
sound and vision	channel 3	1dB
carriers (tilt)	channel 8	1dB
	any channel except channel 1	min. 18dB
	channel l	min. 16dB
Power gain	channel 2	20dB
	CCIR E4	22dB
	CCIR E5	23dB
	CCIR E12	22dB
·	any channel	max. 10dB
Noise figure	channel 3	7dB
	channel 10	7dB
	channel 2	min. 30dB
I.F. rejection	channel 5	min. 40dB
	CCIR E12	min. 60dB
Image rejection	band I any channel band III any channel	min. 60dB min. 40dB
Signal handling, signal input level producing	in channel at nominal gain, on wanted vision carrier from interfering accompanying sound carrier	band I 10mV band III 8mV
1% cross modulation (see note 3)	in band at nominal gain, on wanted vision carrier channel (X) from interfering vision carrier channel (X-2)	band I 80mV band III 25mV
Signal handling,	band I any channel at nominal gain	20mV
signal input level producing overloading	band III any channel at nominal gain	13mV
(see note 4)	both bands at 40dB gain reduction	min. 200mV
I.F. output detuning	after band switching and tuning	max. 200kHz

## V.H.F. TELEVISION TUNER with diede tuning

### ELECTRICAL DATA (contd.)

	Conditions	Value
	10% change in supply voltage	max. 300kHz
Oscillator frequency	3 to 60 second warm up after switch on	max. 50kHz
drift	input signal of 20 to 50mV at nominal gain	20kHz
	temperature change from 25 to 40°C	max. 400kHz
Temperature	operating ambient	+5 to +50°C
ranges	storage	-25 to +60°C

### NOTES

- There is a tuning margin of 2MHz at the extremes of Band III (including CCIR channel E12) and Band I (including channel C) below channel 1 and 1MHz above channel 5.
- 2. A stabilised supply of +28V is required for the tuning diodes to minimise tuning variations caused by mains fluctuations. See figures 1 and 2 for tuning voltages plotted against channel settings and frequencies.
- 3. This is the aerial e.m.f. (referred to 75Ω), which will cause the transference of 1% of the modulation of an unwanted signal to the carrier of the wanted signal.
- 4. A signal causing overloading is that aerial e.m.f. (referred to  $75\Omega$ ) which produces a 30% compression of the synchronisation pulses (of a standard television signal) or a noticeable deterioration of the picture quality.

### Oscillator radiation

The tuner is designed to meet the oscillator radiation quoted in BS905 (1969), providing the conditions for relaxation are adhered to, and no permanent connection is made to the tuner test point (pin 9). Connections between the tuner and i.f. amplifier should be kept as short as possible.

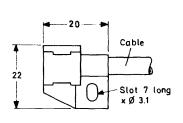
### ACCESSORIES (dimensions in millimetres)

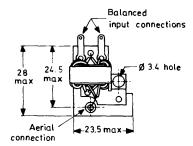
### Immunity shield

In order to meet the immunity requirements of BS905 (1969) it is recommended that the aerial connection should be screened. A suitable screening shield, fixed by a screw to the tuner is available under type No. 4313 135 01170.

### Baluns

A balun transformer, type number ELC1094, is available to convert the aerial input from  $75\Omega$  asymmetric to  $300\Omega$  balanced, for tuners incorporated into export (CCIR) receivers.





Immunity shield and cable clamp for aerial connector

3000 balun transformer D5362

### MOUNTING

The most suitable method of mounting the tuner is soldering directly on to a printed wiring board using the piercing diagram shown under DIMENSIONS AND CONNECTIONS.

SOLDERING CONDITIONS

260°C maximum for 5 seconds maximum

# V.H.F. TELEVISION TUNER with diode tuning

## **ELC1042**

AGAINST CHANNEL SETTING WITH A VISION I, F. OF 34, 65MHz

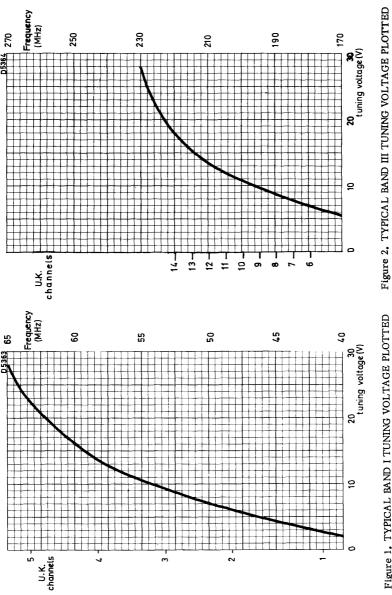
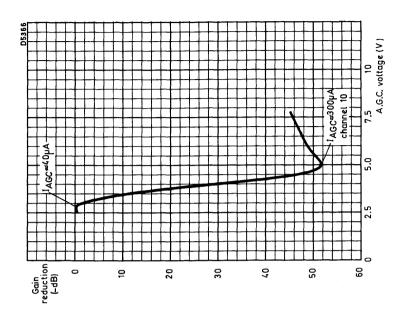


Figure 1, TYPICAL BAND I TUNING VOLTAGE PLOTTED
AGAINST CHANNEL SETTING WITH A
SOUND I, F. OF 38, 15MHz



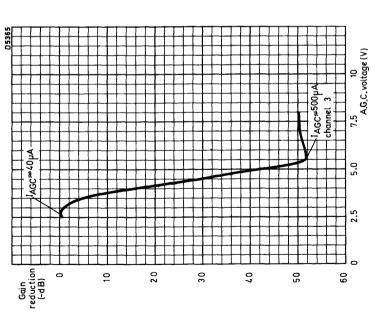


Figure 3. TYPICAL I A.G.C. VOLTAGE PLOTTED AGAINST GAIN REDUCTION

Figure 4. TYPICAL BAND III A.G.C. VOLTAGE PLOTTED AGAINST GAIN REDUCTION

### QUICK REFERENCE DATA

Designed to cover 405 lines v.h.f., and 625 lines (wired distribution) channels, and the u.h.f. channels of the C.C.I.R. system B.

Systems	C.C.I.R	C.C.I.R. systems A, B and I		
Channels	System A	System A System B System I		
band I band III	B1 to B5 B6 to B14	E2 to E4 E5 to E12	IA to IC ID to IJ	
Intermediate frequencies				
picture sound	34.65 MHz 38.15 MHz		39.5 MHz 33.5 MHz	

#### DESCRIPTION

A v.h.f. tuner with electronic tuning and band switching, covering the v.h.f. band I (frequency range 41.5 to 68 MHz, and the v.h.f. band III (frequency range 174 to 230 MHz).

Mechanically, the tuner is built on a low-loss printed-wiring board, carrying all components and housed in a metal case consisting of a rectangular frame and front and rear covers (see Fig. 2). The aerial connection is on the frame side, all other connections (supply voltages, a.g. c. voltage, tuning and switching voltages) are made via feedthrough capacitors in the underside. The mounting method is described below.

The v.h.f. aerial signal is fed via an i.f. trap to a tuned input circuit, which is connected to the emitter of the input transistor BF264. The collector load of this transistor is formed by a double tuned circuit, transferring the signal to the base of the mixer transistor BF195. The oscillator is equipped with a BF194 transistor. The three r.f. circuits are tuned by three capacitance diodes BB105G. Switching between v.h.f. I and III is achieved by five switching diodes BA182.

The collector circuit of the mixer transistor is a single tuned i.f. resonant circuit, at the low end of which the i.f. signal is capacitively coupled out of the tuner.

The tuner requires transistor supply voltages of +12 V, a switching voltage of +12 V, a.g.c. voltages variable from +2.5 V (normal operating point) to about +6 V (maximum a.g.c.), and a tuning voltage variable from +0.3 V to +25 V.

The aerial input of the tuner is asymmetrical. For use in symmetrical aerial systems, aerial transformers (baluns) are available (see ACCESSORIES).

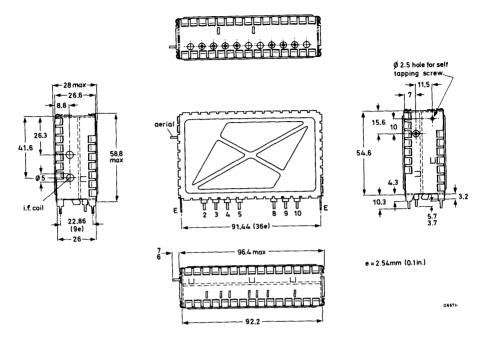


Fig. 1

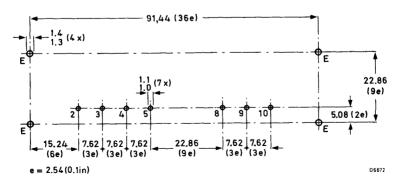
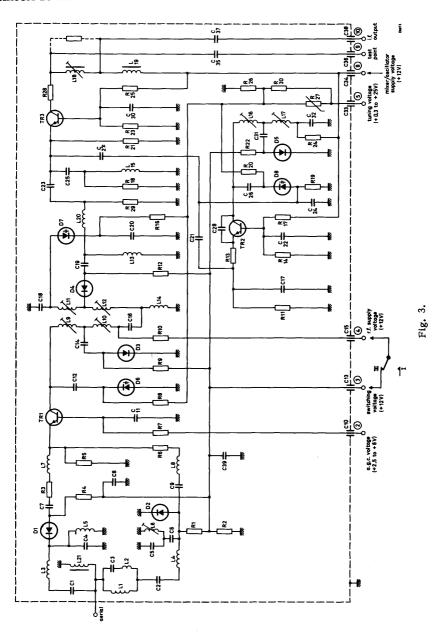


Fig. 2. Piercing diagram viewed from solder side of board.

CIRCUIT DIAGRAM



#### TERMINATIONS

2 = a.g.c. voltage, +2.5 to +6.0 V 3 = switching voltage, v.h.f. III, +12 V (approx. 12.5 mA) 4 = r.f. supply voltage, +12 V (approx 3.2 to 10 mA) 5 = tuning voltage. +0.3 to +25 V8 = mixer/oscillator supply voltage, +12 V (approx. 6.7 mA) 9 = test point 10 = i.f. output E = earth

### MOUNTING

The tuner may be mounted by soldering it on to a printed-wiring board, using the piercing diagram shown in Fig. 2. The tuner may also be mounted by means of a snap-in mount or a bracket; information will be supplied upon request.

The tuner may be mounted anywhere in the receiver and there are no restrictions on orientation.

Dimensions in millimetres

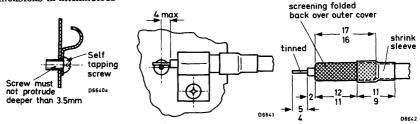


Fig. 4 Recommended fixing method of the aerial cable.

### ACCESSORIES

Aerial input transformer (balun) for converting the aerial input from 75  $\Omega$  asymmetric to  $300 \Omega$  symmetric - type no. ELC1094.

Immunity shield for screening the aerial connection - type no. 4313 132 01910

### ELECTRICAL DATA

storage

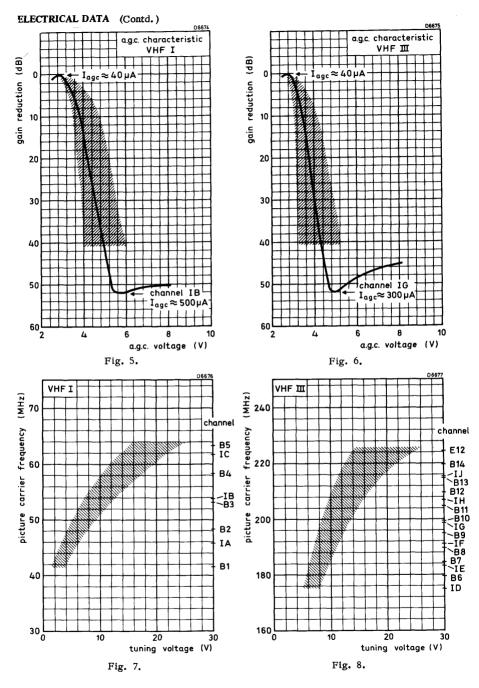
Unless otherwise specified, all electrical values apply at an ambient temperature of  $25 \pm 5$  °C and a supply voltage of  $12 \pm 0.3$  V.

Semiconductors,	
r.f. amplifier	BF264
mixer	BF195
oscillator	BF194
tuning diodes	$3 \times BB105G$
switching diodes	5 × BA182
Ambient temperature range	
operating	+5 to +55 <sup>o</sup> C
storage	-25 to +85 °C

## V.H.F. TELEVISION TUNER with diode tuning

## ELC1042/05

ELECTRICAL DATA	(Contd.)		
Supply voltage Current drawn from	n +12 V supply	+12 V, ±10%	•
band I band III		10 to 16.5 m	(
		22.5 to 29 m	A a.g.c. voltage
A.G.C. voltage (Fi	,		
at nominal ga		+2.5 V	
at 40 dB gain band III	reduction	typ. +4.9 V	
at nominal ga	in	+2.5 V	
at 40 dB gain	reduction	typ. +4.3 V	
A.G.C. current at	40 dB gain reduction		
band I		max. 0.8 m	
band III		max. 0.6 m	A
Tuning voltage rang Current drawn from	ge (Figs. 7 and 8) n 25 V tuning voltage su	$+0.3$ to $+25$ max. $30 \mu A$	<b>V</b>
Switching voltage			
band I		open circuit	
band III		+12 V, ± 10%	70
Frequency ranges	System A	System B	System J
band I	channel B1 (picture	channel E2 (picture	channel IA (picture
	carrier 45 MHz) to	carrier 48.25 MHz)	carrier 45.75 MHz)
	channel B5 (picture carrier 66, 75 MHz)	to channel E4 (picture carrier	to channel IC (picture carrier
	carrier oo. 75 Willz)	62.25 MHz)	61.75 MHz)
band III	channel B6 (picture	channel E5 (picture	channel ID (picture
	carrier 179.75 MHz)	carrier 175.25 MHz)	carrier 175.25 MHz)
	to channel B14 (picture carrier	to channel E12	to channel IJ
	219.75 MHz)	(picture carrier 224. 25 MHz)	(picture carrier 215.25 MHz)
Intermediate freque	· · ·	, <b> ,</b>	
picture	34.65 MHz	38.9 MHz	39.5 MHz
sound	38.15 MHz	33.4 MHz	33.5 MHz



# V.H.F. TELEVISION TUNER with diode tuning

band I

band III

## ELC1042/05

ELECTRICAL DATA (Contd.)			
Input impedance asymmetrical symmetrical	75 $\Omega$ 300 $\Omega$ (see ACCESSORIES)		
V.S.W.R. (between picture carrier and sound carrier)	v.s.w.r. at nom. gain.	max. v.s.w.r. during gain control	
band I (ecept channel B1) band III	max. 3.5 max. 3.5	max. 3.5 max. 3.5	
A.G.C. range band I band III	min. 40 dB min. 40 dB		
R.F. curves  bandwidth band I, except channel B1 band III  tilt band I, except channel B1 band III, except channel B1	typ. 10 to 12 typ. 9 to 20 max. 3 dB max. 3 dB		
Power gain (see also MEASURING METHOD OF POWER GAIN)			
band I, except channel B1 channel B1 channel IA channel IC band III channel ID channel IJ	min. 18 dB min. 16 dB typ. 20 dB typ. 22 dB min. 18 dB typ. 25 dB typ. 24 dB		
Noise figure band I, except channel B1 channel IB band III channel IG	max. 10 dB typ. 7.5 dB max. 10 dB typ. 7.0 dB		
I.F. rejection	••		
band I, channel B2 channel B5 band III	min. 30 dB min. 40 dB min. 60 dB		
Image rejection	mm. oo ub		
1 - 3 T	. 60 10		

min. 60 dB

min. 40 dB

### ELECTRICAL DATA (Contd.)

Signal handling (see also Figs. 9 and 10) Minimum input signal (e.m.f.) producing cross modulation (1%) at nominal gain, in channel wanted signal: picture carrier frequency, interfering signal: sound carrier frequency, typ. 8 to 14 mV typ. 6 to 10 mV v.h.f. I (note 1 overleaf) v.h.f. III in band wanted signal: picture carrier frequency of channel X, interfering signal: picture carrier of channel X -2 typ. 60 to 100 mV typ. 20 to 30 mV (note 1 overleaf) v.h.f. I v.h.f. III Minimum input signal (e.m.f.) producing overloading, at nominal gain typ. 20 mV v.h.f. I v.h.f. III typ. 13 mV Minimum input signal (e.m.f.) at nominal gain producing a shift of the oscillator typ. 20 to 50 mV (note 3 overleaf) frequency of 20 kHz Detuning of the i.f. output circuit as a result max. 200 kHz of bandswitching and tuning Shift of oscillator frequency at a change of the supply voltage of 10% band I max. 300 kHz max. 300 kHz band III during warm -up time (measured between 3 s and 60 s after switching on) band I max. 50 kHz max. 50 kHz band III Drift of oscillator frequency at a change of the ambient temperature from 25 to 40 °C max. 400 kHz band I band III max. 400 kHz

Oscillator radiation (oscillator voltages at the aerial terminal)

The oscillator radiation will be within the limits of BS905: 1969 provided no connection has been made to the test point and the circuit connected to the i.f. output is carefully shielded.

For the oscillator fundamentals, use is made of the relaxed limits, assuming that the design of the i.f. amplifier of the receiver is such that a detuning of the oscillator of > -2.0 MHz or >+0.6 MHz from the nominal frequency will result in unacceptable picture and/or sound degradation.

## V.H.F. TELEVISION TUNER with diode tuning

### ELC1042/05

### ELECTRICAL DATA (Contd.)

Immunity from radiated interference

If the tuner, including the aerial connection (see Fig. 4), is installed in a professional manner, the immunity from radiated interference will be within the limits specified in BS905: 1969.

If a higher safety margin, or another cable connection is required, use can be made of an immunity shield (see ACCESSORIES).

### Microphonics

If the tuner is installed in a professional manner, there will be no noticeable microphonics.

### NOTES:

- 1. This e.m.f. is referred to an impedance of 75  $\Omega$ .

  1% cross modulation means that 1% of the modulation depth of the interfering signal is transferred to the wanted signal.
- 2. This e.m.f. is referred to an impedance of 75  $\Omega$ . Criterion of overloading: 30% compression of the synchronization pulses of a standard television signal or a noticeable deterioration of the picture quality.
- 3. This e.m.f. is referred to an impedance of 75  $\Omega$ .

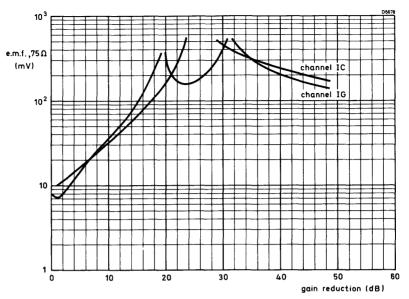


Fig. 9. Cross modulation, in channel.

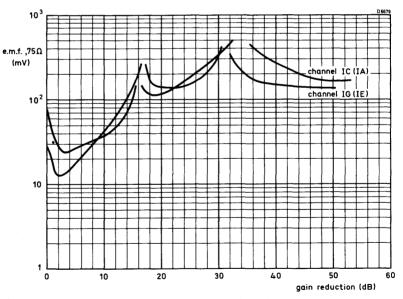
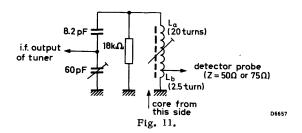


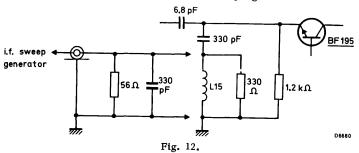
Fig. 10. Cross modulation, in band; the interfering channels are given between brackets.

### MEASURING METHOD OF POWER GAIN

 The i.f. output of the tuner should be terminated with the circuit given below, the test-point (terminal 9) not being connected.



2. Feed an i.f. sweep signal to the v.h.f. I - mixer coupling coil.



 Adjust the trimmer (Fig. 11), tunable coil La/Lb, i.f. output coil of the tuner L18 (Fig. 1), and the coupling between La and Lb to get the resonant curve as given below.

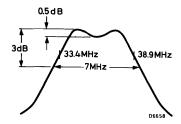


Fig. 13.

### MEASURING METHOD OF POWER GAIN (Contd.)

Display the r.f. and i.f. curve of the tuner at 190 MHz (picture carrier frequency) and make, if necessary, small corrections in the alignment of the i.f. coils  $\rm L_a/L_b$  and L18 to get the markers 39.5 MHz and 33.5 MHz symmetrically on the slopes of the curve, and the peaks at equal amplitude.

Because the output impedance of the dummy circuit is 50 to 75  $\Omega$ , the power gain can be measured in the conventional manner by inserting the tuner and dummy circuits between a 75  $\Omega$  source and a 75  $\Omega$  detector, or between a 50  $\Omega$  source, a matching pad 50/75  $\Omega$ , and a 50  $\Omega$  detector.

### OTHER AVAILABLE VERSIONS

ELC1042: This is identical with the ELC1042/05 except that the i.f. coil L18 of the ELC1042 has four additional turns.

QUICK REFERENCE DATA			
Designed for use in colour and monochrome television receivers.			
Systems	C.C.I.R. systems G an		
Channels	21 to 69		
Intermediate frequencies	ncies System G System		
picture sound	38.9 MHz 33.4 MHz	39.5 MHz 33.5 MHz	

### DESCRIPTION

A u.h.f. tuner with electronic tuning covering the u.h.f. bands IV and V (frequency range  $470\ to\ 860\ MHz$ ).

Mechanically, the tuner is built on a low-loss printed-wiring board, carrying all components, and housed in a metal case consisting of a rectangular frame and front and rear covers (see Fig. 1). The aerial connection is on the frame side, and all other connections (supply voltages, a.g. c. voltage and tuning voltage) are made via feedthrough capacitors in the underside. The mounting method is described below.

The tuner is of the three transistor type, comprising two r.f. stages and an oscillator/mixer. The input circuit is untuned, so that optimum noise figures may be realised, whilst the additional r.f. stage compensates for the increased insertion loss associated with diode tuned circuits.

The coupling between the first and second r.f. stages is by a half-wave tuned line; the coupling between the second r.f. stage and the mixer is by bandpass half-wave tuned lines. The secondary of the passband is coupled to the emitter of the oscillator/mixer stage via a coupling loop, which also provides the inductive feedback for the oscillator.

Half-wave lines, terminated at one end by a fixed capacitor and tuned at the other end by a variable capacitance diode, are used throughout.

The tuner requires transistor supply voltages of +12 V, a.g.c. voltages variable from +2.5 V (normal operating point) to about +7.5 V (maximum a.g.c.), and a tuning voltage variable from +0.3 V to +25 V.

The aerial input of the tuner is asymmetrical. For use in symmetrical aerial systems, aerial transformers (baluns) are available (see ACCESSORIES).

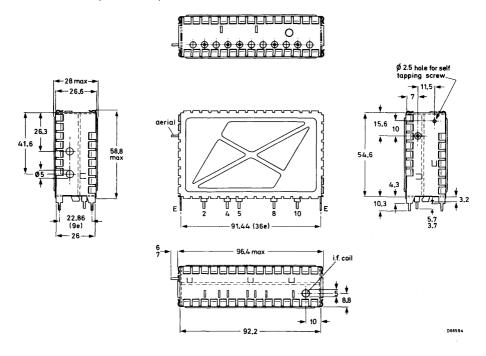


Fig. 1.

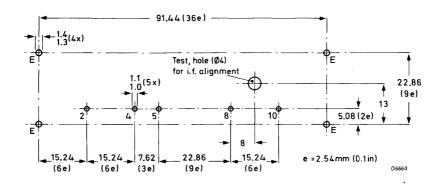


Fig. 2. Piercing diagram viewed from solder side of board.

### CIRCUIT DIAGRAM.

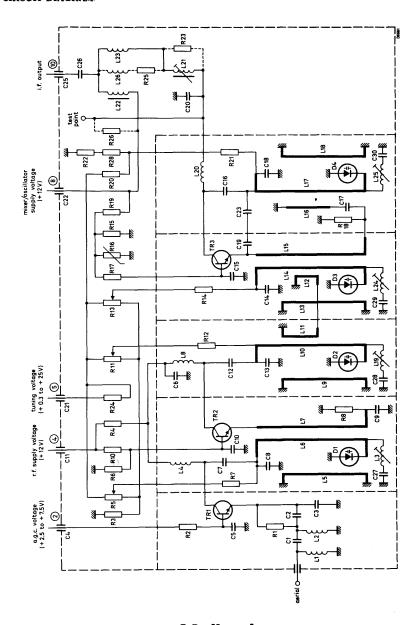


Fig. 3.

### TERMINATIONS

2 = a.g.c. voltage, +2.5 to +7.5 V

4 = r.f. supply voltage, +12 V (approx. 8.8 to 13 mA)

5 = tuning voltage, +0.3 to +25 V

8 = mixer/oscillator supply voltage, +12 V (approx. 3.6 mA)

10 = i.f. output

E = earth

#### MOUNTING

The tuner may be mounted by soldering it on to a printed-wiring board, using the piercing diagram shown in Fig. 2. The tuner may also be mounted by means of a snap-in mount or a bracket; information will be supplied on request.

The tuner may be mounted anywhere in the receiver and there are no restrictions on orientation.

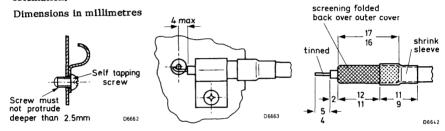


Fig. 4. Recommended fixing method of the aerial cable.

### ACCESSORIES

Aerial input transformer (balun) for converting the aerial input from 75  $\Omega$  asymmetric to 300  $\Omega$  symmetric - type no. ELC1095.

Immunity shield for screening the aerial connection - type no. 4313 132 01910.

#### ELECTRICAL DATA

Unless otherwise specified, all electrical values apply at an ambient temperature of  $25\pm5$   $^{o}C$  and a supply voltage of  $12\pm0.3$  V.

Semiconductors, r.f. amplifiers mixer/oscillator tuning	2 × BF362 BF363 4 × BB105B
Ambient temperature range operating storage	+5 to +55 <sup>o</sup> C -25 to +85 <sup>o</sup> C
Supply voltage Current drawn from +12 V supply	$+12 \text{ V} \pm 10\%$
r.f. amplifiers	8.8 mA (at nominal gain) to 12 to 17 mA (at 30 dB gain reduction)
mixer/oscillator	3.6 mA

## U.H.F. TELEVISION TUNER with diode tuning

### **ELC1043/05**

### ELECTRICAL DATA (Contd.)

A.G.C. voltage (Fig. 5) at nominal gain at 30 dB gain reduction

A.G.C. current at 30 dB gain reduction

Tuning voltage range Slope of tuning characteristic Current drawn from 25 V tuning voltage supply

Frequency range

Margin at the extreme channels

Intermediate frequencies

picture sound

Input impedance asymmetrical symmetrical

V.S.W.R. (between picture carrier and sound carrier)

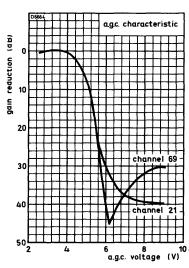


Fig. 5.

+2.5 V 6 V (max. 7.5 V)

max. 1.2 mA

+0.3 to +25 V min. 5 MHz/V

max. 20 μA

channel 21 (picture carrier 471.25 MHz) to channel 69 (picture carrier 855.25 MHz).

min. 3 MHz

system G	System I
38.9 MHz	39.5 MHz
33.4 MHz	33.5 MHz

75 Ω

 $300 \Omega$  (see ACCESSORIES)

at nom. gain without a.g.c.	during gain control up to 30 dB
max. 4	max. 6

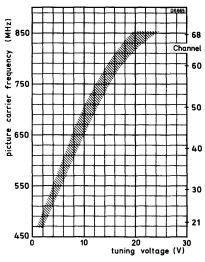


Fig. 6.

ELECTRICAL DATA (Contd.)	
A.G.C. range	min. 30 dB
R.F. curves bandwidth tilt	typ. 10 to 20 MHz 4 dB (0 to 2 dB typical)
Power gain (see also MEASURING METHOD OF POWER GAIN) channel 21 channel 50 channel 69	min. 17 dB typ. 22 dB typ. 22 dB typ. 22 dB
Noise figure channel 21 channel 50 channel 69	max. 10 dB typ. 6 dB typ. 6.5 dB typ. 7 dB
I.F. rejection	min. 60 dB
Image rejection, channels 21 to 61	min. 53 dB
n+4 rejection (obtained between the picture carrier of the wanted channel n and the sound carrier of an unwanted signal spaced 4 channels above the wanted channel).	min. 53 dB
Signal handling  Minimum input signal (e.m.f.) producing cross modulation of 1% at nominal gain,  in channel  wanted signal: picture carrier frequency, interfering signal: sound carrier frequency.	typ. 8 mV (note 1 overleaf)
in band wanted signal: picture carrier frequency of channel X, interfering signal: picture carrier of channel X-5	typ. 25 mV (note 1 overleaf)
Minimum input signal (e.m.f.) producing	
overloading, at nominal gain at maximum a.g.c.	typ. 15 to 20 mV (note 2 overleaf) min. 250 mV (note 2)
Minimum input signal (e.m.f.) at nominal gain producing a shift of the oscillator frequency of 20 kHz	typ. 5 to 15 mV (note 3 overleaf)
Detuning of the i.f. output circuit as a result of tuning	max. 150 kHz
Shift of oscillator frequency at a change of the supply voltage of 10% during warm -up time (measured between 3 s	max. 500 kHz
and 60 s after switching on)	max. 200 kHz
at a gain reduction of 30 dB	max. 100 kHz
Drift of oscillator frequency at a change of the ambient temperature from 25 to 50 °C	max. 1000 kHz



## U.H.F. TELEVISION TUNER with diode tuning

## **ELC1043/05**

### ELECTRICAL DATA (Contd.)

Oscillator radiation (oscillator voltages at the aerial terminal)

The oscillator radiation will be within the limits of BS905: 1969 provided the circuit, connected to the i.f. output, is carefully shielded.

For the oscillator fundamentals, use is made of the relaxed limits, assuming that the design of the i.f. amplifier of the receiver is such that a detuning of the oscillator of > -2.0 MHz or > +0.6 MHz from the nominal frequency will result in unacceptable picture and/or sound degradation.

### Immunity from radiated interference

If the tuner, including the aerial connection (see Fig. 4) is installed in a professional manner, the immunity from radiated interference will be within the limits specified in BS905: 1969.

If a higher safety margin or another cable connection is required, use can be made of an immunity shield (see ACCESSORIES).

### Microphonics

If the tuner is installed in a professional manner, there will be no noticeable microphonics.

### NOTES:

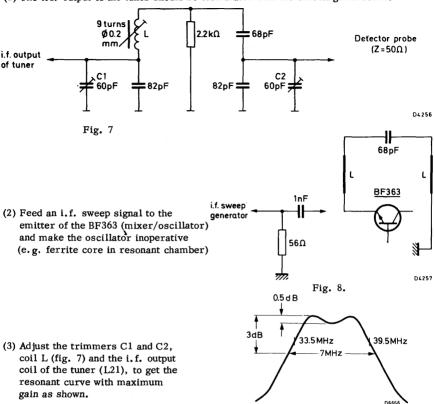
- 1. This e.m.f. is referred to an impedance of 75  $\Omega$ .

  1% cross modulation means that 1% of the modulation depth of the interfering signal is transferred to the wanted signal.
- 2. This e.m.f. is referred to an impedance of 75  $\Omega$ . Criterion of overloading: 30% compression of the synchronization pulses of a standard television signal or a noticeable deterioration of the picture quality.
- 3. This e.m.f. is referred to an impedance of 75  $\Omega$ .



### MEASURING METHOD OF POWER GAIN

(1) The i.f. output of the tuner should be terminated with the circuit given below.



(4) Display the r.f. and i.f. curve of the tuner at 470 MHz and make, if necessary, small corrections in the alignment of C1, C2, and L and L21 to get the markers 38.9 MHz and 33.4 MHz symmetrically on the slopes of the curve and the peaks at equal amplitude.

Fig. 9.

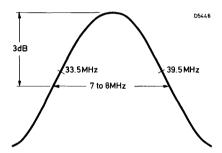
Because the output impedance of the dummy circuit is 50 to 75  $\Omega$ , the power gain can be measured in the conventional manner by inserting the tuner and the dummy circuit between a 75  $\Omega$  source and a 75  $\Omega$  detector, or between a 50  $\Omega$  source, a matching pad 50/75  $\Omega$  and a 50  $\Omega$  detector.

### OTHER AVAILABLE VERSION

ELC1043/06

This type is identical to ELC1043/05 except for the following:-

- The i.f. output circuit consists of an i.f. coil with increased turns damped by a 680Ω resistor. This changes the tuning range and 'Q' of the i.f. output coil, making it suitable for coupling to a block filter input i.f. amplifier.
- 2. The power gain is reduced to a nominal of 12dB and a minimum of 9dB by the damping resistor mentioned above.
- 3. The same dummy circuit is used for measuring power gain but, as a result of damping the i.f. coil, a single tuned response will appear at the output, instead of the double tuned response, as in the case of ELC1043/05. (see below).





## V.H.F./U.H.F. TELEVISION TUNER with diode tuning

### **ELC2000**

### QUICK REFERENCE DATA

Designed to cover the v.h.f. and u.h.f. channels of the C.C.I.R. systems B and G. including the Italian channels.

Systems

C.C.I.R. systems B and G

Channels

E2 to C (band I) E5 to E12 (band III)

E21 to E69 (bands IV and V)

Intermediate frequencies

picture sound 38.9 MHz 33.4 MHz

### DESCRIPTION

A combined v.h.f./u.h.f. tuner with electronic tuning and band switching, covering the v.h.f. band I including the Italian channel C (frequency range 47 to 88 MHz), the v.h.f. band III (frequency range 174 to 230 MHz), and the u.h.f. band (frequency range 470 to 860 MHz).

Mechanically, the tuner is built on a low-loss printed-wiring board, carrying all components, and housed in a metal case consisting of a rectangular frame and front and rear covers (see Fig. 1). The two aerial connections (v.h.f. and u.h.f.) are on the two frame sides, all other connections (supply voltages, a.g.c. voltage, tuning and switching voltages) are made via feedthrough capacitors in the underside. The mounting method is described below.

Electrically, the tuner consists of a v.h.f. and u.h.f. part. The v.h.f. aerial signal is fed via an i.f. trap, combined with a high pass filter, to a tuned input circuit, which is connected to the emitter of the input transistor BF200. The collector load of this transistor is formed by a double tuned circuit, transferring the signal to the base of the mixer transistor BF182. The oscillator is equipped with a transistor BF194. The four r.f. circuits are tuned by four capacitance diodes BB106. Switching between v.h.f. I and III is achieved by four switching diodes BA243/244.

The collector circuit of the mixer transistor is a single tuned i.f. resonant circuit, at the low end of which the i.f. signal is capacitively coupled out of the tuner. An i.f. injection point is provided at the collector of the mixer, for aligning this circuit together with the i.f. amplifier of the television receiver.

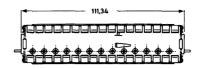
The u.h.f. part of the tuner consists of a tuned input circuit, connected to the emitter of the amplifier transistor BF180. The interstage network between this transistor and the self-oscillating mixer stage is formed by a double tuned circuit. A transistor BF181 acts as a self-oscillating mixer. The four tuned u.h.f. circuits are tuned by four capacitance diodes BB105B.

The output of the self-oscillating mixer is fed to a double tuned i.f. circuit which is connected to the emitter of the v.h.f. mixer transistor BF182, now operating as an i.f. amplifier in grounded base configuration. Band switching between v.h.f. and u.h.f. is achieved by another diode BA243.

The tuner requires transistor supply voltages of +12 V, a switching voltage of +12 V, a.g.c. voltages, variable from +2.4 V (normal operating point) to about +7.5 V (maximum a.g.c.) and a tuning voltage, variable from +0.5 V to +28 V.

The aerial inputs of the tuner are assymmetrical. For use in symmetrical aerial systems, aerial transformers (baluns) are available (see ACCESSORIES).





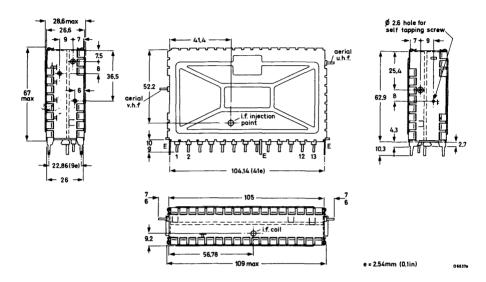


Fig. 1.

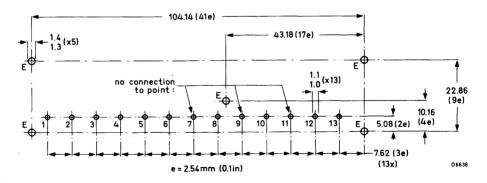
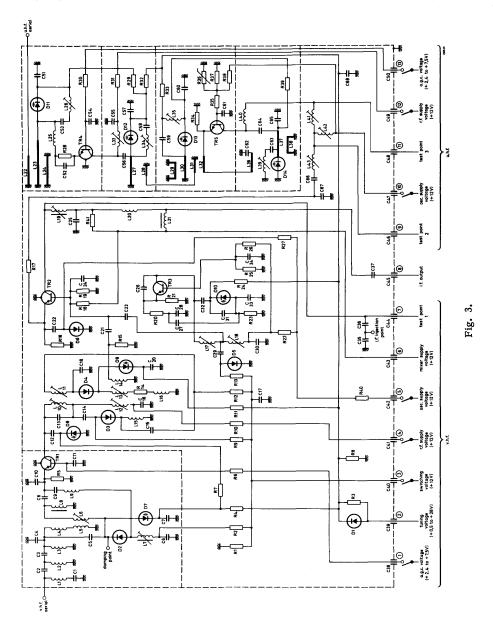


Fig. 2. Piercing diagram viewed from solder side of board. No connection must be made to terminals 7, 9 and 11, as otherwise the oscillator radiation would increase.

# V.H.F./U.H.F.TELEVISION TUNER with diode tuning

# **ELC2000**

# CIRCUIT DIAGRAM



#### **TERMINATIONS**

```
1 = a.g.c. voltage, v.h.f., +2.4 to +7.5 V
2 = tuning voltage, +0.5 to +28 V
3 = switching voltage, +12 V (approx. 20 mA)
4 = r.f. supply voltage, v.h.f., + 12 V (approx. 3 to 10 mA)
5 = oscillator supply voltage, v.h.f., +12 V (approx. 6 mA)
6 = mixer supply voltage, v.h.f., +12 V (approx. 5 mA)
7 = test point 1, v.h.f.
8 = i.f. output
9 = test point 2 (alignment short)
10 = oscillator supply voltage, u.h.f., +12 V (approx. 4 mA)
11 = test point 3, u.h.f.
12 = r.f. supply voltage, u.h.f., +12 V (approx. 2.5 to 9.5 mA)
13 = a.g.c. voltage, u.h.f., +2.4 to +7.5 V
E = earth
```

#### MOUNTING

The tuner may be mounted by soldering it on to a printed-wiring board, using the piercing diagram shown in Fig. 2. The tuner may also be mounted by means of a snap-in mount or a bracket; information will be supplied upon request.

The tuner may be mounted anywhere in the receiver and there are no restrictions on orientation.

#### Dimensions in millimetres

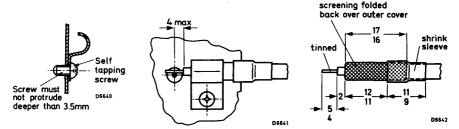


Fig. 4. Recommended fixing method of the aerial cables.

#### ACCESSORIES

Aerial input transformer (balun) v.h.f. - type no. ELC1094 Aerial input transformer (balun) u.h.f. - type no. ELC2092

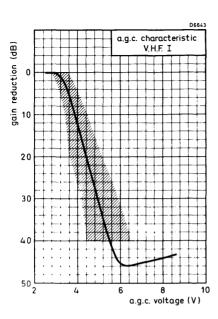
# V.H.F./U.H.F. TELEVISION TUNER with diode tuning

# **ELC2000**

## **ELECTRICAL DATA**

Unless otherwise specified all electrical values apply at an ambient temperature of 25  $\pm$  5  $^{O}C$  and a supply voltage of 12  $\pm$  0.3 V.

Semiconductors bands I and III,		
r.f. amplifier mixer oscillator tuning diodes	BF200 BF182 BF194 4 × BB106	
switching diodes	$5 \times BA243/244$	
bands IV and V, r.f. amplifier mixer/oscillator tuning diodes drift compensating diode	BF180 BF181 4 × BB105B BAW62	
Ambient temperature range operating storage	+5 to +55 <sup>O</sup> C -25 to +85 <sup>O</sup> C	
Supply voltage Current drawn from +12 V supply band I band III bands IV and V	+12 V ± 10%  14 to 21 mA 34 to 41 mA 31.5 to 38 mA	depending on a.g.c. voltage
A.G.C. voltage (Figs. 5, 6 and 7) band I, at nominal gain at 40 dB gain reduction	2.4 V typ. 5.5 V	,
band III, at nominal gain at 40 dB gain reduction	2.4 V typ. 4.5 V	
bands IV and V, at nominal gain at 30 dB gain reduction	2.4 V typ. 5.0 V	
A.G.C. current band I at 40 dB gain reduction band III at 40 dB gain reduction bands IV and V at 30 dB gain reduction	max. 0. 8 mA max. 0. 6 mA max. 0. 7 mA	
Tuning voltage range (Figs. 8, 9 and 10) Current drawn from 28 V tuning voltage supply	$+0.5$ to $+28$ V max. 36 $\mu$ A	
Switching voltage band I band III band IV and V	open circuit +12 V +12 V	



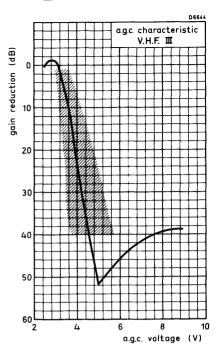


Fig. 5

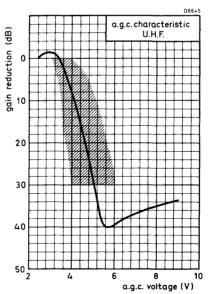


Fig. 7

Fig. 6

# V.H.F./U.H.F. TELEVISION TUNER with diode tuning

# **ELC2000**

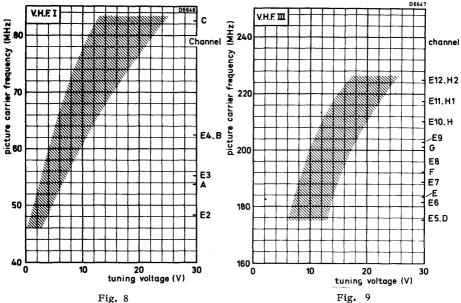


Fig. 8

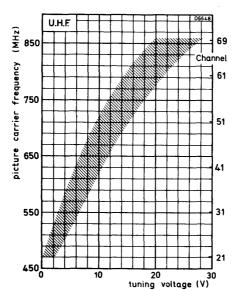


Fig. 10

## **ELECTRICAL DATA** (continued)

Frequency ranges

band I

channel E2 (picture carrier 48.25 MHz) to channel C (picture carrier 82.25 MHz).

margin at the extreme channels

band III

min. 1.2 MHz

channel E5 (picture carrier 175.25 MHz) to channel E12 (picture carrier 224, 25 MHz).

margin at the extreme channels

bands IV and V

min. 2 MHz

channel E21 (picture carrier 471.25 MHz) to channel E69 (picture carrier 855.25 MHz).

300  $\Omega$  (see ACCESSORIES)

margin at the extreme channels

min. 3 MHz

Intermediate frequencies

picture sound

38.9 MHz 33.4 MHz

**75** Ω

Input impedance

asymmetrical symmetrical

V.S.W.R. (between picture carrier and sound carrier)

> band I (except channel C) band III (except channel E12) bands IV and V

at nom. gain | during gain control

worst best best worst va lue va lue va lue va lue (max.) (max.) (max.) (max.) 3 5 3 4 4 5 5

A.G.C. range

band I band III

bands IV and V

min. 40 dB min. 40 dB min. 30 dB

R.F. curves

bandwidth

band I band III

bands IV and V

typ. 10 to 15 MHz typ. 10 to 15 MHz typ. 15 to 25 MHz

ti lt

band I band III

bands IV and V, channels E21 to E60

channels E61 to E69

max. 3 dB

max. 3 dB max. 3 dB

max. 4 dB

# **ELECTRICAL DATA** (continued)

Power gain (see also MEASURING METHOD OF POWER GAIN)

band I	min.	26 đB	
channel E2	typ.	29 dB	
channel C	typ.	32 dB	
band III	min.	25 dB	
channel E5	typ.	28 dB	
channel E11	typ.	28 dB	
bands IV and V	min.	25 dB	
channel E21	typ.	32 dB	
channel E31	typ.	29 dB	
channel E69	typ.	33 dB	
Noise figure			
band I	max.	8.5 dB	
channel E4	typ.	6.5 dB	
band III	max.	8 dB	
channel E9	typ.	6.5 dB	
bands IV and V	max.	12 dB	
channel E21	typ.	8.0 dB	
channel E51	typ.	9.5 dB	
channel E68	typ.	10.5 dB	
I.F. rejection			
band I, channel E2	min.	40 dB	
channel C	min.	60 dB	
band III	min.	60 dB	
bands IV and V	min.	60 dB	
Image rejection			
hand I	min.	40 dB	
hand III	min.	60 dB	
bands IV and V	min.	40 dB	
Signal handling (see also Figs. 12 and 13)  Minimum input signal (e.m.f.) producing	******	10 QD	
cross modulation of 1% at nominal gain, in channel			
wanted signal: picture carrier frequency, interfering channel: sound carrier			
frequency, v.h.f. I	typ.	4 mV ]	
v.h.f. III	typ.	4 mV	note 1
u.h.f.		5 to 10 mV	note 1

continued on next page)

#### ELECTRICAL DATA (continued)

```
Signal handling, (continued)
```

```
in band
```

wanted signal: picture carrier frequency of channel X, interfering signal: picture carrier of channel X-2 (v.h.f.), X-5 (u.h.f.) v.h.f. I v.h.f. III u.h.f.

typ. 15 to 60 mV typ. 10 to 50 mV typ. 15 to 50 mV

Minimum input signal (e.m.f.) producing overloading,

at nominal gain at maximum a.g.c.

typ. 10 mV note 2

Minimum input signal (e.m.f.) at nominal gain producing a shift of the oscillator frequency of 10 kHz.

band I band III bands IV and V min. 25 mV min. 25 mV typ. 10 to 20 mV

max. 400 kHz

Detuning of the i.f. output circuit as a result of bandswitching and tuning with respect of channel E8

Shift of oscillator frequency at a change of the

supply voltage of 10%

 band I
 max. 300 kHz

 band III
 max. 300 kHz

 bands IV and V
 max. 600 kHz

during warm-up time (measured between 5 s and 15 min after switching on)

band I max. 100 kHz
band III max. 100 kHz
bands IV and V max. 250 kHz
at a gain reduction of 30 dB max. 100 kHz

Drift of oscillator frequency at a change of the ambient temperature from 25 to 40  $^{\rm o}{\rm C}$ 

#### Oscillator radiation

The tuner conforms with the radiation requirements of C.I.S.P.R. Recommendation No. 24/3, provided the following conditions are fulfilled:

A low-pass filter (Fig. 11) with a cut-off frequency of about 300 MHz has to be inserted between the v.h.f. aerial terminal of the tuner and the aerial terminal of the receiver. Television receivers with a common v.h.f./u.h.f. connector in combination with a low-pass/high-pass splitter do not need this additional filter.

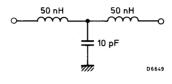


Fig. 11

No connections must be made to the terminals 7, 9 and 11.

Earthing of the tuner and connection to the i.f. amplifier must be made in such a way, that additional radiation is prevented.

#### Microphonics

If the tuner is installed in a professional manner, there will be no appreciable microphonics.

#### NOTES

- 1. This e,m.f. is referred to an impedance of 75  $\Omega$ .
  1% cross modulation means that 1% of the modulation depth of the interfering signal is transferred to the wanted signal.
- This e.m.f. is referred to an impedance of 75 Ω.
   Criterion of overloading: 30% compression of the synchronization pulses of a standard television signal or a noticeable deterioration of the picture quality.
- 3. This e.m.f. is referred to an impedance of 75  $\Omega$ .

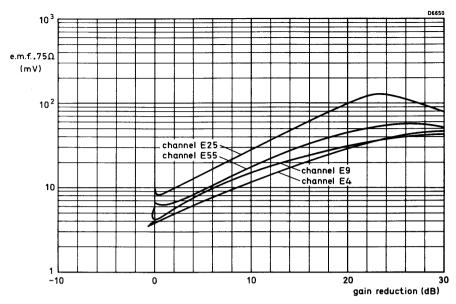


Fig. 12. Cross modulation, in channel.

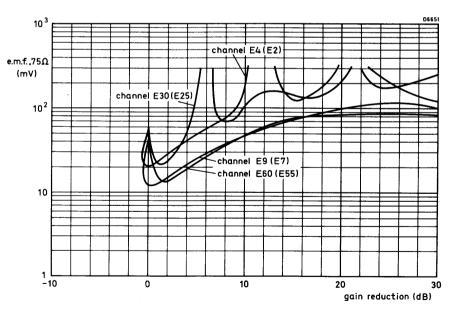


Fig. 13. Cross modulation in band; the interfering channels are given between brackets.

#### APPLICATION

#### Connection of the tuner

For connection of the tuner, see terminal location in Fig. 1. If the tuner is used in receivers, the chassis of which is connected to the mains, isolating capacitors according to the safety rules have to be inserted in the aerial leads.

Five ways of connecting, depending on the number of switches available, are given below.

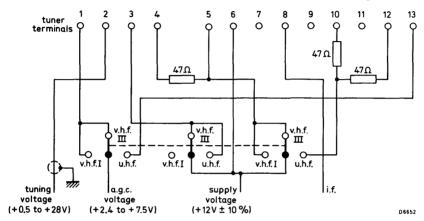


Fig. 14. Connection diagram with three switches.

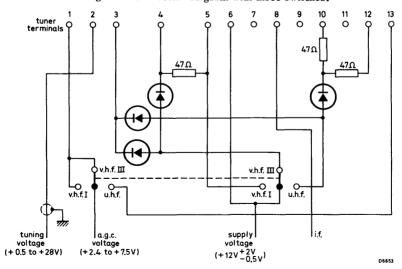


Fig. 15. Connection diagram with two switches.
All diodes: BAX13, BA217 or comparable silicon diodes.

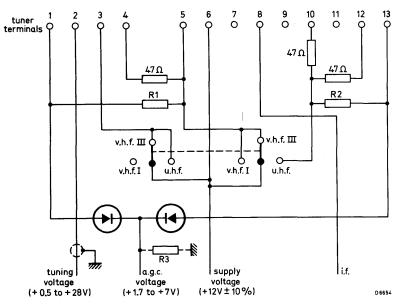


Fig. 16. Connection diagram with two switches. All diodes: BAX13, BA217 or comparable silicon diodes. The values of  $R_1$ ,  $R_2$  and  $R_3$  are depending on a.g.c. circuit.

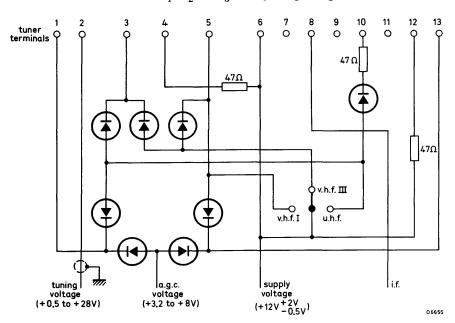


Fig. 17. Connection diagram with one switch.

All diodes: BAX13, BA217 or comparable silicon diodes.

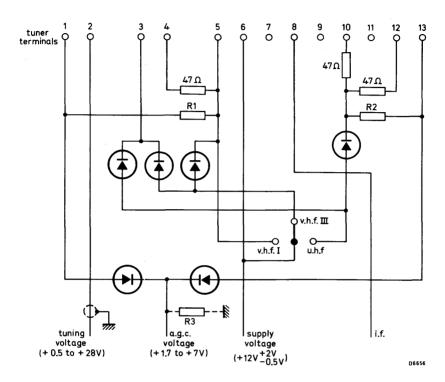


Fig. 18. Connection diagram with one switch.

All diodes: BAX13, BA217 or comparable silicon diodes.

The values of R<sub>1</sub>, R<sub>2</sub> and R<sub>3</sub> are depending on a.g.c. circuit.

# Alignment of i.f. circuit

The tuner is provided with an i.f. injection point at the collector of the mixer, for aligning the i.f. circuit together with the i.f. amplifier of the television receiver (for the position of the i.f. injection point see Fig. 1).

The aligning should be done with the v.h.f. III band tuned. The tuning voltage should be 15 to 20 V.

If this injection method cannot be employed in the television receiver, e.g. because the injection point is not accessible or there is not enough i.f. signal available, the i.f. signal can be fed to test point 3 (terminal 11) via a capacitor of 0.82 to 1 pF.

The tuner must be switched to the u.h.f. position; the tuning voltage should be approx. 10 V. This injection method requires approx. 14 dB less signal than the first method.

No permanent connection must be made to test point 3, otherwise the tuner may exceed the oscillator radiation limits.

#### MEASURING METHOD OF POWER GAIN

The i.f. output of the tuner should be terminated with the circuit given below. The terminals 7. 9 and 11 should not be connected.

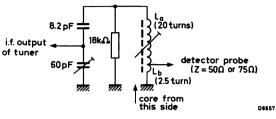


Fig. 19

Switch the tuner to the v.h.f. III band; the tuning voltage should be 15 to 20 V. Feed an i.f. sweep signal (e.m.f. 500 to 1000 mV) to the i.f. injection point. Adjust the trimmer (Fig. 19), tunable coil ( $L_{\rm a}/L_{\rm b}$ ), i.f. output coil of the tuner L19 (Fig. 3) and the coupling between  $L_{\rm a}$  and  $L_{\rm b}$  to get the resonant curve as given below.

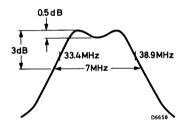


Fig. 20

Then display the r.f. and i.f. curve of the tuner at 190 MHz (picture carrier frequency) and make small corrections in the alignment of the i.f. coils ( $L_{\rm a}/L_{\rm b}$  and L19, if necessary, to get the markers 38.9 MHz symmetrically on the slopes of the curve and the peaks, at equal amplitude.

Because the output impedance of the dummy circuit is 50 to 75  $\Omega$ , the power gain can be measured in the conventional manner by inserting the tuner and the dummy circuit between a 75  $\Omega$  source and a 75  $\Omega$  detector (or between a 50  $\Omega$  source, a matching pad 50/75  $\Omega$ , and a 50  $\Omega$  detector).

# U.H.F. TELEVISION TUNER with diode tuning

#### DEVELOPMENT SAMPLE DATA

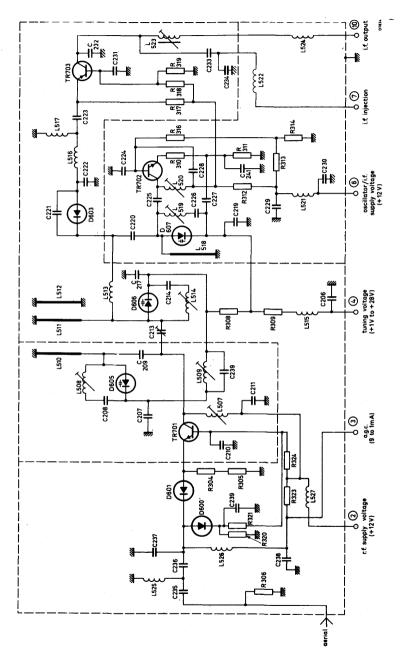
QUICK	REFERENCE DATA
Systems	United Kingdom system
Channels	E21 to E69
	systems I and K
Intermediate frequencies	
picture	39.5 MHz
sound	33.5 MHz

## APPLICATION

This tuner is designed to cover the u.h.f. channels E21 to E69 to meet the special requirements of the United Kingdom.

This information is derived from development samples made available for evaluation, it does not form part of our data handbook system and does not necessarily imply that the device will go into production





11321

# U.H.F. TELEVISION TUNER with diode tuning

#### DESCRIPTION

The U321 is a u.h.f. tuner with electronic tuning, covering the u.h.f. band from 470 to 860 MHz (channels E21 to E69).

Mechanically, the tuner is built on a printed-wiring board, carrying all components, in a metal bousing made of a rectangular frame and front and rear cover (see Fig. 2).

A shielded aerial terminal is on one of the shorter sides of the frame, all the other terminals (supply input stage, a.g.c., tuning voltage, supply oscillator/i.f. stage, i.f. injection and i.f. output) are made via connecting pins (1.32 mm diameter) on the bottom side. Mounting is shown in Fig. 2 and 3.

Electrically the tuner consists of an input circuit with high pass characteristic, followed by a P.I.N. diode attenuator (2 diodes BA379), and a transistor BF480 in grounded base configuration (see Fig. 5). This transistor is operating at an emitter current of about 5 to 8 mA, featuring good noise figures and good handling properties as well. The a.g.c. current for driving the P.I.N. diode attenuator is directly controlled by the a.g.c. system of the receiver. The collector load of the input transistor is formed by a double tuned circuit, transferring the signal to the mixer diode BA280. The selectivity of this circuit at the image frequency has been improved by special means, so that the stringent requirements of the U.K. can be met. The mixer diode is driven by an oscillator, equipped with a transistor BF480. The i.f. signal, originated in the mixer, is amplified by a transistor BF324 in grounded base configuration. The combination of the Schottky-barrier diode BA280 and transistor BF324 features good noise figures and good signal handling properties as well. 3 capacitance diodes BB205B tune the double tuned circuit and the oscillator.

The i.f. output circuit of the tuner is a single tuned one, at the low end of which the i.f. signal is coupled out of the tuner.

A d.c. path to ground for the collector current of the i.f. transistor has to be provided outside the tuner, preferably by a choke of about  $5~\mu H$ . Damping of the i.f. output circuit and matching of the i.f. output to the i.f. circuit of the receiver can be achieved by connecting a series resistor and a parallel capacitor outside the tuner. For details see page 11.

An i.f. injection point is available at the collector of the i.f. transistor, connected to a terminal on the bottom side.

**DEVELOPMENT SAMPLE DATA** 

**Mullard** 

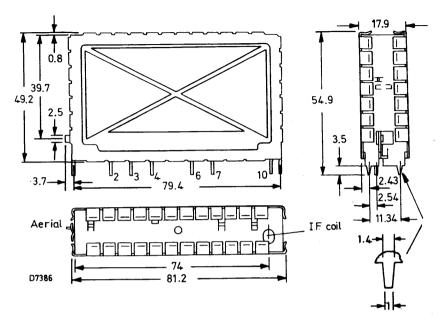


Fig. 2a

```
2 = r.f. supply voltage, +12 V

3 = a.g.c. voltage, +9.2 to +1.5 V

4 = tuning voltage, +1 to +28 V

6 = oscillator/i f. supply voltage, +12 V

7 = i.f. injection point

10 = i.f. output
```



Fig. 2b. I.F. output coil.

Torque for alignment: 2 to 15 mNm

Press-through force : ≥10 N

Mass approx. 75 g

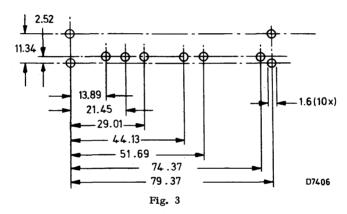
# U.H.F. TELEVISION TUNER with diode tuning

### Mounting

The tuner may be mounted by soldering it on to a printed-wiring board, using the piercing diagram shown in Fig. 3. (The tuner may also be mounted by means of a socket, information will be supplied upon request).

It is recommended that the tuner be installed in the cool part of the receiver cabinet and not exposed to the vibrations of the loudspeaker. There are no restrictions on orientation.

The solderability of the terminals and mounting tabs is according to IEC68-2, test Ta (230  $\pm$  10 °C, 2  $\pm$  0.5 s). The resistance to soldering heat is according to IEC68-2, test Tb (260  $\pm$  5 °C, 10  $\pm$  1 s).



Hole pattern of the printed wiring board (viewed on solder side)

**DEVELOPMENT SAMPLE DATA** 

**Mullard** 

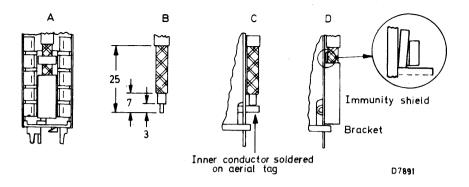


Fig. 4

Fixing of the aerial cable

Recommended cable: DAVU wire CX4004 (outer sheath diameter 5.32 mm)

The aerial cable should be connected as follows:

- strip the cable according to Fig. 4B
- fix the cable as indicated in Fig. 4C and solder the inner conductor on the aerial tag.
- insert lugs on immunity shield under the tabs on tuner body, push the shield into position so that the locating tags snap into place in the tuner body.

# U.H.F. TELEVISION TUNER with diode tuning

#### **ELECTRICAL DATA**

The electrical values are measured on the u.h.f. tuner alone.

Unless otherwise specified all electrical values apply at an ambient temperature of  $25 \pm 5$  °C, a relative humidity of  $60 \pm 15\%$ , a supply voltage of  $12 \pm 0.3$  V and an a.g.c. current of -9 mA  $\pm 0.2$  mA.

Within the given tolerance range of supply voltage and a.g.c. current, only insignificant deviations from the specified values can be expected.

Under the extreme conditions of temperature and humidity as given below, the tuner will function normally, but some specified limits may be exceeded.

#### General

#### Semiconductors

P.I.N. diode	2 × BA 379
r.f. amplifier	BF480
tuning diode	$3 \times BB205B$
mixer diode	BA280
oscillator	BF480
i.f. amplifier	BF 324

#### Environmental conditions

Temperature, operating	+5 to +55 °C
Temperature, storage	-25 to +85 °C
Humidity, operating	max. 90% R. H.

## Voltages and currents

Supply voltage

Voltage

Note:	The supply voltage	e of the input s	tage should l	e filtered to	avoid	hum modulation in
	one of the P. I. N.	diodes when th	he attenuator	is biased to	higher	attenuation ratios.

one of the P.I.N. diodes when the attenuator is biased to higher attenuation ratios.

Under most unfavourable conditions a ripple of ≥ 7.5 mV peak-to-peak may produce a just visible disturbance.

## Current drawn from +12 V supply

Current for r.f. stage terminal 2 at nominal gain at 26 dB gain reduction	typ. 16 mA typ. 13 mA
Current for oscillator and i.f. stage terminal	$6 \leq 16 \mathrm{mA}$
A.G.C. current at nominal gain at 26 dB gain reduction	-9 ± 0.5 mA typ5.6 mA

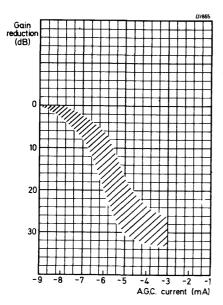
For a.g.c. characteristics see Fig. 7

#### **DEVELOPMENT SAMPLE DATA**



+12 V ± 10% (+10% -15%)\*

<sup>\*</sup>A supply voltage of +12 V -15% is admissible, if a possible deterioration of gain, noise figure, signal handling, oscillator shift and drift is accepted.



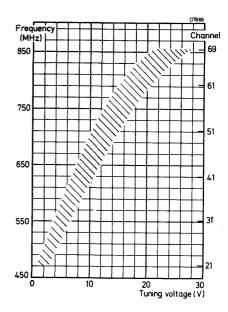


Fig. 4 A.G.C. characteristic

Fig. 5 Tuning characteristic

Tuning voltage range (Fig. 5)

+1 to +28 V

Current drawn from +28 V tuning voltage supply

max.  $0.5 \mu A$ Slope of tuning characteristic min. 5 MHz/V

Note: The source impedance of the tuning voltage offered to terminal 4 must be max. 47 k $\Omega$  at tuning voltages below 3 V.

## Frequencies

Frequency range channel E21 (picture carrier 471.25 MHz) to channel E69 (picture carrier 855. 25 MHz). Margin at the extreme channels: min. 3 MHz

Immediate frequency

39.5 (38.9) MHz picture sound 33.5 (33.4) MHz

> The oscillator frequency is higher than the aerial signal frequency.

Note: The tuner is aligned in such a way that the i.f. of both systems can be applied. The tilt limit is valid for 39.5/33.5 MHz.

# U.H.F. TELEVISION TUNER with diode tuning

## Wanted signal characteristics

Input impedance asymmetrical

**75** Ω

V.S.W.R. at picture carrier frequency, at nominal gain

max. 5

Reflection coefficient at picture carrier frequency, at nominal gain

max. 66%

A.G.C. range

min. 26 dB; typ. 31 dB

R.F. curves

bandwidth

typ. 18 MHz

tilt (only for i.f. 38.9/33.4 MHz)

on any channel the amplitude difference between the top of the r.f. resonant curve and the picture carrier marker, the sound carrier marker, or any frequency between them will not exceed 3 dB at nominal gain, and 4 dB in the a.g.c. range between nominal gain and 20 dB gain reduction.

Power gain (see also Measuring method

of power gain) 18 dB min. channel E21 23 dB typ. channel E40 typ. 22 dB channel E69 24 dB typ.

Gain difference between any two

channels 3dB typ. Noise figure max. 10 dB channel E21 typ. 6.5 dB channel E40 typ. \ 7.5 dB channel E69 typ. 8.0 dB

Overloading

Input signal producing 1 dB gain compression at nominal gain typ. 88 dB ( $\mu$ V) into 75  $\Omega$ 

Input signal producing either a detuning of the oscillator of +300 kHz or -1000 kHz or stopping of the oscillations at nominal gain

typ. 100 dB ( $\mu$ V) into 75  $\Omega$ 

**DEVELOPMENT SAMPLE DATA** 

# Unwanted signal characteristics

Image rejection (measured at picture

carrier frequency) channels E21 to E60

min. 53 dB

I.F. rejection (measured at picture

carrier and colour sub-carrier frequency)

min. 60 dB

 $N \pm 4$  rejection

Interference signal for an interference ratio of 53 dB referred to wanted picture carrier (picture to sound carrier ratio of 7 dB; wanted signal 60 dB ( $\mu$ V); tuner operating at nominal gain)

tvp. 80 dB ( $\mu$ V) into 75  $\Omega$ 

Cross modulation

Input signal producing 1% cross modulation, i.e. 1% of the modulation depth of the interfering signal is transferred to the wanted signal.

In channel cross modulation

(wanted signal: picture carrier frequency; interfering signal: sound carrier frequency)

at nominal gain (wanted input

level 60 dB (µV)

typ. 84 dB ( $\mu$ V)·into 75  $\Omega$ 

at 26 dB gain reduction

(wanted input level 86 dB (µV))

typ. 100 dB ( $\mu$ V) into 75  $\Omega$ 

In band cross modulation

(wanted signal: picture carrier of channel N; interfering signal: picture carrier of channel N  $\pm$  5)

manner iv ± 5)

at nominal gain (wanted input level 60 dB ( $\mu$ V)

at 26 dB gain reduction

(wanted input level 86 dB (µV))

typ. 90 dB ( $\mu$ V) into 75  $\Omega$ 

min. 100 dB ( $\mu$ V) into 75  $\Omega$ 

Out of band cross modulation at nominal gain, v.h.f. I

v.h.f. III

min. 108 dB ( $\mu$ V) into 75  $\Omega$ 

min. 94 dB ( $\mu$ V) into 75  $\Omega$ 

# Oscillator characteristics

Pulling

Input signal of tuned frequency producing a shift of the oscillator frequency of 10 kHz, at nominal gain

10 kmz, at nommar gam

Shift of oscillator frequency at a change of the supply

voltage of 5%

typ. 84 dB ( $\mu$ V) into 75  $\Omega$ 

max. 500 kHz

# U.H.F. TELEVISION TUNER with diode tuning

# Oscillator characteristics (Contd.)

Drift of oscillator frequency during warm-up time (after the tuner has been completely out of operation for 15 min, measured between 5 s and 15 min after switching on)

max. 250 kHz

during warm-up time (after the input stage is in operation for 15 min, measured between 2 s and 15 min after switching on the oscillator/i.f. stage)

max. 250 kHz

at a change of the ambient temperature from +25 to +40 °C (measured after 3 cycles from +25 to +55 °C)

max. 1000 kHz

### I.F. circuit characteristics

Bandwidth of i.f. output circuit 1)

 $5 \text{ MHz} \pm 500 \text{ kHz}$ 

Bandwidth variation of i.f. output circuit as a result of r.f. tuning

max. 350 kHz

Note: I.F. output of the tuner terminated with a modified circuit of Fig. 6, i.e. a 100 pF capacitor is connected in parallel with C1, R1 is short circuited, tuning voltage is 15 V.

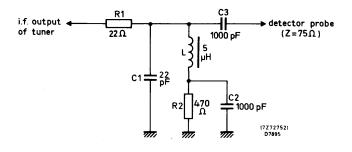


Fig. 6

<sup>1)</sup> I.F. output of the tuner terminated with the circuit shown in Fig. 6; tuning voltage 15 V.

DEVELOPMENT SAMPLE DATA

# I.F. circuit characteristics (Contd.)

Detuning of the i.f. output circuit as a result of r.f. tuning

max, 350 kHz

Note: I.F. output of the tuner terminated with a modified circuit of Fig. 6, i.e. a 100 pF capacitor is connected in parallel with C1, R1 is short circuited; tuning voltage is 15 V.

Tuning range of i.f. output coil 1)

max. 33 to min. 40 MHz

Attenuation between i.f. injection point and i.f. output of the tuner

typ. 23 dB

#### Miscellaneous

Radio interference

Oscillator radiation and oscillator voltage at the aerial terminal

Within the limits of C. I. S. P. R. 24/3 (1970) and VDE 0872/7.72. For the oscillator radiation use is made of the relaxed limit of 3 mV/m (70 dBµV/m).

Microphonics

There will be no microphonics, provided the tuner is installed in a professional manner.

Surge protection

Protection against voltages

max. 8 kV

Note: Three discharges of a 470 pF capacitor into the aerial terminal.

Protection against flashes

max. 30 kV, 400 mWs

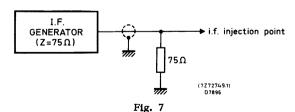
Note: A flashover circuit producing flashes with frequencies of 1 to 20 Hz for 30 s is connected to the aerial terminal.

#### ADDITIONAL INFORMATION

#### I.F. injection

The tuner is provided with an i.f. injection point at the collector of the i.f. transistor (coupled via a capacitor to terminal 7). The i.f. generator can be connected directly to this point (Fig. 7).

The tuner needs normal supply voltages and a tuning voltage of 15 V; the i.f. output should be loaded with the circuit shown in Fig. 6.



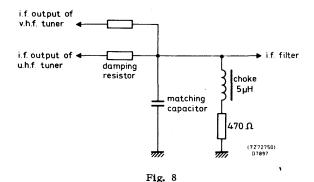
<sup>1)</sup> I.F. output of the tuner terminated with the circuit shown in Fig. 6; tuning voltage 15 V.

# U.H.F. TELEVISION TUNER with diode tuning

# Connection of the i.f. amplifier

The tuner needs a d.c. path from the i.f. output terminal (10) to earth, preferably via a choke of approx.  $5 \mu H$  outside the tuner (Fig. 8).

In the case where the tuner is used in combination with a v.h.f. tuner, this choke can be common for both tuners; a resistor in series with the choke can make ineffective the i.f. output circuit of the switched off tuner. For damping the i.f. output circuit and matching the i.f. output impedance of the tuner to the i.f. amplifier, a series resistor and a parallel capacitor as shown in Fig. 8 should be used.



# Measuring method of power gain

The i.f. output of the tuner should be terminated with the RC-circuit given in Fig. 6.

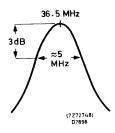


Fig. 9

The RC-circuit roughly matches the i.f. output impedance to 75  $\Omega$  at the resonant frequency of the i.f. output circuit, which should be tuned to 36.15 MHz; the bandwidth should be approx. 5 MHz (Fig. 9).

Because the input and output impedances of the tuner are now 75  $\Omega$ , the power gain can be measured in the conventional manner by inserting tuner and RC-circuit between a 75  $\Omega$  source and a 75  $\Omega$  detector.

**DEVELOPMENT SAMPLE DATA** 

# Alignment of the i.f. output coil

The i.f. output coil should be adjusted with a brass tool with a blade as shown in Fig. 10. A suitable tool is available under catalogue number 7122 009 47680.

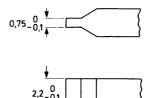


Fig. 10

(7Z75077) D7899

# **ACCESSORIES**

Immunity shield

3122 121 24910

# DEVELOPMENT SAMPLE DATA

This all electronic tuner is identical to the U321, but has been designed for use in colour or monochrome television receivers fitted with closed loop digital tuning systems requiring an oscillator sample output. The tuner features good noise figure and improved signal handling properties, particularly for cross-modulation, by the use of a large signal r.f. transistor and schottky diode mixer. The tuner is equipped with a P.I.N. diode attenuator in the input circuit which is controlled by the A.G.C. system of the receiver.

QUICK REFERENCE DATA					
Supply voltage		+12	v		
Tuning voltage range		+1 to +28	v		
Current drawn from +28 V tuning vo	oltage supply	<0.5	$\mu \mathbf{A}$		
Frequency range	(	Channels 21 - 69			
V.S.W.R.		<b>≤</b> 5			
Power gain	typ.	23	dB		
Noise factor	typ.	7	dB		
Image rejection		≥53	dB		
A.G.C. control range		≥26	dB		
A.G.C. control current					
maximum gain	typ.	9	mA.		
-26 dB	typ.	5	mA.		

#### **OUTLINE AND DIMENSIONS**

See Page 4

#### ACCESSORIES

Immunity shield

Code No. 3122 121 24911

This information is derived from development samples made available for evaluation, It does not form part of our data handbook system and does not necessarily imply that the device will go into production



## ELECTRICAL DATA

Supply voltage		+12	V
Tuning voltage range		+1 to +28	V
Current drawn from +28 V tuning voltage supply		<0.5	μΑ
Frequency range	Channe	ls 21 - 69	
V.s.w.R.		≤5	
Power gain	typ.	23	dB
Noise factor	typ.	7	dB
Image rejection		≥53	dB
A.G.C. control range		≥26	dB
A.G.C. control current maximum gain -26 dB	typ.	9 5	mA mA

## Cross modulation:

Input signal e.m.f., 75  $\Omega$ , which will cause the transference of 1% of the modulation of an unwanted signal to the carrier of the wanted signal.

### In channel cross-modulation:

(wanted signal: vision carrier frequency; interfering signal: sound carrier frequency)

At maximum gain	typ.	25	mV
with a.g.c.	typ.	≥100	mV

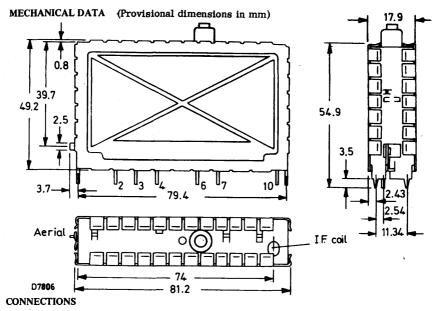
#### In band cross-modulation:

(wanted signal: vision carrier of channel N; interfering signal: vision carrier of channel N  $\pm$  3)

At maximum gain	typ.	50	mV
with a.g.c.	typ.	≥ <b>2</b> 00	mV
Out of band cross-modulation:			
VHF I	typ.	>500	mV
VHF III	typ.	>100	mV
n $\pm$ 4: Interfering vision carrier, e.m.f., 75 $\Omega$ , for -53 dB: (at maximum gain)	typ.	20	mV
Oscillator stability:			
Shift for a change in supply voltage of $\pm 5\%$		≤500	kHz
Drift for a change of ambient temperature from 25 °C to 50 °C		≤1000	kHz

# U.H.F. TELEVISION TUNER For use with digital tuning systems

ELECTRICAL DATA (Oscillator sample)				
Oscillator sample voltage, p.d., 75 $\Omega$	Min.	Typ.	Max.	
$V_{\text{supply}}$ = +10. 8 to +13. 2 V, $V_{\text{tuning}}$ +0. 5 to +28 V, $T_{\text{amb}}$ = +5 to +55 °C	13	-	100	mV
$V_{supply} = 12 \text{ V}, T_{amb} = 25 ^{\circ}\text{C}$	-	33	-	mV
Impedance of oscillator sample port	-	75	-	Ω
V.S.W.R. at oscillator sample port				
Oscillator frequency <600 MHz	-	3	4	
Oscillator frequency >600 MHz	-	2	3	
Reflection coefficient at oscillator sample port				
Oscillator frequency <600 MHz	-	50	60	%
>600 MHz	-	33	50	%
Harmonic content of oscillator sample				
Harmonics below 1000 MHz (2nd harmonic of fundamentals ≤ 500 MHz) with reference to fundamental	-15	-20	_	dB
R.F. rejection at oscillator sample port				
Tuner input of wanted frequency, 5 mV e.m.f., 75 $\Omega$ , tuner operated at nominal gain. $V_{\mbox{signal}}$ at sample port reference oscillator fundamental	-17	-24	- 34	dB
I.F. rejection at oscillator sample port				
<ul> <li>I.F. signal converted from tuner input of wanted frequency, 5 mV e.m.f., 75 Ω; tuner operated at nominal gain.</li> <li>I.F. signal at sample port, reference oscillator fundamental</li> </ul>	-20	- 35	_	dB
Radio interference	20	<b>5</b> 5		
Oscillator port open or terminated 75 $\Omega$ oscillator radiation	-	-	3	mV/m

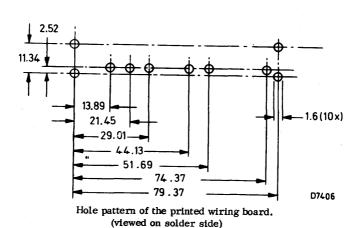


- 2. V supply
- +12 V

- 6. V supply oscillator/i.f.
- +12 V

- 3. A.G.C.
- 9 mA to 1 mA
- 7. I.F. injection.

- 4. tuning
- +1 V to +28 V
- 10. I.F. output



U322 U322LO

This information is derived from development samples made available for evaluation. It does not form part of our data handbook system and does not necessarily imply that the device will go into production

# U.H.F. TELEVISION TUNERS

#### **QUICK REFERENCE DATA**

Systems	C.C.I.R. systems G, H, I and K	
Channels	E21 to E69	
Intermediate frequencies picture	systems G and H	systems I and K
	38,9 MHz	39,5 MHz
sound	33,4 MHz	33,5 MHz

#### **APPLICATION**

These tuners are designed to cover the u.h.f. channels E21 to E69 of C.C.I.R. systems G, H, I and K. In combination with a suitable v.h.f. tuner, e.g. V311, V314 or V315 they can be used in v.h.f./u.h.f. receivers. The aerial inputs and i.f. outputs of both tuners can then be connected in parallel without additional circuitry.

The U322LO is a special version of the U322: an output voltage from the local oscillator is made available for driving digital tuning systems. Apart from this the tuners are identical.



3122 128 54252 3122 128 64851

#### DESCRIPTION

The tuners are u.h.f. tuners with electronic tuning, covering the u.h.f. band from 470 to 860 MHz.

Mechanically, the tuners are built on a printed-wiring board, carrying all components, in a metal housing made of a rectangular frame and front and rear cover (see Fig. 2a). All connections (aerial, supply voltages, a.g.c. voltage, tuning voltage, i.f. injection, i.f. output) are made via terminals on the underside. The mounting method is shown in Fig. 3. Tuner U322LO has a coaxial socket on the top of the frame for coupling out the oscillator sample.

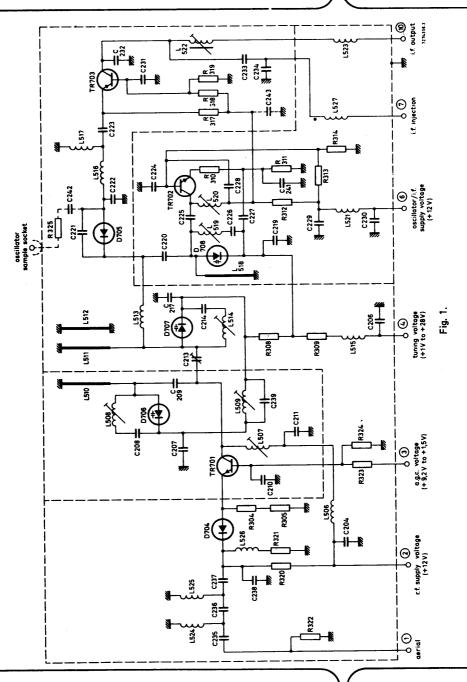
Electrically, the tuner consists of an input circuit with a high-pass characteristic, followed by a P-I-N diode attenuator (1 diode BA379) and the input transistor BF480 in grounded-base configuration. This transistor operates at an emitter current of about 8 to 10 mA, featuring good noise figures and good signal handling properties. It also supplies the current drive for the P-I-N diode attenuator, controlled by an a.g.c. voltage fed to the transistor's base. This combination has good signal handling properties throughout the a.g.c. range. The collector load of the input transistor is formed by a double tuned circuit, transferring the signal to the mixer diode BA280 (or MBD102). The selectivity of this circuit at the image frequency has been improved by special means. The mixer diode BA280 (or MBD102) is driven by an oscillator, equipped with a transistor BF480. At the U322LO the oscillator sample is coupled out of the mixer via a small capacitor in series with a resistor.

The i.f. signal, originated in the mixer, is amplified by a transistor BF324 in grounded-base configuration. The combination of the Schottky-barrier diode BA280 (or MBD102) and the i.f. transistor BF324 also features good noise figures and good signal handling properties. Three capacitance diodes BB105B tune the double tuned circuit and the oscillator.

The i.f. output circuit of the tuner is a single tuned one, at the low end of which the i.f. signal is coupled out of the tuner. A d.c. path to earth for the collector current of the i.f. transistor BF324 has to be provided outside the tuner, preferably by a choke of about 5  $\mu$ H. Damping of the i.f. output circuit and matching of the i.f. output to the i.f. circuit of the receiver can be achieved by connecting a series resistance and a parallel capacitance outside the tuner.

An i.f. injection point has been provided at the collector of the i.f. transistor, connected to terminal 7.







#### **MECHANICAL DATA**

Dimensions in mm

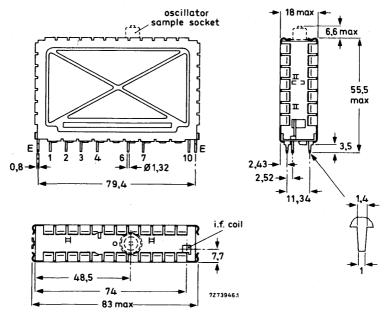


Fig. 2a The oscillator sample socket, drawn with dotted lines, applies only to tuner U322LO.

Terminal 1 = aerial
2 = r.f. supply voltage, + 12 V
3 = a.g.c. voltage, + 9,2 to + 1,5 V
4 = tuning voltage, + 1 to + 28 V
6 = oscillator/i.f. supply voltage, + 12 V
7 = i.f. injection point
10 = i.f. output

Note: When the tuner is operated together with a v.h.f. tuner, only the supply voltage at terminal 6 should be switched off during v.h.f. operation.



Fig. 2b I.F. output coil.

Torque for alignment : 2 to 15 mNm Press-through force : ≥ 10 N

Mass

approx. 75 g

#### Mounting

The tuner may be mounted by soldering it on to a printed-wiring board, using the piercing diagram shown in Fig. 3. (The tuner may also be mounted by means of a socket. Information will be supplied upon request.)

It is recommended that the tuner be installed in the cool part of the receiver cabinet and not exposed to the vibrations of the loudspeaker. There are no restrictions on orientation.

The solderability of the terminals and mounting tabs is according to IEC 68-2, test Ta (230  $\pm$  10 °C, 2  $\pm$  0,5 s). The resistance to soldering heat is according to IEC 68-2, test Tb (260  $\pm$  5 °C, 10  $\pm$  1 s).

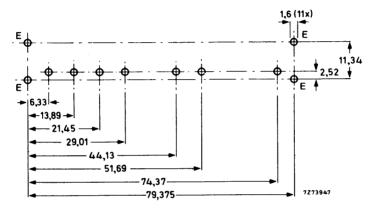


Fig. 3 Piercing diagram viewed from solder side of board.

For connection to the socket on the top of tuner U322LO a coaxial plug has to be used; type 3/2-50 (manufacturer: Daut und Rietz) is recommended.



#### **ELECTRICAL DATA**

The electrical values are measured on the u.h.f. tuner alone, but they are also valid for the u.h.f. tuner in combination with a v.h.f. tuner V311, V314 or V315. Unless otherwise specified all electrical values apply at an ambient temperature of  $25 \pm 5$  °C, a relative humidity of  $60 \pm 15\%$ , a supply voltage of  $12 \pm 0.3$  V and an a.g.c. voltage of  $9.2 \pm 0.2$  V.

Within the given tolerance range of supply voltage and a.g.c. voltage only insignificant deviations from the specified values can be expected. Under the extreme conditions of temperature and humidity as given below, the tuner will function normally, but some specified limits may be exceeded.

#### General

_		
		tors

P-I-N diode r.f. amplifier

mixer oscillator

tuning diodes i.f. amplifier

Ambient temperature range

operating storage

Relative humidity

BA379

BF480

BA280 (or MBD102)

BF480 3 x BB105B BF324

+ 5 to + 55 °C -25 to + 85 °C

max. 90%

# Voltages and currents

Supply voltage

+ 12 V ± 10%

Note: The supply voltage at terminal 2 (input stage) should be filtered to avoid hum modulation in the P-I-N diode when the attenuator is biased to higher attenuation ratios.

Current drawn from + 12 V supply

r.f. amplifier, at nominal gain r.f. amplifier, at 30 dB gain reduction oscillator/i.f. amplifier A.G.C. voltage (Fig. 4), at nominal gain

A.G.C. voltage, at 30 dB gain reduction

Note: A.G.C. voltages between 0 and +

typ. 13 mA typ. 4,5 mA

max. 16 mA + 9,2 ± 0,5 V

min. + 1,5 V

Note: A.G.C. voltages between 0 and + 10 V may be applied without risk of damage.

A.G.C. current (Fig. 4)

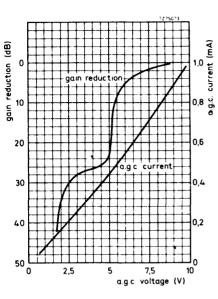
during gain control (0 to 30 dB)

at nominal gain at 30 dB gain reduction

max. +1 mA

typ. + 0.76 to + 0.97 mA

typ. + 0,20 mA



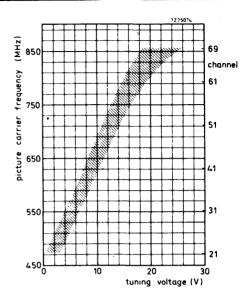


Fig. 4.

Fig. 5.

+ 1 to + 28 V

Tuning voltage range (Fig. 5)

Current drawn from + 28 V tuning voltage supply

max. 0.5 μA Slope of tuning characteristic min. 5 MHz/V

Note: The source impedance of the tuning voltage offered to terminal 4 must be maximum 47 k $\Omega$ at tuning voltages below 3 V.

Oscillator sample signal; only valid for U322LO

at + 12 V supply voltage and

T<sub>amb</sub> = + 25 °C within the given tolerance range of supply voltage and given operating temperature range, and within the

tuning voltage range

+ 0.5 to + 30 V

typ. 90 dB ( $\mu$ V) into 75  $\Omega$ 

82 dB ( $\mu$ V) into 75  $\Omega$ min. max. 100 dB ( $\mu$ V) into 75  $\Omega$ 

Note: A tuning voltage higher than + 28 V will not be harmful for the tuner and may be applied at the user's own risk. Under this condition the published reverse voltage limit of the oscillator tuning diode will be exceeded; the oscillator frequency will never decrease with increasing tuning voltage.



#### Frequencies

Frequency range

channel E21 (picture carrier 471, 25 MHz) to channel E69 (picture carrier 855, 25 MHz). Margin at the extreme channels: min. 3 MHz.

Intermediate frequencies

picture sound

systems G, H systems I, K 38,9 MHz 39.5 MHz 33.4 MHz 33.5 MHz

The oscillator frequency is higher than the aerial signal frequency.

Note. The tuner is aligned in such a way that the i.f. frequencies of the four systems can be applied.

#### Wanted signal characteristics

Input impedance

asymmetrical

**75** Ω

Input impedance of oscillator sample socket; only valid for U322LO

asymmetrical

**75 Ω** 

V.S.W.R. and reflection coefficient

at picture carrier frequency, at

nominal gain

v.s.w.r. reflection coefficient max. 5

max. 66%

V.S.W.R. and reflection coefficient at oscillator sample socket; only valid for U322LO

Mullard

v.s.w.r. at fosc < 600 MHz

v.s.w.r. at fosc > 600 MHz

reflection coefficient at  $f_{\rm OSC}$  < 600 MHz reflection coefficient at fosc > 600 MHz max. 4 (tvp. 3) max. 3 (typ. 2)

max. 60% (typ. 50%) max. 50% (typ. 33%)

R.F. curves, bandwidth

R.F. curves, tilt (only for i.f. 38,9/33,4 MHz)

tvp. 18 MHz

on any channel the amplitude difference between the top of the r.f. resonant curve and the picture carrier marker, the sound carrier marker, or any frequency between them will not exceed 3 dB at nominal gain, and 4 dB in

the a.g.c. range between nominal gain and

20 dB gain reduction.

min, 30 dB A.G.C. range

Power gain (see also Measuring method of power gain)	min. 19 dB
channel E21	typ. 23 dB
channel E40	typ. 22 dB
channel E69	typ. 26 dB
Gain difference between any two channels	max. 4 dB
Noise figure	max. 10 dB
channel E21	typ. 7 dB
channel E40	typ. 7,5 dP
channel E69	typ. 8 dB

Overloading

Input signal producing 1 dB gain compression at nominal gain

Input signal producing either a detuning of the oscillator of + 300 kHz or -1000 kHz or stopping of the oscillations at nominal gain

typ. 100 dB (μV) into 75 Ω

typ. 88 dB ( $\mu$ V) into 75  $\Omega$ 

### **Unwanted signal characteristics**

Image rejection (measured at picture carrier frequency) channels E21 to E60

min. 46 dB; typ. 53 dB

Harmonic content of oscillator sample; only valid for U322LO Suppression of harmonics which fall into the frequency range below 1000 MHz (second harmonics of fundamentals

below 500 MHz)

min. 15 dB (typ. 20 dB) below oscillator fundamental

R.F. rejection at oscillator sample socket; only valid for U322LO Signal voltage at oscillator sample socket (input signals of wanted frequency 70 dB ( $\mu$ V) into 75  $\Omega$ ; tuner operating at nominal gain)

min. 17 dB (typ. 24 to 34 dB) below oscillator fundamental

I.F. rejection (measured at picture carrier and colour sub-carrier frequency)

min. 60 dB

I.F. rejection at oscillator sample socket; only valid for U322LO I.F. signals at oscillator sample socket

I.F. signals at oscillator sample socket (converted from input signals of wanted frequency 70 dB ( $\mu$ V) into 75  $\Omega$ ; tuner operating at nominal gain)

min. 20 dB (typ. 35 dB) below oscillator fundamental



# 3122 128 54252 3122 128 64851

N ± 4 rejection

Interference signal for an interference ratio of 53 dB referred to wanted picture carrier (picture to sound carrier ratio of 10 dB; wanted signal 60 dB (µV); tuner operating at nominal gain)

max. 80 dB ( $\mu$ V) into 75  $\Omega$ 

Cross modulation

Input signal producing 1% cross modulation, i.e. 1% of the modulation depth of the interfering signal is transferred to the wanted signal.

In channel cross modulation (wanted signal: picture carrier frequency; interfering signal: sound carrier frequency)

at nominal gain (wanted input

level 60 dB (µV))

typ. 84 dB ( $\mu$ V) into 75  $\Omega$ 

at 26 dB gain reduction (wanted input level 86 dB (µV)

typ. 100 dB ( $\mu$ V) into 75  $\Omega$ 

In band cross modulation (wanted signal: picture carrier of channel N; interfering signal: picture carrier of channel N ± 5)

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at nominal gain (wanted input

level 60 dB (µV))

typ. 88 dB ( $\mu$ V) into 75  $\Omega$ 

at 26 dB gain reduction

(wanted input level 86 dB (µV))

typ. 100 dB ( $\mu$ V) into 75  $\Omega$ 

Out of band cross modulation, at nominal gain

min. 108 dB ( $\mu$ V) into 75  $\Omega$ 

v.h.f. I

v.h.f. 111

min. 94 dB ( $\mu$ V) into 75  $\Omega$ 

#### Oscillator characteristics

**Pulling** 

Input signal of tuned frequency producing

a shift of the oscillator frequency of 10 kHz, at nominal gain

tvp. 84 dB ( $\mu$ V) into 75  $\Omega$ 

Shift of oscillator frequency

at a change of the supply voltage of 5%

max.500 kHz

Drift of oscillator frequency during warm-up time (after the tuner has been completely out of operation for 15 min, measured between 5 s and 15 min after switching on)

max. 250 kHz

during warm-up time (after the input stage is in operation for 15 min, measured between 2 s and 15 min after switching.on the oscillator/i.f. stage)

max. 250 kHz

at a change of the ambient temperature from + 25 to + 40 °C (measured after 3 cycles from + 25 to + 55 °C) 470 to 790 MHz 790 to 860 MHz

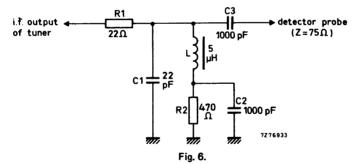
max. 500 kHz max. 650 kHz

#### I.F. circuit characteristics

Bandwidth of i.f. output circuit

5 MHz ± 500 kHz

Note: I.F. output of the tuner terminated with the circuit shown in Fig. 6; tuning voltage 15 V.



Bandwidth variation of i.f. output circuit as a result of r.f. tuning

max, 350 kHz

Note: I.F. output of the tuner terminated with a modified circuit of Fig. 6, i.e. a 100 pF capacitor is connected in parallel with C1 and R1 is short-circuited; tuning voltage 15 V.

Detuning of the i.f. output circuit as a result of r.f. tuning

max, 350 kHz

Note: I.F. output of the tuner terminated with a modified circuit of Fig. 6, i.e. a 100 pF capacitor is connected in parallel with C1 and R1 is short-circuited; tuning voltage 15 V.



# 3122 128 54252 3122 128 64851

Minimum tuning range of i.f. output coil

33 to 40 MHz

Note: I.F. output of the tuner terminated with the circuit shown in Fig. 6; tuning voltage 15 V.

Attenuation between i.f. injection point

and i.f. output of the tuner

typ. 23 dB

Miscellaneous

Radio interference

Oscillator radiation and oscillator

voltage at the aerial terminal

Within the limits of C.I.S.P.R. 13

(1975) and VDE 0872/7.72.\*
For the oscillator radiation use

is made of the relaxed limit of

3 mV/m (70 dBμV/m).

Microphonics

There will be no microphonics,

provided the tuner is installed in a professional manner.

Surge protection

Protection against voltages

max. 8 kV

Note: Three discharges of a 470 pF capacitor into the aerial terminal.

Protection against flashes

max. 30 kV, 400 mWs

Note: A flashover circuit producing flashes with frequencies of 1 to 20 Hz for 30 s is connected to the aerial terminal.

<sup>\*</sup> For U322LO: when the oscillator sample socket is either open or terminated with a shielded resistor of 75  $\Omega$ .

#### ADDITIONAL INFORMATION

#### I.F. injection

The tuner is provided with an i.f. injection point at the collector of the i.f. transistor (coupled via a capacitor to terminal 7). The i.f. generator can be connected directly to this point (Fig. 7). The tuner needs normal supply voltages and a tuning voltage of 15 V; the i.f. output should be loaded with the circuit shown in Fig. 6.

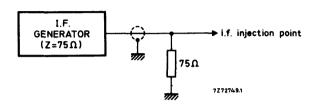
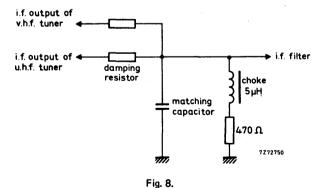


Fig. 7.

# Connection of the i.f. amplifier

The tuner needs a d.c. path from the i.f. output terminal (10) to earth, preferably via a choke of approx.  $5 \mu H$  outside the tuner (Fig. 8). Where the tuner is used in combination with a v.h.f. tuner, this choke can be common for both tuners; a resistor in series with the choke can make ineffective the i.f. output circuit of the switched-off tuner. For damping the i.f. output circuit and matching the i.f. output impedance of the tuner to the i.f. amplifier, a series resistor and a parallel capacitor as shown in Fig. 8 should be used.





#### Measuring method of power gain

The i.f. output of the tuner should be terminated with the RC-circuit given in Fig. 6.

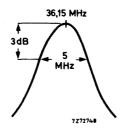


Fig. 9.

The RC-circuit roughly matches the i.f. output impedance to 75  $\Omega$  at the resonant frequency of the i.f. output circuit, which should be tuned to 36,15 MHz; the bandwidth should be approx. 5 MHz (Fig. 9). Because the input and output impedances of the tuner are now 75  $\Omega$ , the power gain can be measured in the conventional manner by inserting tuner and RC-circuit between a 75  $\Omega$  source and a 75  $\Omega$ 

#### Alignment of the i.f. output coil

detector.

The i.f. output coil should be adjusted with a brass tool with a blade as shown in Fig. 10. A suitable tool is available under catalogue number 7122 005 47680.

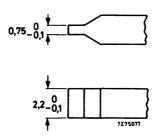


Fig. 10.

#### **ACCESSORIES**

Connector assembly for use of tuner U322 or U322LO in combination with v.h.f. tuner V311 (or VD1), V314 or V315: connector, catalogue number 3112 200 20720; washer, catalogue number 3112 221 01220;

Mullard

clamp, catalogue number 3112 274 13220.

# V.H.F. TELEVISION TUNER with diode tuning

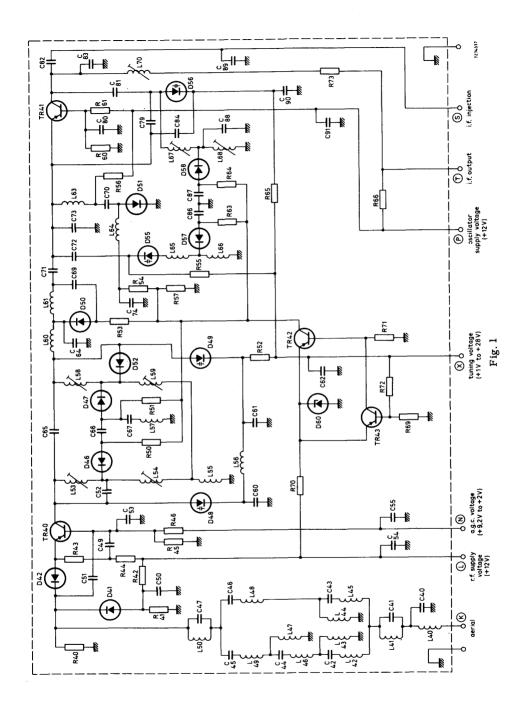
**V311** 

#### DEVELOPMENT SAMPLE DATA

QUICK REFERENCE DATA		
Systems	C.C.I.R. sy	stems B and I
Channels	system B	system I
v.h.f. I	NZ1 to E4	IA to IC
v.h.f. III	E5 to E12	ID to IJ
Intermediate frequencies		
picture	38,9 MHz	39,5 MHz
sound	33,4 MHz	33,5 MHz

#### APPLICATION

This-tuner is designed to cover the v.h.f. channels of C.C.I.R. systems B and I. In combination with the u.h.f. tuner U322 it can be used in v.h.f. /u.h.f. receivers. The aerial inputs and i.f. outputs of both tuners can then be connected in parallel without additional circuitry.



#### DESCRIPTION

The V311 is a v.h.f. tuner with electronic tuning, covering the v.h.f. band I (44 to 68 MHz) and the v.h.f. band III (174 to 230 MHz). Switching between the bands is done automatically by a built-in comparator circuit.

Mechanically, the tuner is built on a printed-wiring board, carrying all components, in a metal housing made of a rectangular frame and front and rear cover (see Fig. 2). All connections (aerial, supply voltage, a.g.c. voltage, tuning voltage, i.f. injection, i.f. output) are made via terminals on the under side. The mounting method is shown in Fig. 3.

Electrically the tuner consists of two input circuits in parallel (bands I and III) with band-pass characteristics, followed by a P-I-N diode attenuator (2 diodes BA379) and the input transistor AF379 in grounded-base configuration. This transistor operates at an emitter current of about 4 to 12 mA, featuring good noise figures and good signal handling properties. It also supplies the current drive for the P-I-N diode attenuator, controlled by an a.g.c. voltage fed to the transistor's base. This combination has good signal handling properties throughout the a.g.c. range.

The collector load of the input transistor is formed by a double tuned circuit, transferring the signal to the self-oscillating mixer AF367. The selectivity of this circuit at the intermediate frequency has been improved.

Four capacitance diodes BB106 tune the double-tuned circuit and the oscillator.

The i.f. output circuit of the tuner is a single tuned one, at the low end of which the i.f. signal is coupled out of the tuner. A d.c. path to earth for the collector current of the mixer has to be provided outside the tuner, preferably by a choke of about 5  $\mu H$ . Damping of the i.f. output circuit and matching of the i.f. output to the i.f. circuit of the receiver can be achieved by connecting a series resistance and a parallel capacitance outside the tuner.

An i.f. injection point has been provided at the collector of the mixer transistor, connected to terminal S.

A comparator circuit supplying the automatic switching-over between bands I and III consists of two p-n-p transistors, the emitters of which have the same stabilized 5,6 V reference voltage, thereby supplying a very good temperature and supply voltage dependence. The voltage divider at the input of the circuit consists of two high-ohmic resistors to prevent unacceptable loading of the tuning voltage.

DEVELOPMENT SAMPLE DATA

**Mullard** 

MECHANICAL DATA Dimensions in mm

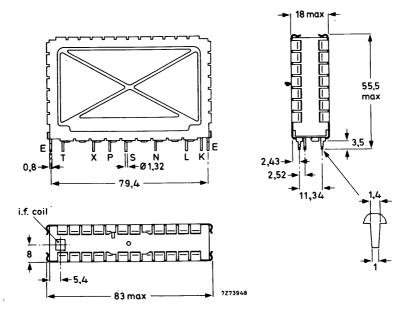


Fig. 2a.

Terminal T = i.f. output

X = tuning voltage, +1 to +28 V

P = self-oscillating mixer supply voltage, +12 V

S = i.f. injection point

N = a.g.c. voltage, +9,2 to +2 V

L = r.f. supply voltage, +12 V

K = aerial

Note: When the tuner is operated together with a u.h.f. tuner, only the supply voltage at terminal P should be switched off during u.h.f. operation.



Fig. 2b. I.F. output coil.

Torque for alignment: 2 to 15 mNm Press-through force : ≥ 10 N

Mass

approx. 80 g

# Mounting

The tuner may be mounted by soldering it on to a printed-wiring board, using the piercing diagram shown in Fig. 3. (The tuner may also be mounted by means of a socket. Information will be supplied upon request.)

It is recommended that the tuner be installed in the cool part of the receiver cabinet and not exposed to the vibrations of the loudspeaker. There are no restrictions on orientation.

The solderability of the terminals and mounting tabs is according to IEC 68-2, test Ta  $(230 \pm 10 \, ^{\circ}\text{C}, 2 \pm 0, 5 \, \text{s})$ . The resistance to soldering heat is according to IEC 68-2, test Tb  $(260 \pm 5 \, ^{\circ}\text{C}, 10 \pm 1 \, \text{s})$ .

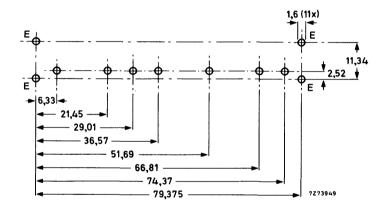


Fig. 3. Piercing diagram viewed from solder side of board.

DEVELOPMENT SAMPLE DATA

Mullard

#### **ELECTRICAL DATA**

The electrical values are measured on the v.h.f. tuner alone, but they are also valid for the v.h.f. tuner in combination with a u.h.f. tuner U322.

Unless otherwise specified all electrical values apply at an ambient temperature of  $25 \pm 5$  °C, a relative humidity of  $60 \pm 15\%$ , a supply voltage of  $12 \pm 0.3$  V and an a.g.c. voltage of  $9,2 \pm 0,2 \text{ V}$ .

Within the given tolerance range of supply voltage and a.g.c. voltage only insignificant deviations from the specified values can be expected.

Under the extreme conditions of temperature and humidity as given below, the tuner will function normally, but some specified limits may be exceeded.

# General

#### Semiconductors

P-I-N diodes	2 x BA 379
r.f. amplifier	AF379
self-oscillating mixer	AF367
tuning diodes	4 x BB106
switching diodes	BA220; 6 x BA243
bandswitch comparator	BZX79; 2 x BC558

Ambient temperature range operating storage	+5 to +55 °C -25 to +85 °C
Relative humidity	max. 90%

#### Voltages and currents

Supply voltage

Note: The supply voltage	at terminal I (innut stage	should be filtered to avoid hum

modulation in one of the P-I-N diodes when the attenuator is biased to higher attenuation ratios.

#### Current drawn from +12 V supply

r.f. amplific	er + bandswitch circuit	
v.h.f. I,	at nominal gain	typ. 40 mA
	at 40 dB gain reduction	typ. 42 mA
v.h.f. III,	at nominal gain	typ. 40 mA
	at 40 dB gain reduction	typ. 42 mA
self-oscillat	ing mixer	typ. 4,5 mA

#### Bandswitching

Switching between v.h.f. I and v.h.f. III is done automatically within the tuner. If the tuner operates together with a u.h.f. tuner only the supply voltage at terminal P should be switched off during u.h.f. operation.



 $+12 \text{ V} \pm 10\%$ 

A.G.C. voltage (Figs. 4, 5 and 6) at nominal gain at 40 dB gain reduction

+9,2 ± 0,5 V min. +2 V

Note: A.G.C. voltages between 0 and +10 V may be applied without risk of damage.

A.G.C. current (Fig. 7), during gain control (0 to 40 dB)

at nominal gain at 40 dB gain reduction max. +1 mA typ. +0,8 mA typ. +0,2 mA

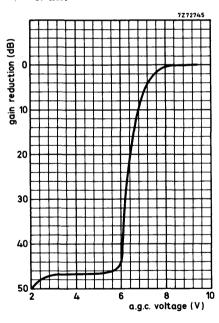
Tuning voltage range (Fig. 8)

+1 to +28 V

Current drawn from +28 V tuning voltage supply (Fig. 9)

-4 to +11 µA

Note: The source impedance of the tuning voltage offered to terminal X must be max. 47 k $\Omega$ .



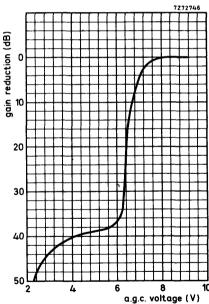


Fig. 4
A.G.C. voltage characteristic, channel E2; typical curve.

Fig. 5
A.G.C. voltage characteristic, channel E5; typical curve.

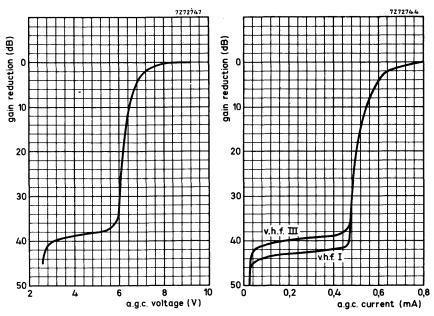


Fig. 6
A.G.C. voltage characteristic, channel E12; typical curve.

Fig. 7
A. G. C. current characteristic; typical curves.

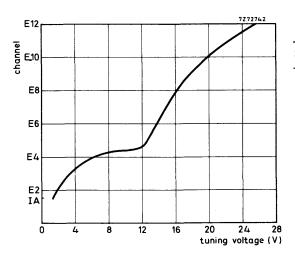


Fig. 8
Tuning voltage characteristic;
typical curve.

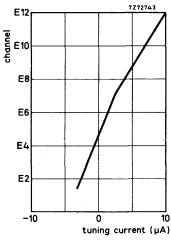


Fig. 9
Tuning current characteristic;
typical curve.

#### Frequencies

Frequency	ranges
-----------	--------

v.h.f. I

channel NZ1 (picture carrier 45,25 MHz) to channel E4 (picture carrier 62,25 MHz). Margin at the extreme channels: min.

1 MHz.

v.h.f. III

channel E5 (picture carrier 175, 25 MHz) to channel E12 (picture carrier 224, 25 MHz). Margin at the extreme channels: min.

1.5 MHz.

Intermediate frequencies

picture sound 
 system B
 system I

 38,9 MHz
 39,5 MHz

 33,4 MHz
 33,5 MHz

The oscillator frequency is higher than the aerial signal frequency.

Note: The tuner is aligned in such a way that the i.f. frequencies of both systems can be applied.

#### Wanted signal characteristics

Input impedance asymmetrical

75 Ω

minimum value between picture carrier and sound carrier frequency

maximum value at picture carrier frequency

V.S.W.R.

Reflection coefficient

A.G.C. range

R.F. curves

bandwidth tilt (only for i.f. 38,9/33,4 MHz)

max. 4 max. 60% max. 4 max. 60%

min. 40 dB

typ. 10 MHz

on any channel the amplitude difference between the top of the r.f. resonant curve and the picture carrier marker, the sound carrier marker, or any frequency between them will not exceed 3 dB at nominal gain, and 4 dB in the a.g.c. range between nominal gain and 20 dB gain reduction.

DEVELOPMENT SAMPLE DATA



Power gain (see also 'Measuring method of power gain') channel E3 channel E5 channel E12	min. 20 dB typ. 25 dB typ. 25 dB typ. 26 dB
Gain difference between any two channels  Noise figure channel E3 channel E5 channel E12	typ. 4 dB max. 9 dB typ. 5 dB typ. 6,5 dB typ. 7 dB
Overloading Input signal producing 1 dB gain compression at nominal gain	typ. 88 dB (μV) into 75 Ω
Input signal producing either a detuning of the oscillator of +300 kHz or -1000 kHz or stopping of the oscillations at nominal gain	typ. 90 dB ( $\mu V$ ) into 75 $\Omega$
Unwanted signal characteristics	
Image rejection (measured at picture carrier frequency)	min. 53 dB
I.F. rejection (measured at picture carrier frequency) channel IA to E12	min. 60 dB
Note: At colour sub-carrier frequency max. 6 dB less rejec	ction.
Cross modulation Input signal producing 1% cross modulation, i.e. 1% of the interfering signal is transferred to the wanted signal.	ne modulation depth of the
In channel cross modulation (wanted signal: picture carrier frequency; interfering signal: sound carrier frequency)	
at nominal gain (wanted input level 60 dB ( $\mu V$ )	typ. 70 dB (μV) into 75 Ω

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at 40 dB gain reduction (wanted input level 100 dB (μV)

In band cross modulation (wanted signal: picture carrier of channel N; interfering signal: picture carrier of channel N ± 2 for v.h.f. I or channel N ± 3 for v.h.f. III at nominal gain (wanted input

at 40 dB gain reduction (wanted input level 100 dB (µV)

level 60 dB (µV)

typ. 106 dB ( $\mu V$ ) into 75  $\Omega$ 

typ. 100 dB ( $\mu V$ ) into 75  $\Omega$ 

typ. 106 dB ( $\mu$ V) into 75  $\Omega$ 

# V.H.F. TELEVISION TUNER with diode tuning

**V311** 

Out of band cross modulation at nominal gain

v.h.f. I, interfering from v.h.f. III interfering from u.h.f.

v.h.f. III, interfering from v.h.f. I interfering from u.h.f.

to be established

#### Oscillator characteristics

#### Pulling

Input signal of tuned frequency producing a shift of the oscillator frequency of 10 kHz, at nominal gain v.h.f. I

v.h.f. III

typ. 73 dB ( $\mu$ V) into 75  $\Omega$  typ. 73 dB ( $\mu$ V) into 75  $\Omega$ 

Shift of oscillator frequency at a change of the supply voltage of 5%

max. 250 kHz

Drift of oscillator frequency during warm-up time (after the tuner has been completely out of operation for 15 min, measured between 5 s

and 15 min after switching on) max. 250 kHz

during warm-up time (after the input stage is in operation for 15 min, measured between 2 s and 15 min after switching on the self-oscillating mixer stage)

max, 250 kHz

at a change of the ambient temperature from +25 to +40 °C (measured after 3 cycles from +25 to +40 °C)

max. 300 kHz

DEVELOPMENT SAMPLE DATA

**Mullard** 

# I.F. circuit characteristics

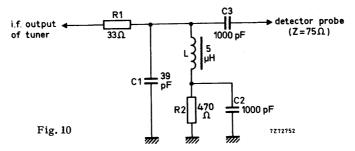
Bandwidth of i.f. output circuit 1)

5 MHz

Bandwidth variation of i.f. output circuit as a result of r.f. tuning and bandswitching (reference: v.h.f. III)

max. 350 kHz

Note: I.F. output of the tuner terminated with a modified circuit of Fig. 10, i.e. a 100 pF capacitor is connected in parallel with Cl and Rl is short circuited; tuning voltage is 15 V.



Detuning of the i.f. output circuit as

a result of r.f. tuning and bandswitching (reference; v.h.f. III), excluded channel E2 channel E2

max. 350 kHz max. 450 kHz

Note: I.F. output of the tuner terminated with a modified circuit of Fig. 10, i.e. a 100 pF capacitor is connected in parallel with C1 and R1 is short circuited; tuning voltage is 15 V.

Tuning range of i.f. output coil 1)

max. 33 to min. 40 MHz

Attenuation between i.f. injection point and i.f. output of the tuner

typ. 23 dB

<sup>1)</sup> I.F. output of the tuner terminated with the circuit shown in Fig. 10; tuning voltage is 15 V.

#### Miscellaneous

Radio interference

Oscillator radiation and oscillator voltage at the aerial terminal

Within the limits of C.I.S.P.R. 24/3 (1970) and VDE 0872/7.72. For the oscillator radiation above 200 MHz use is made of the relaxed limit of 2 mV/m (66 dBuV/m).

Microphonics

There will be no microphonics, provided the tuner is installed in a professional manner.

Surge protection

Protection against voltages

max. 8 kV

Note: Three discharges of a 470 pF capacitor into the aerial terminal.

Protection against flashes

max. 30 kV, 400 mWs

Note: A flash-over circuit producing flashes with frequencies of 1 to 20 Hz for 30 s is connected to the aerial terminal.

#### ADDITIONAL INFORMATION

#### I.F. injection

The tuner is provided with an i.f. injection point at the collector of the mixer transistor (coupled via a capacitor to terminal S). The i.f. generator can be connected directly to this point (Fig. 11).

The tuner needs normal supply voltages and a tuning voltage of 15 V; the i.f. output should be loaded with the circuit shown in Fig. 10.

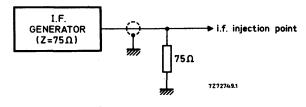


Fig. 11

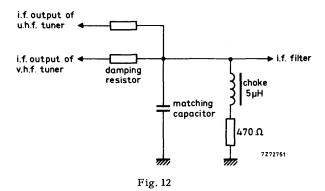
DEVELOPMENT SAMPLE DATA

Mullard

### Connection of the i.f. amplifier

The tuner needs a d.c. path from the i.f. output terminal (T) to earth, preferably via a choke of approx.  $5 \mu H$  outside the tuner (Fig. 12).

In the case where the tuner is used in combination with a u.h.f. tuner, this choke can be common for both tuners; a resistor in series with the choke can make ineffective the i.f. output circuit of the switched off tuner. For damping the i.f. output circuit and matching the i.f. output impedance of the tuner to the i.f. amplifier, a series resistor and a parallel capacitor as shown in Fig. 12 should be used.



# Measuring method of power gain

The i.f. output of the tuner should be terminated with the RC-circuit given in Fig. 10.

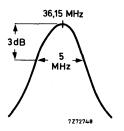


Fig. 13

The RC-circuit roughly matches the i.f. output impedance to 75  $\Omega$  at the resonant frequency of the i.f. output circuit, which should be tuned to 36,15 MHz; the bandwidth should be approx. 5 MHz (Fig. 13).

Because the input and output impedances of the tuner are now 75  $\Omega$ , the power gain can be measured in the conventional manner by inserting tuner and RC-circuit between a 75  $\Omega$  source and a 75  $\Omega$  detector.

# Alignment of the i.f. output coil

The i.f. output coil should be adjusted with a brass tool with a blade as shown in Fig. 14. A suitable tool is available under catalogue number 7122 009 47680.

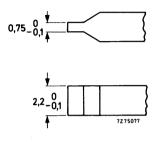


Fig. 14

# **ACCESSORIES**

Connector assembly for use of tuner V311 in combination with u.h.f. tuner U322 (or UD1): connector, catalogue number 3112 200 20720; washer, catalogue number 3112 221 01220; clamp, catalogue number 3112 274 13220.

This information is derived from development samples made available for evaluation. It does not form part of our data handbook system and does not necessarily imply that the device will go into production

# V.H.F TELEVISION TUNER

#### **QUICK REFERENCE DATA**

Systems C.C.I.R. systems B and G

Channels
v.h.f. I E2 to C
v.h.f. III Morocco 4 to E12

Intermediate frequencies
picture 38,9 MHz

33,4 MHz

#### APPLICATION

sound

This tuner is designed to cover the v.h.f. channels of C.C.I.R. systems B and G, including the Italian and Moroccan channels.

In combination with the u.h.f. tuner U322 it can be used in v.h.f./u.h.f. receivers. The aerial inputs and i.f. outputs of both tuners can then be connected in parallel without additional circuitry.

The tuner is compatible with tuner V315. It is also compatible with tuner V311 except for the band switching.



#### DESCRIPTION

The V314 is a v.h.f. tuner with electronic tuning, covering the v.h.f. band I (47 to 88 MHz) and the v.h.f. band III (162 to 230 MHz). Switching between the bands is done by connecting the supply voltage to terminal V for band I and to terminal P for band III.

Mechanically, the tuner is built on a printed-wiring board, carrying all components, in a metal housing made of a rectangular frame and front and rear cover (see Fig. 2a). All connections (aerial, supply voltage, a.g.c. voltage, tuning voltage, i.f. injection, i.f. output) are made via terminals on the underside. The mounting method is shown in Fig. 3.

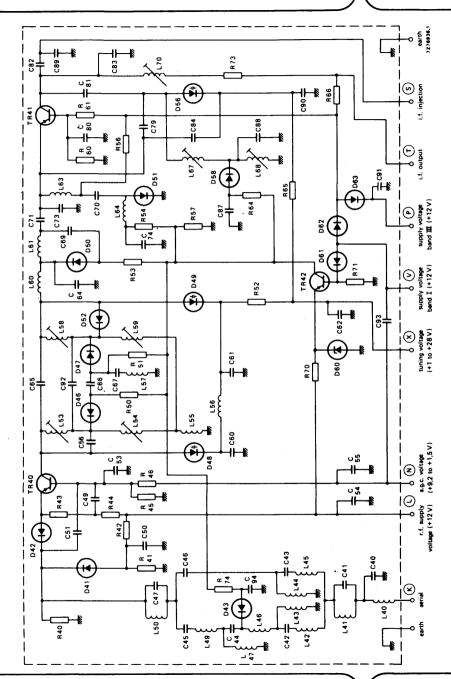
Electrically the tuner consists of two input circuits in parallel (bands I and III) with band-pass characteristics, followed by a P-I-N diode attenuator (2 diodes BA379) and the input transistor AF379 in grounded-base configuration. This transistor operates at an emitter current of about 4 to 12 mA, featuring good noise figures and good signal handling properties. It also supplies the current drive for the P-I-N diode attenuator, controlled by an a.g.c. voltage fed to the transistor's base. This combination has good signal handling properties throughout the a.g.c. range. The collector load of the input transistor is formed by a double tuned circuit, transferring the signal to the self-oscillating mixer AF367. The selectivity of this circuit at the intermediate frequency has been improved. Three capacitance diodes BB106 tune the double tuned circuit and the oscillator.

The i.f. output circuit of the tuner is a single tuned one, at the low end of which the i.f. signal is coupled out of the tuner. A d.c. path to earth for the collector current of the mixer has to be provided outside the tuner, preferably by a choke of about 5  $\mu$ H. Damping of the i.f. output circuit and matching of the i.f. output to the i.f. circuit of the receiver can be achieved by connecting a series resistance and a parallel capacitance outside the tuner.

An i.f. injection point has been provided at the collector of the mixer transistor, connected to terminal S.

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# **MECHANICAL DATA**

Dimensions in mm

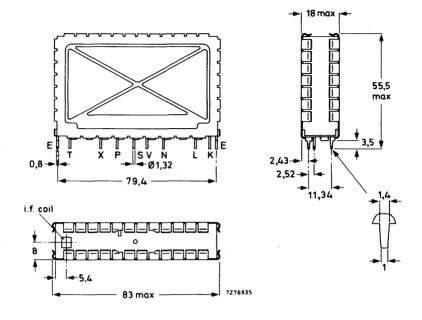


Fig. 2a.

Terminal T = i.f. output X = tuning voltage, + 1 to + 28 V P = supply voltage, band III, + 12 V S = i.f. injection point V = supply voltage, band I, + 12 V N = a.g.c. voltage, + 9,2 to + 1,5 V L = r.f. stage supply voltage, + 12 V K = aerial

Note: When the tuner is operated together with a u.h.f. tuner, only the supply voltage at terminals P and V should be switched off during u.h.f. operation.



Fig. 2b I.F. output coil. Torque for alignment: 2 to 15 mNm Press-through force : ≥ 10 N

Mass

approx. 80 g

#### Mounting

The tuner may be mounted by soldering it on to a printed-wiring board, using the piercing diagram shown in Fig. 3. (The tuner may also be mounted by means of a socket. Information will be supplied upon request.)

It is recommended that the tuner be installed in the cool part of the receiver cabinet and not exposed to the vibrations of the loudspeaker. There are no restrictions on orientation.

The solderability of the terminals and mounting tabs is according to IEC 68-2, test Ta  $(230 \pm 10 \,^{\circ}\text{C}, 2 \pm 0.5 \,\text{s})$ . The resistance to soldering heat is according to IEC 68-2, test Tb  $(260 \pm 5 \,^{\circ}\text{C}, 10 \pm 1 \,\text{s})$ .

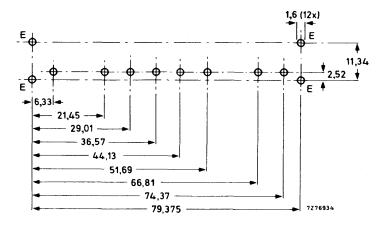


Fig. 3 Piercing diagram viewed from solder side of board.



#### **ELECTRICAL DATA**

The electrical values are measured on the v.h.f. tuner alone, but they are also valid for the v.h.f. tuner in combination with a u.h.f. tuner U322. Unless otherwise specified all electrical values apply at an ambient temperature of 25  $\pm$  5 °C, a relative humidity of 60  $\pm$  15%, a supply voltage of 12  $\pm$  0.3 V and an a.g.c. voltage of 9.2 ± 0.2 V.

Within the given tolerance range of supply voltage and a.g.c. voltage only insignificant deviations from the specified values can be expected. Under the extreme conditions of temperature and humidity as given below, the tuner will function normally, but some specified limits may be exceeded.

#### General

Semiconductors

P-I-N diodes 2 x BA379 r.f. amplifier AF379 self-oscillating mixer AF367 tuning diodes 3 x BB106

switching diodes 4 x BA220; 6 x BA243 switching transistor BC558

BZX79 - C5V6 voltage regulator diode

Ambient temperature range

operating + 5 to + 55 °C -25 to +85 °C storage

Relative humidity Voltages and currents

Supply voltage + 12 V ± 10%

Note: The supply voltage at terminals P and V should be filtered.

Current drawn from + 12 V supply

r.f. amplifier, v.h.f. I, at nominal gain typ. 40 mA v.h.f. I, at 40 dB gain reduction tvp. 42 mA r.f. amplifier, v.h.f. III, at nominal gain tvp. 40 mA v.h.f. III, at 40 dB gain reduction typ. 42 mA self-oscillating mixer typ. 5 mA

Band switching

For operation in band I the supply voltage must be connected to terminal V, for band III operation to terminal P. If the tuner operates together with a u.h.f. tuner only the supply voltage at terminals P and V should be switched off during u.h.f. operation.

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max 90%



A.G.C. voltage (Figs 4 to 7)

+ 9,2 ± 0,5 V

at nominal gain at 40 dB gain reduction

min. + 1,5 V

Note: A.G.C. voltages between 0 and + 10 V may be applied without risk of damage.

A.G.C. current

during gain control (0 to 40 dB)

at nominal gain

at 40 dB gain reduction

Tuning voltage range (Figs 8 and 9)

Current drawn from + 28 V tuning voltage supply

max. + 1 mA

typ. + 0,8 mA

typ. + 0,2 mA

+ 1 to + 28 V

max. 150 nA

Note: The source impedance of the tuning voltage offered to terminal X must be maximum 47 k $\Omega$ .

#### Switching current

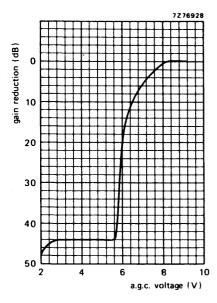


Fig. 4 A.G.C. voltage characteristic, channel E2; typical curve.

# max. 16 mA

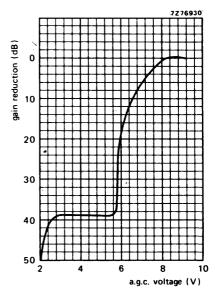


Fig. 5 A.G.C. voltage characteristic, channel C; typical curve.

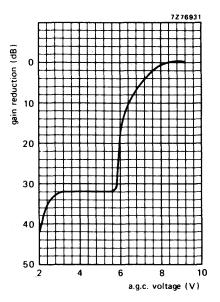


Fig. 6 A.G.C. voltage characteristic, channel M4; typical curve.

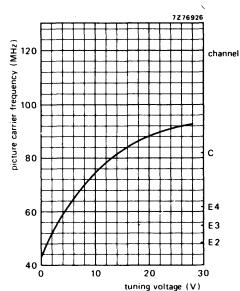


Fig. 8 Tuning voltage characteristic, v.h.f. I; typical curve.

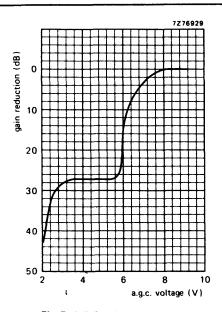


Fig. 7 A.G.C. voltage characteristic, channel E12; typical curve.

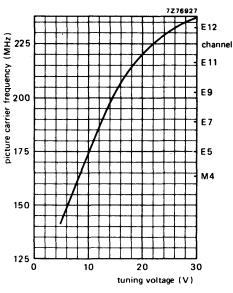


Fig. 9 Tuning voltage characteristic, v.h.f. III; typical curve.

#### Frequencies

Frequency ranges

v.h.f. 1

v.h.f. III

channel E2 (picture carrier 48,25 MHz) to channel C (picture carrier 82,25 MHz)

Margin at the extreme channels: min. 1 MHz. channel M4 (picture carrier 163,25 MHz) to channel E12 (picture carrier 224.25 MHz)

Margin at the extreme channels: min. 1 MHz.

Intermediate frequencies

picture sound

38,9 MHz 33.4 MHz

The oscillator frequency is higher than the aerial signal frequency.

#### Wanted signal characteristics

Input impedance

asymmetrical

**75** Ω

minimum value between picture maximum value at picture carrier

frequency

max. 4

max. 60%

carrier and sound carrier frequency

V.S.W.R.

Reflection coefficient

R.F. curves, bandwidth R.F. curves, tilt

max. 4 max. 60%

typ. 12 MHz on any channel the amplitude difference

between the top of the r.f. resonant curve and the picture carrier marker, the sound carrier marker, or any frequency between them will not exceed 3 dB at nominal gain,

and 4 dB in the a.g.c. range between nominal gain and 20 dB gain reduction. min. 40 dB

A.G.C. range

Power gain (see also Measuring method of power gain) channel E3

channel E5 channel E12 Gain difference between any two channels min, 20 dB

typ. 20 dB typ. 25 dB

typ. 25 dB typ. 4 dB

Noise figure	max. 9 dB
channel E3	typ. 5 dB
channel E5	typ. 6,5 dB
channel E12	typ. 7 dB
Overloading	

Input signal producing 1 dB gain compression at nominal gain 88 dB ( $\mu$ V) into 75  $\Omega$ týp. Input signal producing either a detuning of the oscillator of + 300 kHz

or -1000 kHz or stopping of the oscillations at nominal gain typ. 90 dB ( $\mu$ V) into 75  $\Omega$ 

#### Unwanted signal characteristics

Image rejection (measured at picture

carrier frequency) min. 53 dB

I.F. rejection (measured at picture carrier frequency)

channel E2 to E12 60 dB

Note: At colour sub-carrier frequency maximum 6 dB less rejection.

#### Cross modulation

Input signal producing 1% cross modulation, i.e. 1% of the modulation depth of the interfering signal is transferred to the wanted signal.

In channel cross modulation (wanted signal: picture carrier frequency; interfering signal: sound carrier frequency)

at nominal gain (wanted input

level 60 dB (µV) 70 dB (μV) into 75 Ω tvp.

at 40 dB gain reduction (wanted input level 100 dB (µV) typ. 106 dB ( $\mu$ V) into 75  $\Omega$ 

In band cross modulation (wanted signal: picture carrier of channel N; interfering signal: picture carrier of channel N ± 2 for v.h.f. I or channel N ± 3 for v.h.f. III

at nominal gain (wanted input

level 60 dB (µV) typ. 88 dB ( $\mu$ V) into 75  $\Omega$ 

at 40 dB gain reduction (wanted input level 100 dB (µV) 100 dB ( $\mu$ V) into 75  $\Omega$ typ.

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Out of band cross modulation at nominal gain

v.h.f. I, interfering from v.h.f. III v.h.f. I, interfering from u.h.f.

v.h.f. III, interfering from v.h.f. I

v.h.f. III, interfering from u.h.f.

to be established

#### Oscillator characteristics

Pulling

Input signal of tuned frequency producing a shift of the oscillator frequency of

10 kHz, at nominal gain

V.h.f. I

v.h.f. I typ. 73 dB ( $\mu$ V) into 75  $\Omega$  v.h.f. III typ. 73 dB ( $\mu$ V) into 75  $\Omega$ 

Shift of oscillator frequency

at a change of the supply voltage of 5% max. 250 kHz

Drift of oscillator frequency

during warm-up time (after the tuner has been completely out of operation for 15 min, measured between 5 s and 15 min after switching on)

during warm-up time (after the input stage is in operation for 15 min, measured between 2 s and 15 min after switching on the self-oscillating

at a change of the ambient temperature from +25 to +40 °C (measured after 3 cycles from +25 to +60 °C)

max. 250 kHz

max. 250 kHz

max. 300 kHz

I.F. circuit characteristics

mixer stage)

Bandwidth of i.f. output circuit

6 MHz

Note: I.F. output of the tuner terminated with the circuit shown in Fig. 10; tuning voltage is 15 V; band III at minimum gain.

Bandwidth variation of i.f. output circuit as a result of r.f. tuning and band switching (reference: v.h.f. III)

max. 350 kHz

Note: I.F. output of the tuner terminated with a modified circuit of Fig. 10, i.e. a 100 pF capacitor is connected in parallel with C1 and R1 is short-circuited; tuning voltage is 15 V.

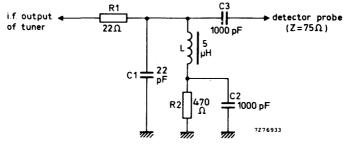


Fig. 10.



Detuning of the i.f. output circuit as a result of r.f. tuning and band switching (reference; v.h.f. III)

max, 350 kHz

Note: I.F. output of the tuner terminated with a modified circuit of Fig. 10, i.e. a 100 pF capacitor is connected in parallel with C1 and R1 is short-circuited; tuning voltage is 15 V.

Tuning range of i.f. output coil

max. 34 to min. 41 MHz

Note: I.F. output of the tuner terminated with the circuit shown in Fig. 10; tuning voltage is 15 V.

Attenuation between i.f. injection point and i.f. output of the tuner

typ. 23 dB

#### Miscellaneous

Radio interference Oscillator radiation and oscillator

voltage at the aerial terminal

Within the limits of C.I.S.P.R. 13 (1975) and VDE 0872/7.72. For the oscillator radiation above

200 MHz use is made of the relaxed limit of 2 mV/m

(66 dBµV/m).

Microphonics

There will be no microphonics, provided the tuner is installed in

a professional manner.

Surge protection

Protection against voltages

max. 8 kV

Note: Three discharges of a 470 pF capacitor into the aerial terminal.

Protection against flashes

max. 30 kV, 400 mWs

Note: A flashover circuit producing flashes with frequencies of 1 to 20 Hz for 30 s is connected to the aerial terminal.

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#### ADDITIONAL INFORMATION

#### I.F. injection

The tuner is provided with an i.f. injection point at the collector of the mixer transistor (coupled via a capacitor and a resistor to terminal S). The i.f. generator can be connected directly to this point (Fig. 11).

The tuner needs normal supply voltages and a tuning voltage of 15 V; the i.f. output should be loaded with the circuit shown in Fig. 10.

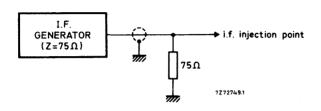


Fig. 11.

#### Connection of the i.f. amplifier

The tuner needs a d.c. path from the i.f. output terminal (T) to earth, preferably via a choke of approx. 5  $\mu$ H outside the tuner (Fig. 12). Where the tuner is used in combination with a u.h.f. tuner, this choke can be common for both tuners; a resistor in series with the choke can make ineffective the i.f. output circuit of the switched off tuner. For damping the i.f. output circuit and matching the i.f. output impedance of the tuner to the i.f. amplifier, a series resistor and a parallel capacitor as shown in Fig. 12 should be used. (During v.h.f. operation the voltage across the 470  $\Omega$  resistor is 1 to 1.2 V.)

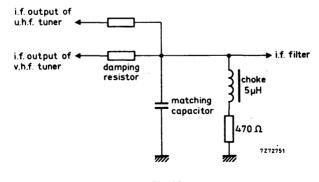


Fig. 12.



#### Measuring method of power gain

The i.f. output of the tuner should be terminated with the RC-circuit given in Fig. 10.

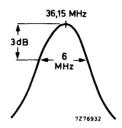


Fig. 13.

The RC-circuit roughly matches the i.f. output impedance to 75  $\Omega$  at the resonant frequency of the i.f. output circuit, which should be tuned to 36,15 MHz; the bandwidth should be approx. 6 MHz (Fig. 13).

Because the input and output impedances of the tuner are now 75  $\Omega$ , the power gain can be measured in the conventional manner by inserting tuner and RC-circuit between a 75  $\Omega$  source and a 75  $\Omega$  detector.

#### Alignment of the i.f. output coil

The i.f. output coil should be adjusted with a brass tool with a blade as shown in Fig. 14. A suitable tool is available under catalogue number 7122 005 47680.

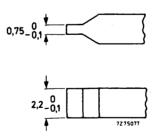


Fig. 14.

#### **ACCESSORIES**

Connector assembly for use of tuner V314 in combination with u.h.f. tuner U322: connector, catalogue number 3112 200 20720; washer, catalogue number 3112 221 01220; clamp, catalogue number 3112 274 13220.

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This information is derived from development samples made available for evaluation. It does not form part of our data handbook system and does not necessarily imply that the device will go into production

### V.H.F. TELEVISION TUNERS

#### QUICK REFERENCE DATA

Systems	C.C.I.R. systems B and G
Channels v.h.f. I v.h.f. III	E2 to S1 S2 to min. S17 (typ. S19)
Intermediate frequencies picture sound	38,9 MHz 33,4 MHz

#### **APPLICATION**

These tuners are designed to cover the v.h.f. channels of C.C.I.R. systems B and G, including the S channels for cable television.

In combination with the u.h.f. tuner U322 they can be used in v.h.f./u.h.f. receivers. The aerial inputs and i.f. outputs of both tuners can then be connected in parallel without additional circuitry.

The tuners are compatible with tuner V314. They are also compatible with tuner V311, except for the band switching.

The V315LO is a special version of the V315: an output voltage from the local oscillator is made available for driving digital tuning systems. Apart from this the tuners are identical.



#### DESCRIPTION

The tuners are v.h.f. tuners with electronic tuning, covering the v.h.f. band I (47 to 111 MHz) and the v.h.f. band III (111 to 279 MHz). Switching between the bands is done by external band switching.

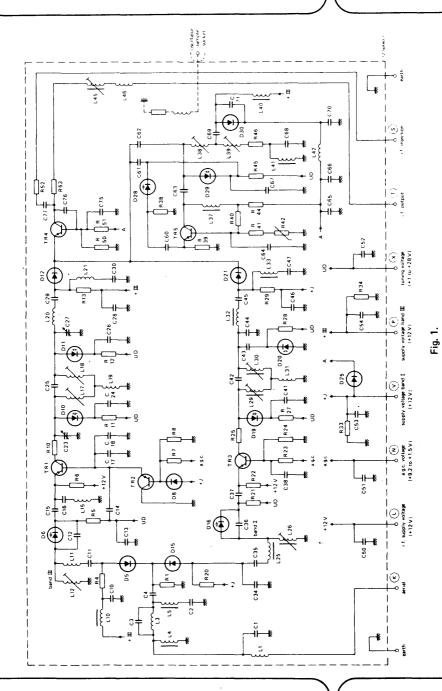
Mechanically, the tuners are built on a printed-wiring board, carrying all components, in a metal housing made of a rectangular frame and front and rear cover (see Fig. 2a). All connections (aerial, supply voltage, a.g.c. voltage, tuning voltage, i.f. injection, i.f. output) are made via terminals on the underside. The mounting method is shown in Fig. 3, Tuner V315LO has a coaxial socket on the top of the frame, for coupling out the oscillator sample.

Electrically the tuner consists of two tunable input circuits in parallel (bands I and III), each followed by an r.f. transistor in grounded-base configuration (BF939 for band I, BF967 for band III). The collector load of each input transistor is formed by a double tuned circuit, transferring the signal to the mixer BF324 fed by the oscillator BF198. Eight capacitance diodes BB209 tune the double-tuned circuit and the oscillator.

The i.f. output circuit of the tuner is a single tuned one, at the low end of which the i.f. signal is coupled out of the tuner. A d.c. path to earth for the collector current of the mixer has to be provided outside the tuner, preferably by a choke of about 5  $\mu$ H. Damping of the i.f. output circuit and matching of the i.f. output to the i.f. circuit of the receiver can be achieved by connecting a series resistance and a parallel capacitance outside the tuner.

An i.f. injection point has been provided at the collector of the mixer transistor, connected to terminal S.







#### **MECHANICAL DATA**

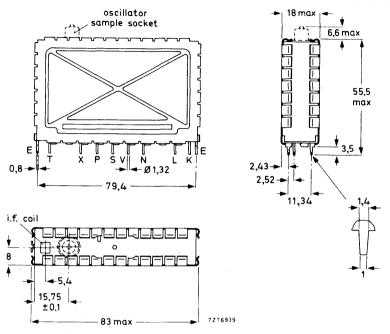


Fig. 2a The oscillator sampling socket, drawn with dotted lines, applies only to tuner V315LO.

Terminal T = i.f. output
X = tuning voltage, + 1 to + 28 V
P = supply voltage, band III, + 12 V
S = i.f. injection point
V = supply voltage, band I, + 12 V
N = a.g.c. voltage, + 9,2 to 1,5 V
L = r.f. stage supply voltage, + 12 V
K = aerial

Note: When the tuner is operated together with a u.h.f. tuner, only the supply voltage at terminals P and V should be switched off during u.h.f. operation.



Fig. 2b I.F. output coil.

Torque for alignment: 2 to 15 mNm

Press-through force: ≥ 10 N

Mass

approx. 80 g



Dimensions in mm

#### Mounting

The tuner may be mounted by soldering it on to a printed-wiring board, using the piercing diagram shown in Fig. 3. (The tuner may also be mounted by means of a socket. Information will be supplied upon request.)

It is recommended that the tuner be installed in the cool part of the receiver cabinet and not exposed to the vibrations of the loudspeaker. There are no restrictions on orientation.

The solderability of the terminals and mounting tabs is according to IEC 68-2, test Ta (230  $\pm$  10  $^{\circ}$  C, 2  $\pm$  0,5 s). The resistance to soldering heat is according to IEC 68-2, test Tb (260  $\pm$  5  $^{\circ}$  C, 10  $\pm$  1 s).

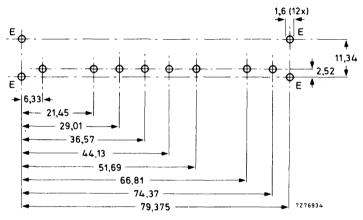


Fig. 3 Piercing diagram viewed from solder side of board.

For connection to the socket on the top of tuner V315LO a coaxial plug has to be used; type 3/2-50 (manufacturer: Daut und Rietz) is recommended.



#### ELECTRICAL DATA

The electrical values are measured on the v.h.f. tuner alone, but they are also valid for the v.h.f. tuner in combination with a u.h.f. tuner U322. Unless otherwise specified all electrical values apply at an ambient temperature of 25 ± 5 °C, a relative humidity of 60 ± 15%, a supply voltage of 12 ± 0,3 V and an a.g.c. voltage of  $9.2 \pm 0.2 \text{ V}$ .

Within the given tolerance range of supply voltage and a.g.c. voltage only insignificant deviations from the specified values can be expected. Under the extreme conditions of temperature and humidity as given below, the tuner will function normally, but some specified limits may be exceeded.

#### General

C	icon	4	

**BF939** r.f. amplifier, band I r.f. amplifier, band III BF967 BF324 mixer

oscillator **BF198** 8 x BB209 tuning diodes

2 x BA220: 2 x BA283 BC558 switching transistor

Ambient temperature range

+ 5 to + 55 °C operating -25 to +85 °C storage max. 90% Relative humidity

Voltages and currents

switching diodes

+ 12 V ± 10% Supply voltage

Note: The supply voltage at terminals P and V should be filtered.

Current drawn from + 12 V supply

typ. 5,8 mA r.f. amplifier, v.h.f. I, at nominal gain typ.12,5 mA v.h.f. I, at 40 dB gain reduction typ. 10 mA r.f. amplifier, v.h.f. III, at nominal gain typ. 20 mA v.h.f. III, at 40 dB gain reduction mixer and oscillator tvp. 12 mA

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#### Band switching

For operation in band I the supply voltage must be connected to terminal V, for band III operation to terminal P. If the tuner operates together with a u.h.f. tuner only the supply voltage at terminals P and V should be switched off during u.h.f. operation.

BA182: 2 x BA244:

A.G.C. voltage (Figs 4 to 7) at nominal gain  $+ 9.2 \pm 0.5 \text{ V}$  at 40 dB gain reduction + 1.5 V

Note: A.G.C. voltages between 0 and + 10 V may be applied without risk of damage.

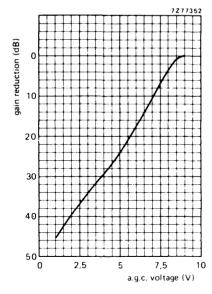
A.G.C. current

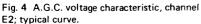
during gain control (0 to 40 dB) max. + 0,5 mA min. - 2 mA

at nominal gain typ. + 0,3 mA

at 40 dB gain reduction typ. -1,2 mA
Tuning voltage range (Figs 8 and 9) + 1 to + 28 V
Current drawn from + 28 V tuning voltage supply max. 400 nA

Note: The source impedance of the tuning voltage offered to terminal X must be max. 47 k $\Omega$ . Switching current max. 16 mA





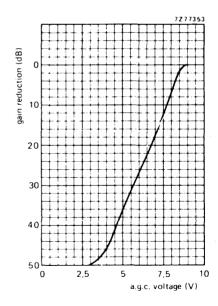


Fig. 5 A.G.C. voltage characteristic, channel S1; typical curve.



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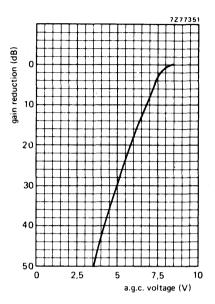


Fig. 6 A.G.C. voltage characteristic, channel S2; typical curve.

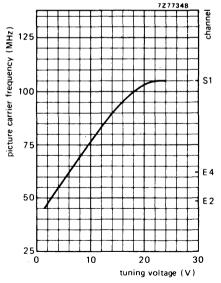
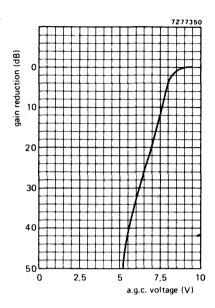


Fig. 8 Tuning voltage characteristic, v.h.f. I; typical curve.



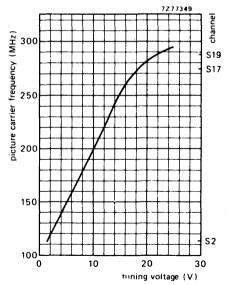


Fig. 9 Tuning voltage characteristic, v.h.f. III; typical curve.

Oscillator sample signal; only valid for V315LO

at + 12 V supply voltage and

Tamb = + 25 °C

within the given tolerance range of supply voltage and given operating-

temperature range, and within the tuning voltage range + 0,5 to + 30 V

min. 80 dB ( $\mu$ V) into 75  $\Omega$  max. 100 dB ( $\mu$ V) into 75  $\Omega$ 

tvp. 84 dB ( $\mu$ V) into 75  $\Omega$ 

Note: A tuning voltage higher than + 28 V will not be harmful for the tuner and may be applied at the user's own risk. Under this condition the published reverse voltage limit of the oscillator tuning diode will be exceeded; the oscillator frequency will never decrease with increasing tuning voltage.

Frequency of oscillator sample signal; only valid for V315LO

v.h.f. I v.h.f. III 87,15 to 144,15 MHz 151,15 to 312,15 MHz (typ. 326,15 MHz)

#### **Frequencies**

Frequency ranges

v.h.f. I

channel E2 (picture carrier 48,25 MHz) to channel S1 (picture carrier 105,25 MHz) Margin at the extreme channels: min. 1 MHz.

v.h.f. III

channel S2 (picture carrier 112,25 MHz) to channel S17 (picture carrier 273,25 MHz) Margin at the extreme channels: min. 1 MHz.

Intermediate frequencies

picture sound 38,9 MHz

33,4 MHz

The oscillator frequency is higher than the aerial signal frequency.

#### Wanted signal characteristics

Input impedance asymmetrical

75  $\Omega$ 

Input impedance of oscillator sample socket; only valid for V315LO asymmetrical 75  $\Omega$ 

V.S.W.R. and reflection coefficient

minimum value between picture carrier and sound maximum value at picture carrier

frequency

carrier frequency

max. 4 max. 60% max. 4 max. 60%

V.S.W.R. and reflection coefficient at oscillator sample socket; only valid for V315LO

v.s.w.r., v.h.f. I

reflection coefficient

v.s.w.r.

v.s.w.r., v.h.f. III reflection coefficient, v.h.f. I max. 2 max. 2 max. 33%

reflection coefficient, v.h.f. III

max. 33%



#### 3122 128 51170 3122 128 51200

R.F. curves, bandwidth

R.F. curves, tilt

typ. 12 MHz

on any channel the amplitude difference

between the top of the r.f. resonant curve and the picture carrier marker, the sound carrier marker, or any frequency between them will not exceed 3 dB at nominal gain, and 4 dB in the a.g.c. range between nominal gain and

20 dB gain reduction.

A.G.C. range, except channel F2

A.G.C. range, channel E2

Power gain (see also Measuring method of power gain) channel E3

channel E5 channel E12

Gain difference between any two channels

min. 40 dB min. 30 dB

min. 20 dB

typ. 23 dB

tvp. 23 dB typ. 23 dB

tvp. 6 dB

#### Unwanted signal characteristics

Image rejection (measured at picture carrier

frequency), channels E2 to E12

min, 60 dB

Harmonic content of oscillator sample; only valid for V315LO

Suppression of harmonics which fall

into the frequency range below 1000 MHz

min. 15 dB below oscillator fundamental

min. 20 dB below oscillator fundamental

R.F. rejection at oscillator sample socket; only valid for V315LO

Signal voltage at oscillator sample socket

(input signals of wanted frequency 70 dB (μV)

into 75  $\Omega$ , tuner operating at nominal gain)

1.F. rejection (measured at picture carrier

frequency), channels E3 to E55 channel E2

min. 60 dB min, 55 dB

Note: At colour sub-carrier frequency max. 6 dB less rejection.

I.F. rejection at oscillator sample socket; only valid for V315LO

I.F. signals at oscillator sample socket

(input signals of wanted frequency 70 dB ( $\mu$ V)

into 75  $\Omega$ , tuner operating at nominal gain)

min. 20 dB below oscillator fundamental

Cross modulation

Input signal producing 1% cross modulation, i.e. 1% of the modulation depth of the interfering signal is transferred to the wanted signal.

In channel cross modulation (wanted signal: picture carrier frequency; interfering signal: sound carrier frequency)

Mullard

at nominal gain (wanted input level 60 dB (µV)

at 40 dB gain reduction (wanted input

level 100 dB (µV)

to be established



In band cross modulation (wanted signal: picture carrier of channel N; interfering signal: picture carrier of channel N  $\pm$  2 for v.h.f. I or channel N  $\pm$  3 for v.h.f. III)

at nominal gain (wanted input level 60 dB ( $\mu$ V)

at 40 dB gain reduction (wanted

input level 100 dB (µV)

to be established

Out of band cross modulation at nominal gain

v.h.f. I, interfering from v.h.f. III

v.h.f. I, interfering from u.h.f. v.h.f. III, interfering from v.h.f. I

v.h.f. III, interfering from u.h.f.

to be established

#### Oscillator characteristics

Pullina

Input signal of tuned frequency producing a shift of the oscillator frequency of 10 kHz, at nominal gain

v.h.f. I v.h.f. III typ. **75** dB ( $\mu$ V) into 75  $\Omega$  typ. **75** dB ( $\mu$ V) into 75  $\Omega$ 

Shift of oscillator frequency

at a change of the supply voltage of 5%

Drift of oscillator frequency

Drift of oscillator frequency during warm-up time (after the tuner has been completely out of operation

for 15 min, measured between 5 s and 15 min after switching on)

max. 250 kHz

max. 400 kHz

during warm-up time (after the input stage is in operation for 15 min, measured between 2 s and 15 min after switching on the self-oscillating

mixer stage)

max. 250 kHz

at a change of the ambient temperature from + 25 to + 40 °C (measured after 3 cycles from + 25 to + 60 °C)

max. 650 kHz

I.F. circuit characteristics

Bandwidth of i.f. output circuit

6 MHz

Note: I.F. output of the tuner terminated with the circuit shown in Fig. 10; tuning voltage is 15 V; band III at minimum gain.

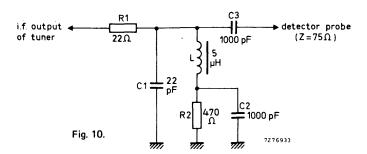
Bandwidth variation of i.f. output circuit as a result of r.f. tuning and

band switching (reference: v.h.f. III)

max. 350 kHz

Note: I.F. output of the tuner terminated with a modified circuit of Fig. 10, i.e. a 100 pF capacitor of connected in parallel with C1 and R1 is short-circuited; tuning voltage is 15 V.





Detuning of the i.f. output circuit as a result of r.f. tuning in band III

max. 350 kHz

Note: I.F. output of the tuner terminated with a modified of Fig. 10, i.e. a 100 pF capacitor is connected in parallel with C1 and R1 is short-circuited; tuning voltage is 15 V.

Tuning range of i.f. output coil

max, 34 to min, 41 MHz

Note: I.F. output of the tuner terminated with the circuit shown in Fig. 10; tuning voltage is 15 V.

Attenuation between i.f. injection point and i.f. output of the tuner

tvp. 23 dB

#### Miscellaneous

Radio interference Oscillator radiation and oscillator voltage at the aerial terminal

Within the limits of C.I.S.P.R. 13 (1975)

and VDE 0872/7.72.\*

For the oscillator radiation above 200 MHz use is made of the relaxed limit of 2 mV/m

 $(66 \text{ dB}\mu\text{V/m}).$ 

Microphonics There will be no microphonics, provided the tuner is installed in a professional manner.

Surge protection

Protection against voltages

max. 8 kV

Note: Three discharges of a 470 pF capacitor into the aerial terminal.

Protection against flashes max. 30 kV, 400 mWs

Note: A flashover circuit producing flashes with frequencies of 1 to 20 Hz for 30 s is connected to the aerial terminal.

• For V315LO: when the oscillator sample socket is either open or terminated with a shielded resistor of 75  $\Omega$ .

#### ADDITIONAL INFORMATION

#### I.F. injection

The tuner is provided with an i.f. injection point at the collector of the mixer transistor (coupled via a capacitor and a resistor to terminal S). The i.f. generator can be connected directly to this point (Fig. 11).

The tuner needs normal supply voltages and a tuning voltage of 15 V; the i.f. output should be loaded with the circuit shown in Fig. 10.

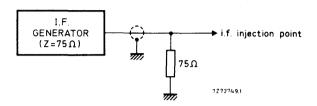


Fig. 11.

#### Connection of the i.f. amplifier

The tuner needs a d.c. path from the i.f. output terminal (T) to earth, preferably via a choke of approx.  $5~\mu\text{H}$  outside the tuner (Fig. 12). Where the tuner is used in combination with a u.h.f. tuner, this choke can be common for both tuners; a resistor in series with the choke can make ineffective the i.f. output circuit of the switched-off tuner. For damping the i.f. output circuit and matching the i.f. output impedance of the tuner to the i.f. amplifier, a series resistor and a parallel capacitor as shown in Fig. 12 should be used (During v.h.f. operation the voltage across the 470  $\Omega$  resistor is 1 to 1,2 V.)

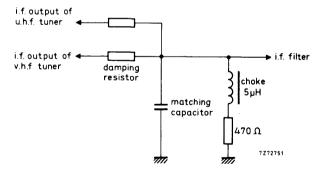
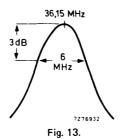


Fig. 12.



#### Measuring method of power gain

The i.f. output of the tuner should be terminated with the RC-circuit given in Fig. 10.

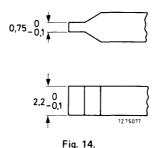


The RC-circuit roughly matches the i.f. output impedance to 75  $\Omega$  at the resonant frequency of the i.f. output circuit, which should be tuned to 36,15 MHz; the bandwidth should be approx. 6 MHz (Fig. 13).

Because the input and output impedances of the tuner are now 75  $\Omega$ , the power gain can be measured in the conventional manner by inserting tuner and RC-circuit between a 75  $\Omega$  source and a 75  $\Omega$  detector.

#### Alignment of the i.f. output coil

The i.f. output coil should be adjusted with a brass tool with a blade as shown in Fig. 14. A suitable tool is available under catalogue number 7122 005 47680.



#### **ACCESSORIES**

Connector assembly for use of tuner V315 or V315LO in combination with u.h.f. tuner U322(or UDI): connector, catalogue number 3112 200 20720;

Mullard

connector, catalogue number 3112 200 20720; washer, catalogue number 3112 221 01220;

clamp, catalogue number 3112 274 13220.

# MONOCHROME TELEVISION ASSEMBLIES





#### OUICK REFERENCE DATA

Designed for use with  $110^{\rm O}$  picture tubes with a neck diameter of 28mm. The unit is suitable for use with line output transformer AT2048/11 for transistor drive, and line linearity control units AT4042/02 or AT4042/14.

Line deflection coils (parallel connected) inductance Field deflection coils (parallel connected) resistance 3.3mH 7.5Ω



#### GENERAL

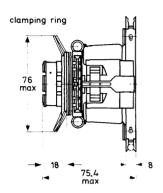
The design of the coil is such as to bring the centre of deflection into the conical part of the picture tube; the coil should therefore be pushed right up the neck of the tube until it touches the cone. Picture shift magnets are mounted on the rear moulding of the coil, and adjustable raster correction magnets are fitted, mounted on 'stalks' for vertical pincushion correction. Facilities are provided on the periphery of the moulding, for mounting small plastic bonded magnets for correcting the corners of the raster.

To orientate the raster correctly, the unit should be rotated by hand on the neck of the picture tube. A screw-tightened clamping ring permits it to be locked in the desired position, both axially and radially.

The assembly is manufactured in flame retardant material to conform to BS415 and IEC 65.

#### DIMENSIONS (millimetres)

#### First angle projection



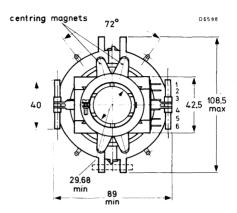


Fig. 1

#### MOUNTING

For optimum raster shape the soldering tag plate must be positioned as shown in Fig. 1.

#### CONNECTIONS

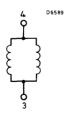


Fig. 2

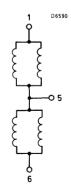


Fig. 3

#### ELECTRICAL DATA (at 25°C)

Line deflection coils, parallel connected (see Fig. 2)		
Connections to pins	3 and 4	
Inductance	$3.3 \pm 5\%$	mΗ
Resistance	6.1 ± 10%	Ω
Deflection current, peak-to-peak (at 18kV and beam deviation of 495mm on a 61 cm (24in) reference picture tube)	2.3 ± 6.5%	A
Field deflection coils, parallel connected (see Fig. 3)		
Connection to pins	1 and 6	
Inductance	17 ± 10%	mH
Resistance	7.5 ± 8%	Ω
Deflection current, peak-to-peak (at 18kV and beam deviation of 390mm on a 61cm (24in) reference picture tube	1.1 ± 5.5%	A
Maximum voltage between line and field coils (50Hz)	2	kV
MAXIMUM OPERATING TEMPERATURE	105	°C
ADJUSTMENT RANGE OF CENTRING MAGNETS	Ø5 to Ø45	mm

#### RASTER DISTORTION

The raster edges fall between the two rectangles, as shown in Fig. 4.

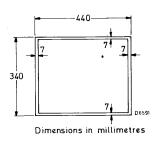


Fig. 4

#### ADJUSTMENTS

#### Correction of eccentricity

After adjustment of the linearity of the deflection current, the eccentricity of the picture tube and the deflection unit can be corrected by means of two independently movable, diametrically magnetised centring magnets. By turning the magnets with respect to each other, the magnetic force of the resultant field of both magnets is adjusted. The direction of the resultant magnetic field is adjusted by turning the magnets simultaneously. It should be noted that the centring magnets cannot be used for compensating the effects of non-linearity or of phase differences between synchronisation and time base, as otherwise the correction needed becomes excessive and, even if the correction is within the range of the magnets, curved lines may appear in the centre of the raster.

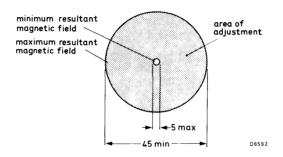


Fig. 5

#### APPLICATION

Deflection coil	Line output transformer	Line linearity control unit	Output device
AT1040/15	AT2048/11	AT4042/02	BU205
		or	
		AT4042/14	

QUICK REFERENCE DATA	
Monitor tube, diagonal neck diameter	24 cm (9 in) 28 mm
Deflection angle	90 °
Line deflection current, edge to edge at 14 kV	8, 6 A (p-p)
Inductance of line coils, parallel connected	93 μΗ
Field deflection current, edge to edge at 14 kV	0, 425 A (p-p)
Resistance of field coils, series connected	27 Ω

#### APPLICATION

This deflection unit has been designed for use with a  $90^{\circ}$  black and white monitor tube type M24-100 W in conjunction with:

line output transformer AT2102/01; linearity control unit AT4036 and;

line output transistor BD160.

#### DESCRIPTION

The saddle-shaped line deflection coils are moulded so that the deflection centre is well within the conical part of the picture tube.

The field deflection coils are wound on a Ferroxcube yoke ring which is flared so that the frame and line deflection centres coincide.

For centring and pin-cushion distortion see under "Correction facilities".

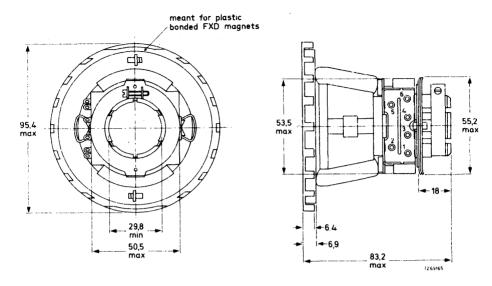


Fig. 1

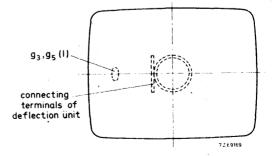
The unit is provided with soldering pins for connection. The pin numbering in Fig. 1 corresponds to that in the connection diagram (Figs. 3 and 4).

#### MOUNTING

The unit should be mounted as far forward as possible on the neck of the picture tube, so that it touches the cone. For optimum raster shape, the coil should be mounted as shown in Fig. 2.

To orientate the raster correctly, the unit may be rotated on the neck of the picture tube. A clamping ring locks the unit both axially and radially.

Fig. 2 Front view of picture tube.



## DEFLECTION UNIT for black and white monitors

## AT1071/01

#### ELECTRICAL DATA

Line deflection coils, parallel connected (Fig. 3) terminals 3 and 4

 $\begin{array}{ccc} \text{Inductance} & & \text{typ. } 93 \, \mu\text{H} \\ \text{Resistance} & & \text{typ. } 0,15 \, \Omega \end{array}$ 

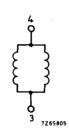
Field deflection coils, series connected (Fig. 4)

terminals 1 and 5 Inductance

Inductance typ. 56 mH Resistance typ. 27  $\Omega$  Maximum peak voltage between terminals of

line and field coils (50 Hz) 2000 V

Maximum operating temperature 95 °C



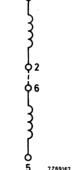


Fig. 4 Field coils

The following characteristics are measured at an e.h.t. of  $14\,kV$  on a  $24\,cm$  (9 in) reference tube, type M24-100 W.

#### Sensitivity

Fig. 3 Line coils

Deflection current edge to edge

 $\begin{array}{ll} \text{in line direction} & 8,6 \quad A \text{ (p-p)} \\ \text{in field direction} & 0,425 \ A \text{ (p-p)} \\ \end{array}$ 

Geometric distortion (measured without correction magnets and centring ring)

Pin-cushion, barrel and trapezium The edges of the raster fall within distortion the two rectangles shown in Fig. 5.

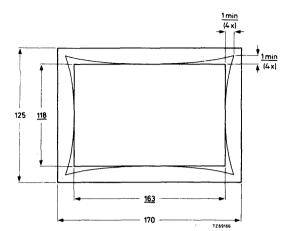


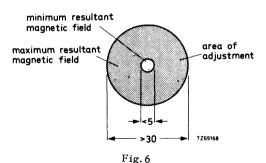
Fig. 5

#### CORRECTION FACILITIES

#### For centring

After adjustment of the linearity of the deflection current, the eccentricity of the picture tube and the deflection unit can be corrected by means of two independently movable centring magnets of plastic-bonded Ferroxdure. These magnets are magnetized diametrically. By turning the magnets with respect to each other the resulting field strength is varied. The direction of the resulting magnetic field is adjusted by turning the magnets simultaneously.

These centring magnets can not be used for compensating the effects of non-linearity or of phase differences between the synchronization and time base, as otherwise the correction needed becomes excessive. Even if the correction is within the range of the magnets, curved lines may appear in the centre of the raster.



#### For pin-cushion distortion

This can be corrected by moving magnets of plastic-bonded Ferroxdure (catalogue number 3122 104 95000) which may be mounted in the rim at the front of the deflection unit.



#### DEFLECTION UNIT

#### QUICK REFERENCE DATA

Picture tube		
diagonal	31 cm (12 in)	24 cm (9 in)
	34 cm (14 in)	31 cm (12 in)
neck diameter	max. 20,9 mm	max. 20,9 mm
Deflection angle	110º	90o
Line deflection current for full scan, at 11 kV	5,02 A (p-p)	4,05 A (p-p)
Inductance of line coils, parallel connected	25	5 μΗ
Field deflection current for full scan, at 11 kV	1,1 A (p-p)	0,91 A (p-p)
Resistance of field coils, parallel connected	2,7	Ω

#### **APPLICATION**

The deflection unit has been designed for use with 31 cm (12 in) or 34 cm (14 in)  $110^{\circ}$  black and white picture tubes, or 24 cm (9 in) or 31 cm (12 in)  $90^{\circ}$  black and white monitor tubes. The unit is used in conjunction with:

- line output transformer AT2140/10 or AT2140:
- linearity control unit AT4042/39;
- line driver transformer AT4043/56.

#### DESCRIPTION

The saddle shaped line deflection coils are moulded so that the deflection centre is well within the conical part of the picture tube. The field deflection coils are wound on a Ferroxdure yoke ring which is flared so that the frame and line deflection centres coincide. Provisions are made for centring, and correction of pin-cushion distortion. The unit meets the self-extinguishing and non-dripping requirements of IEC 65.

#### MOUNTING

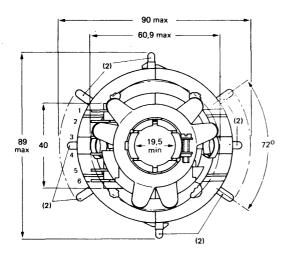
The unit should be mounted as far forward as possible on the neck of the picture tube, so that it touches the cone.

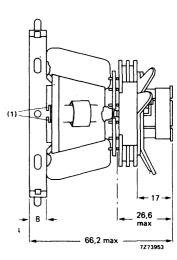
To orient the raster correctly, the unit may be rotated by hand on the neck of the picture tube, with which it makes a slip fit. A screw-tightened clamping ring permits it to be locked, both axially and radially, in the desired position.



#### **MECHANICAL DATA**

Dimensions in mm





 $255 \mu H \pm 5\%$ 

455  $\mu$ H/Ω ± 8%

 $0.56 \Omega$ 

Fig. 1 Deflection unit AT1074. Facilities for fitting correction magnets:

- (1) for bracket with plastic-bonded FXD magnet strip, catalogue number 3122 137 10160;
- (2) for plastic-bonded FXD magnets, catalogue number 3122 104 94120.

The unit is provided with solder pins for connection. The pin numbering in Fig. 1 corresponds to that in the connection diagrams (Fig. 2).

#### **ELECTRICAL DATA**

The electrical values apply at an ambient temperature of 25 °C.

Line deflection coils, parallel connected (Fig. 2a)

terminals 3 and 4

terminals 1 and 6

Inductance

Resistance

L/ 11

Field deflection coils, parallel connected (Fig. 2b)

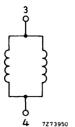
Inductance 7,7 mH ± 8%

Resistance 2,7  $\Omega$ 

L/R 2,87 mH/ $\Omega \pm 10\%$ 

Maximum d.c. voltage between terminals of line and field coils 500 V

Maximum operating temperature 95 °C





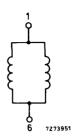


Fig. 2b Field coils.

The following characteristics are measured at an e.h.t. of 11 kV on a 31 cm (12 in) reference picture tube.

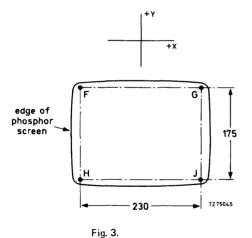
#### Sensitivity

Deflection current edge to edge in line direction

in field direction

110°	
5,02 A (p-p) 1,1 A (p-p)	4,05 A (p-p)
1,1 A (p-p)	0,91 A (p-p)

Geometric distortion measured without correction magnets, on a 31 cm (12 in) reference picture tube.



1 1y. J.

#### CORRECTION FACILITIES

#### For centring

After adjustment of the linearity of the deflection current, the eccentricity of the picture tube and the deflection unit can be corrected by means of two independently movable centring magnets of plasticbonded Ferroxdure. These magnets are magnetized diametrically. By turning the magnets with respect to each other the resulting field strength is varied. The direction of the resulting magnetic field is adjusted by turning the magnets simultaneously.

These centring magnets cannot be used for compensating the effects of non-linearity or of phase differences between the synchronization and time base, as otherwise the correction needed becomes excessive. Even if the correction is within the range of the magnets, curved lines may appear in the centre of the raster.

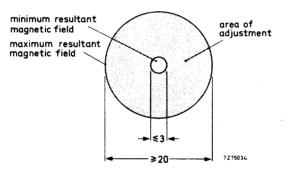


Fig. 4.

#### For geometric distortion

The unit has provisions for mounting brackets for magnet strips\* to correct pin-cushion distortion and for magnets\*\* to correct the raster corners, see Fig. 1.

Plastic-bonded Ferroxdure magnet strips (with bracket) are available on request (catalogue number 3122 137 10160).

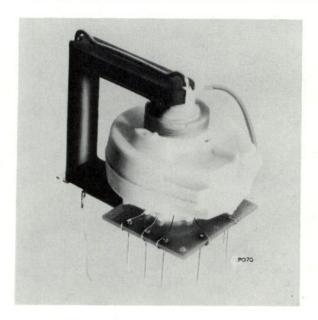
Mullard

\*\* Plastic-bonded Ferroxdure magnets are available on request (catalogue number 3122 104 94120).

#### QUICK REFERENCE DATA

Designed for use in a fully transistorised monochrome television receiver to provide the line scan for the  $110^{0}$  picture tubes with 28mm neck diameter. Intended for use in conjunction with the deflection coil AT1040/15, line linearity control units AT4042/02 or AT4042/14, the line output transistor BU205 and a semiconductor e.h.t. rectifier.

Supply voltage	150	v
Е.Н.Т.	18	kV



#### GENERAL

The primary, secondary and the e.h.t. windings are placed on one limb of the Ferroxcube U and I cores. The e.h.t. coil is encapsulated in a flame retardant polyester.

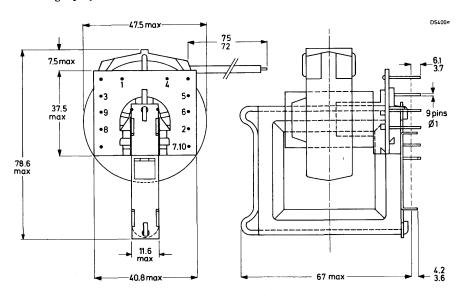


Fig. 1

ELECTRICAL DATA			
Beam current	35	435	- μΑ
Supply voltage	150	150	V
*Supply current	240	290	mA
E.H.T.	17.8	16.2	kV
Internal resistance of e.h.t.	-	4	$M\Omega$
Collector to emitter voltage of BU205, peak-to-peak	960	-	v
Deflection current, peak-to-peak	2.2	-	Α
Overscan	6.5	10	%
Low voltage supply			
pin 5 to earth	31	-	V
pin 6 to earth	12	-	v
pin 8 to 9	6.3	-	v

Note: - The maximum operating temperature of the transformer and core is  $105^{\circ}\text{C}$ . This allows for a maximum operating temperature of  $70^{\circ}\text{C}$ .

\*With 20W of low voltage power

#### MOUNTING

The transformer can be mounted either on a printed-wiring board or a metal chassis. When mounting on a printed-wiring board (Fig. 2), the transformer is secured by its four mounting pins and two screws.

A separation of at least 25mm must be maintained between the transformer and any adjacent metal parts, to avoid reducing the efficiency of the transformer. A free passage of air round the transformer is required to allow sufficient cooling to maintain an operating ambient temperature below  $70^{\circ}\text{C}$ .

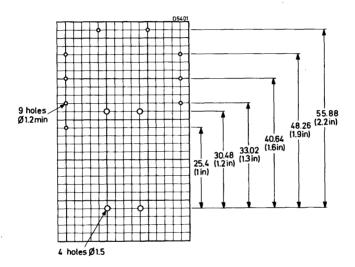
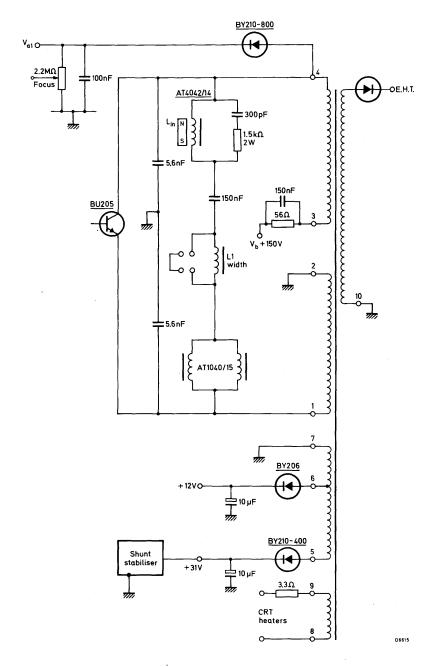


Fig. 2

CIRCUIT DIAGRAM

See overleaf



# LINE OUTPUT TRANSFORMER for black and white monitors

# AT2102/01

QUICK REFE	RENCE DATA		
I <sub>eht</sub>	0,03	0, 23	mA
E.H.T.	14, 5	13, 2	kV
$R_{i(eht)}$	6		MΩ
$I_{p-p}$ deflection	8		A
Supply voltage (VB)	12	12	v
current (I <sub>B</sub> )	830	1100	mA
Voltages of auxiliary windings	-102 V <sub>(p)</sub> , +	820 V <sub>(p)</sub>	

#### APPLICATION

This transformer has been designed to provide the required scanning amplitude for 24 cm (9 inch) 90° monitor tubes with a neck diameter of 28 mm intransistor equipped monitors presenting 625 lines at 50 frames per second (CCIR) or 525 lines at 60 frames per second (USA).

It is intended for use in conjunction with:

- deflection unit AT 1071/01;
- linearity control unit AT4036;
- line output transistor BD160;
- booster (efficiency) diode BYX55, BYX 71;
- e.h.t. rectifier device TV 18KT.

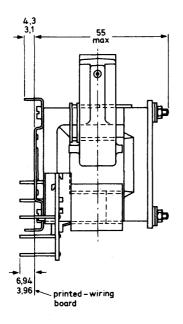
See also circuit diagram of Fig. 3.

## DESCRIPTION

The magnetic circuit of the transformer comprises Ferroxcube U and I-cores clamped together with brackets.

The primary windings and the auxiliary windings are situated on one leg of the core, the e.h.t. winding and the coupling winding are situated on the other leg. The e.h.t. winding is encapsulated in flame retardent polyester. The whole transformer meets the self-extinguishing requirements of IEC publication 65, para. 14.4 and UL492, para. 280-SE 1. The transformer is provided with four mounting pins.

External circuit connection is made to connecting pins, positioned as indicated in Fig. 1 enabling the unit to be soldered directly into a printed-wiring board.



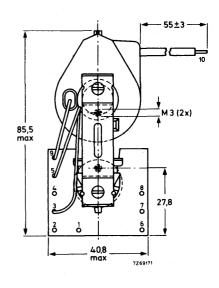


Fig. 1

# MOUNTING

For mounting on a printed-wiring board the fit of the connecting and mounting pins in a printed-wiring grid with a pitch of 2,54 mm (0,1 in) is illustrated in Fig. 2. The transformer core must be earthed

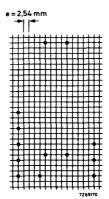


Fig. 2 Hole pattern for mounting on a printed-wiring board (solder side).

Grid holes 1,3 ± 0,1 mm.

# LINE OUTPUT TRANSFORMER for black and white monitors

# AT2102/01

# Temperature

The operating temperature of the core and the coils should not exceeded 95  $^{\rm o}{\rm C}$  under worst conditions, i.e. taking into account:

- over-voltage on the windings;
- low atmospheric pressure (at high altitudes) implying bad cooling by convection;
- high room temperature (up to 45 °C).

To satisfy these requirements it may be desirable to provide ample cool air circulation around the transformer.

# Distances

The following minimum distances between the transformer and neighbouring conductive flat surfaces must be maintained (it should be noticed that edges of conductive parts must have a greater distance):

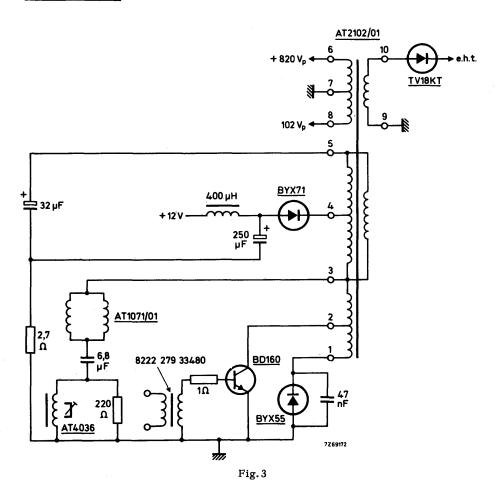
- from the e.h.t. winding, radially 20 mm, axially 12 mm;
- from the e.h.t. cap and lead 20 mm;
- from the primary coil 10 mm;
- between the upper edge of the rectifier socket and the primary coil 10 mm.

The transformer, and the leads and components carrying high voltage pulses should be kept free from metal particles, solder drops, etc.

**ELECTRICAL DATA**Measured in the circuit shown in Fig. 3 (auxiliary windings unloaded).

E.H.T. supply	I <sub>eht</sub> e.h.t. R <sub>i(eht)</sub>	mA kV MΩ	0, 03 14, 5	0, 23 13, 2
	$v_{B}$	v	12	12
Power supply	$v_B$	v	24	23,7
	I <sub>average</sub>	mA	830	1100
Output transistor	VCEM	v	180	
Deflection	I <sub>p-p</sub> Overscan	A	8	
Defrection	variation	%		3
Auxiliary windings,	connecting pin 6 connecting pin 8	v <sub>p</sub>	+820 -102	

# Application circuit



AT2140/10

This information is derived from development samples made available for evaluation. It does not form part of our data handbook system and does not necessarily imply that the device will go into production

# LINE OUTPUT TRANSFORMER

# QUICK REFERENCE DATA

l <sub>eht</sub>	0 μΑ   100 μΑ	<del></del>
E.H.T.	11 kV 10,2 kV	,
R <sub>i(eht)</sub>	8 MΩ	
Supply voltage (V <sub>B</sub> ) Supply current (I <sub>B</sub> )	8,8 V 8,8 V 920 mA 1100 m/	^
Deflection current	4,2 A (p-p) 4,1 A (	
Auxiliary voltages	15 V (d.c.), 75 V (d.c.), 200 V (	(d.c.)

# **APPLICATION**

This transformer has been designed to provide the required scanning amplitude for 31 cm (12 in) and 34 cm (14 in) 90° black and white monitor tubes with a neck diameter of 20 mm in video display monitors presenting 625 lines at 50 frames per second (CCIR) or 525 lines at 60 frames per second (USA).

It is intended for use in conjunction with:

deflection unit AT1074; adjustable linearity control unit AT4042/39:

line driver transformer AT4043/56.

# DESCRIPTION

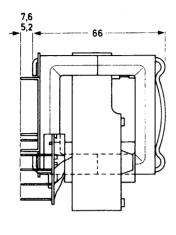
The magnetic circuit of the transformer comprises two Ferroxcube U-cores, clamped together with a bracket. The primary winding, the auxiliary windings and the e.h.t. winding are situated on one leg of the core. An e.h.t. rectifier diode is incorporated in the transformer. All windings are encapsulated in flame retardent polyester. The whole transformer meets the self-extinguishing and non-dripping properties of the American Underwriters' Laboratories rating mentioned in UL94SE-1.

The transformer is provided with four mounting pins. External circuit connection is made to connecting pins, enabling the unit to be soldered directly into a printed-wiring board.



# **MECHANICAL DATA**

Dimensions in mm



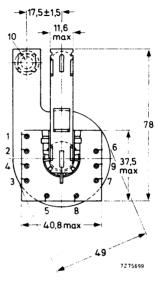


Fig. 1 Line output transformer AT2140/10.

# MOUNTING

The transformer may be mounted on a printed-wiring board. The fit of the connecting and mounting pins in a printed-wiring grid with a pitch of 2,54 mm (0,1 in) is illustrated in Fig. 2. The core of the transformer must be earthed.

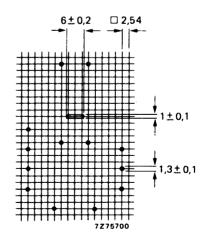


Fig. 2 Hole pattern for mounting on a printed-wiring board (solder side).

# Temperature

The operating temperature of the core and the coils should not exceed 90 °C, under worst conditions, i.e. taking into account:

over-voltage on the windings;

low atmospheric pressure (at high altitudes) implying bad cooling by convection; high room temperature (up to 45 °C)

To satisfy these requirements it may be desirable to provide ample cool air circulation around the transformer.

### Distances

The following minimum distances between the transformer and neighbouring conductive flat surfaces must be maintained (in proportion to their sharpness protruding parts must have a greater distance):

a. From the e.h.t. winding, radially 18 mm, axially 10 mm. b. From the e.h.t. lead 15 mm.

The transformer, and the leads and components carrying high-voltage pulses should be kept free from metal particles, solder drops etc.

# ELECTRICAL DATA (see also Fig. 3)

E.H.T. supply	l <sub>eht</sub> E.H.T. <sup>R</sup> i(eht)	0 μA 11 kV	100 μA 10,2 kV ΜΩ
Power supply	V <sub>B</sub>	8,8 V 920 mA	8,8 V 1100 mA
Output transistor	VCEM ICM	220 V 3,6 A	220 V 3,7 A
Deflection	Current Flyback ratio (average)	4,2 A (p-p) 9,4 %	4,1 A (p-p) 9,4 %
	Overscan variation	0 %	0 %

Auxiliary windings connecting pin 1 connecting pin 2 connecting pin 4

15 V (d.c.)

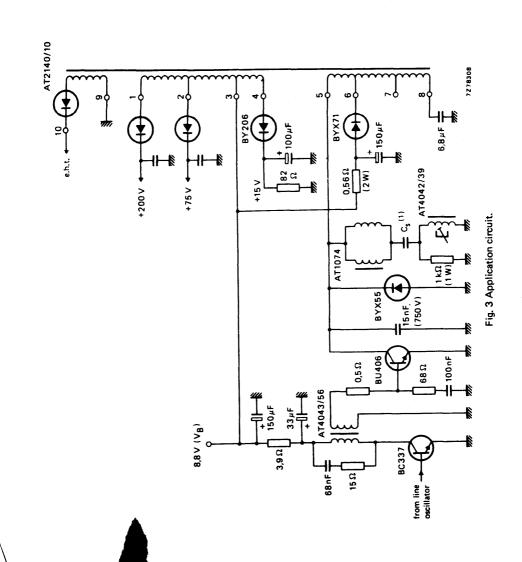
200 V (d.c.)

75 V (d.c.)









March 1978



# ADJUSTABLE LINEARITY CONTROL UNIT

# **APPLICATION**

This linearity control unit has been designed for use in black and white monitors with 24 cm (9 in) or 31 cm (12 in) 90° monitor tubes. It can be used in conjunction with deflection unit AT1071/03, line output transformer AT2102/02 and line driver transformer AT4043/56.

## DESCRIPTION

The unit consists of a coil wound on a Ferroxcube rod and two Ferroxdure magnets. One of these magnets has the shape of a half ring and is placed around the Ferroxcube rod under the coil: The other magnet is cylindrical; it is placed parallel to and clamped against the Ferroxcube rod opposite the first one. This magnet is provided with a square hole to facilitate turning of it to adjust the biasing field and so the linearity of the line deflection.

# MECHANICAL DATA

Dimensions in mm

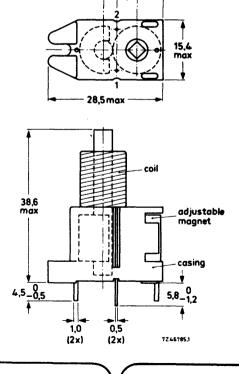


Fig. 1 Adjustable linearity control unit AT4036.



# **ELECTRICAL DATA**

When a sawtooth current (without S-correction) of 6 A (p-p), frequency 15 625 Hz, flyback ratio 18%, flows through the linearity control unit (one connection point to earth), the correction voltage is adjustable between 1,05 and 1,95 V.

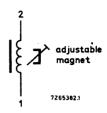


Fig. 2 Circuit diagram.

#### MOUNTING

The unit can be mounted either on printed-wiring boards by means of its two connection pins and two mounting pins (see Fig. 3), or on metal chassis by bending the two mounting pins and/or by means of a screw through an aperture in the casing (see Fig. 4). To prevent distortion of the magnetic field no iron part should approach the magnetic parts nearer than 3 mm. The coil should be shunted with a 1 W carbon resistor to damp ringing phenomena.

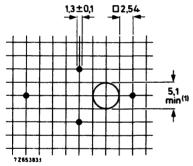


Fig. 3 Hole pattern for mounting on a printed-wiring board.
(1) Hole for bottom adjustment, if required.

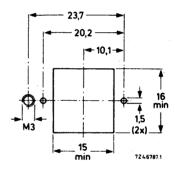


Fig. 4 Hole pattern for mounting on a chassis.

# QUICK REFERENCE DATA

For use in conjunction with:

- Monochrome deflection unit AT1040/15 and line output transformer AT2048/11
- Colour deflection units AT1027/AT1029 Series and line output transformer AT2055 Series.



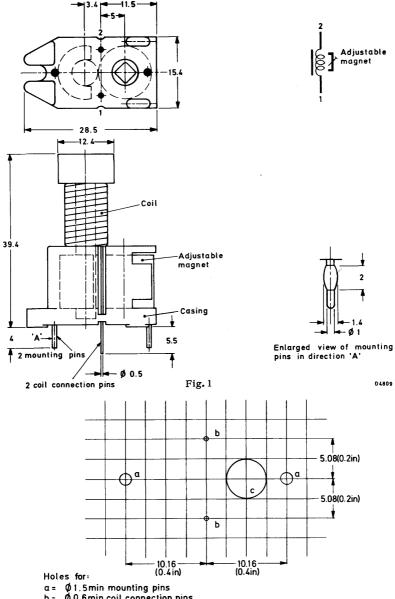
# GENERAL

The unit can be mounted either on the printed-wiring board by the two connection and two mounting pins (see figs. 1 and 2), or on a metal chassis by bending the two mounting pins, and/or by a screw through an aperture in the casing.

To prevent distortion of the magnetic field, a separation of 3mm must be maintained between the unit and any metal part.

A series resistor and capacitor are connected across the coil to damp out ringing. If under fault conditions the coil goes open circuit, the capacitor will limit the scanning coil current and enables a low wattage resistor to be used.





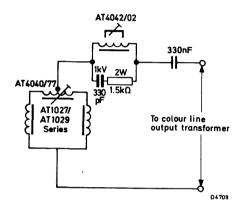
b = 0.6min coil connection pins c = 0.5min bottom adjustment

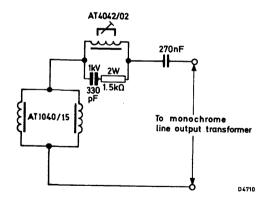
Fig. 2 Piercing di**a**gram D5399

# ELECTRICAL DATA

When a sawtooth current (without S-correction) of 2.8A peak-to-peak and frequency 15.625kHz, having a flyback of 18% flows through the linearity control unit, the correction voltage will be adjustable between 15V and 26V.

# APPLICATION CIRCUITS







# APPLICATION

This non-adjustable linearity control unit is designed for use in black and white television sets equipped with  $110^{\rm o}$  deflection angle picture tube.

It is intended for use in conjunction with:

- deflection unit AT 1040/15;
- line output transformer AT 2048/12.

# DESCRIPTION

The unit consists of a coil wound on a Ferroxcube rod, and a Ferroxdure magnet, which is placed around the rod next to the coil.

Dimensions in mm

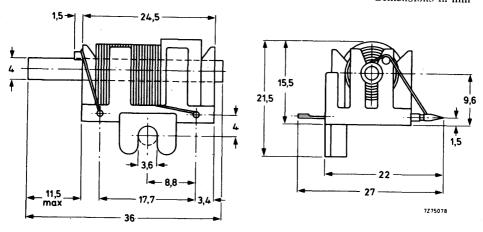


Fig. 1

# **ELECTRICAL DATA**

When a saw-tooth current (without S-correction) of 2, 2 A(p-p), frequency 15625 Hz, flyback ratio 18%, flows through the linearity control unit, the correction voltage is 17 V.

# MOUNTING

The unit can be mounted on printed-wiring boards by means of its two connection pins and two mounting pins (see Fig. 1). To prevent distortion of the magnetic field no iron part should approach the magnetic parts anywhere nearer than 3 mm.

# ADJUSTABLE LINEARITY CONTROL UNIT

# **APPLICATION**

This linearity control unit has been designed for use in black and white monitors with 31 cm (12 in) or 38 cm (15 in) 1100 monitor tubes. It can be used in conjunction with deflection unit AT1038/20, and line output transformer AT2102/05.

### DESCRIPTION

The unit consists of a coil wound on a Ferroxcube rod, two Ferroxdure magnets and one magnet of plastic-bonded Ferroxdure. The last mentioned magnet is placed around the Ferroxcube rod, above the coil. One of the Ferroxdure magnets has the shape of a half ring; it is placed around the Ferroxcube rod under the coil. The other Ferroxdure magnet is cylindrical; it is positioned parallel to and clamped against the Ferroxcube rod opposite the first one. It is provided with a square hole to facilitate turning to adjust the biasing field and so the linearity of the line deflection.

#### **MECHANICAL DATA**

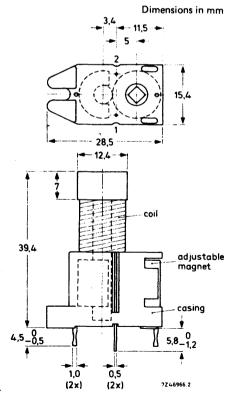


Fig. 1 Adjustable linearity control unit AT4034/01.



#### **ELECTRICAL DATA**

When a sawtooth current (without S-correction) of 2,4 A (p-p), frequency 15 625 Hz, flyback ratio 18%, flows through the linearity control unit (one connection point to earth), the correction voltage is adjustable between 12 and 24 V.

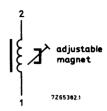


Fig. 2 Circuit diagram.

#### MOUNTING

The unit can be mounted either on printed-wiring boards by means of its two connection pins and two mounting pins (see Fig. 3), or on metal chassis by bending the two mounting pins and/or by means of a screw through an aperture in the casing (see Fig. 4). To prevent distortion of the magnetic field no iron part should approach the magnetic parts nearer than 3 mm. The coil should be shunted with a 1 W carbon resistor of 1500  $\Omega$  to damp ringing phenomena.

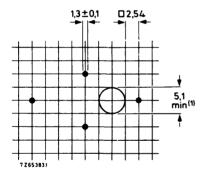


Fig. 3 Hole pattern for mounting on a printed-wiring board. (1) Hole for bottom adjustment, if required.

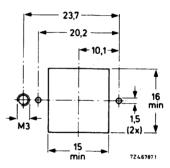


Fig. 4 Hole pattern for mounting on a chassis.

# ADJUSTABLE LINEARITY CONTROL UNIT

# **APPLICATION**

This linearity control unit has been designed for use in black and white monitors with 31 cm (12 in) or 38 cm (15 in) 110° monitor tubes. It can be used in conjunction with deflection unit AT1038/40, line output transformer AT2102/04 and line driver transformer AT4043/59. The unit is also to be used in colour television sets with a 110° colour picture tube.

# DESCRIPTION

The unit consists of a coil, mounted on a Ferroxcube rod, two Ferroxdure magnets and one plasto-ferrite magnet. One magnet has the shape of a ring and is placed around the Ferroxcube rod above the coils. One has the shape of a half ring and is placed around the Ferroxcube rod under the coils. The third magnet is prindrical; it is positioned to and clamped against the Ferroxcube rod opposite the half ring magnet. It is provided with a square hole to facilitate turning to adjust the biasing field and, therefore, the linearity of the line deflection.

**MECHANICAL DATA** 

Dimensions in mm

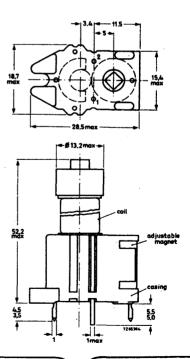


Fig. 1 Adjustable linearity control unit AT4042/08.



# **ELECTRICAL DATA**

When a sawtooth current of 6 A (p-p), frequency 15 625 Hz, fly-back ratio 18% (without S-correction) flows through the linearity control unit (coils connected in parallel, one connection point to earth), the correction voltage is adjustable between 15 and 25 V.

With a sawtooth current of 4,65 A (p-p) the correction voltage is adjustable between 8 and 15 V.

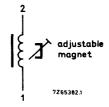


Fig. 2 Circuit diagram.

# MOUNTING

The unit can be mounted either on printed-wiring boards by means of its two connection pins and two mounting pins, or on metal chassis by bending the two mounting pins and/or by means of a screw through an aperture in the casing (see Fig. 4). To prevent distortion of the magnetic field, no iron part should approach the magnetic parts nearer than 3 mm. The coils should be shunted with carbon resistors to damp ringing phenomena; the value of resistor depends on applied line output transformer.

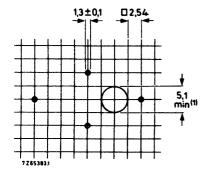


Fig. 3 Hole pattern for mounting on a printed-wiring board.
(1) Hole for bottom adjustment, if required.

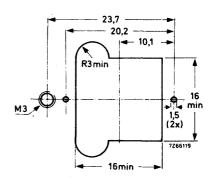


Fig. 4 Hole pattern for mounting on a chassis.

# DEVELOPMENT SAMPLE DATA

This information is derived from development samples made available for evaluation. It does not form part of our data handbook system and does not necessarily imply that the device will go into production

# LINE DRIVER TRANSFORMER

# **APPLICATION**

This transformer has been designed for use in black and white monitors. The required supply voltage is 24 V. The transformer is used in conjunction with deflection unit AT1038/40, line-output transformer AT2102/04, and linearity control unit AT4042/08.

# **MECHANICAL DATA**

Dimensions in mm

The magnetic circuit of the transformer comprises two Ferroxcube U-cores. The unit is provided with pins for mounting on a printed-wiring board.

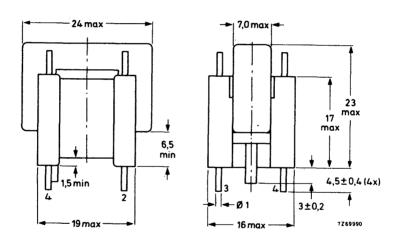


Fig. 1 Line driver transformer AT4043/59.

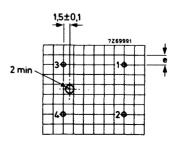


Fig. 2 Hole pattern for mounting on a printedwiring board (component side). Hole diameter 1,3 + 0,1 mm. e = 2,54 mm (0,1 in).



# **ELECTRICAL DATA**

**Application circuit** 

Inductance (primary, 1-2)

Leakage inductance (secondary) Transformation ratio

Maximum operating temperature

6.1 mH

12 µH ± 15%

4,18:1 95 °C

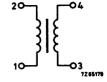


Fig. 3 Circuit diagram.

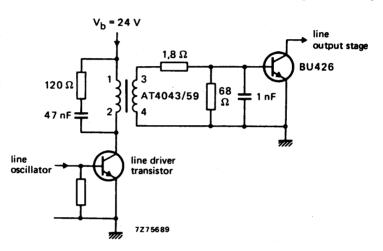
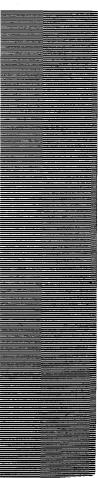


Fig. 4.

# COLOUR TELEVISION ASSEMBLIES



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# **DEFLECTION UNIT**

• with built-in 4-pole coils for symmetrizing of the line and field astigmatism

# **QUICK REFERENCE DATA**

Picture tube, gun arrangement	in line
diagonal	66 cm (26 in)
neck diameter	36,5 mm
Deflection angle	1100
Line deflection current, edge to edge at 25 kV	6,35 A p-p
Inductance of line coils, parallel connected	1,11 mH
Field deflection current, edge to edge at 25 kV	3,4 А р-р
Resistance of field coils, series connected	3,0 Ω
4-pole coils,	
sensitivity for line direction	±.max. 34 mm/A
sensitivity for field direction	± max. 23 mm/A
resistance (series connected)	1,6 Ω

## APPLICATION

This deflection unit has been designed for use with the 110° colour picture tube types A66-500X and A66-510X in CTV receivers in conjunction with:

AT1081

diode-split line output transformer line output transistor BU208A linearity control unit AT4042/38

# DESCRIPTION

multipole unit

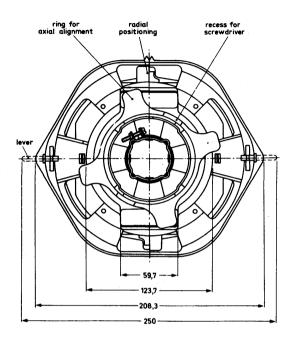
The saddle-shaped line and field deflection coils, and the Ferroxcube yoke ring with 4-pole unit, are supported by a plastic cap. This set is built into a plastic coaxial housing, which is provided with a plastic axial alignment ring. This ring enables the set to be axially adjusted over a distance of 6 mm, after the complete unit has been fastened on the neck of the picture tube with a clamping ring. The screw of the clamping ring is accessible with a screwdriver via a recess in the axial alignment ring. To correct the raster orientation with the complete unit in position on the picture tube neck, the coil assembly can be rotated by means of the protruding parts on the supporting ring, which can be reached by the top and bottom recesses in the coaxial housing. The whole coil assembly is locked in the required position by pushing the levers down until they block.

The unit meets the self-extinguishing requirements of IEC 65 para. 14.4 and UL94, SE1.



# MECHANICAL DATA Outlines

# Dimensions in mm



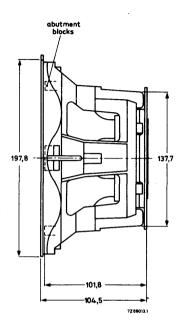


Fig. 1.

The unit is provided with soldering pins for connection.

# Mounting

To obtain easily reproducible and accurate alignment of the picture tube and the deflection unit, the cone of the picture tube has a moulded indexing ridge to centre the deflection unit housing. The deflection unit is brought into correct position by alignment of the protrusion on the housing with the location mark on the cone of the tube. The unit must be pressed against the cone, so that the housing is indexed by the moulded ridge on the cone. The unit is then fixed by tightening the screw in the clamping ring at the rear. The screw should be tightened with a torque of 1,2 to 1,4 Nm.

# **ELECTRICAL DATA**

Line coils, parallel connected inductance

resistance at 25 °C

Line deflection current, edge to edge at 25 kV

Field coils, series connected

inductance resistance at 25 °C

Field deflection current, edge to edge at 25 kV

4-pole coils, sensitivity for line direction

sensitivity for field direction

resistance (series connected)

Maximum operating temperature

1,11 mH ± 4% 1,2 Ω ± 10%

6,35 A p-p

,,55 ∧ b-h

3.5 mH ± 10%

3,0 Ω ± 7%

3,35 A p-p

± max. 34 mm/A

± max. 23 mm/A

1,6 Ω

95 °C

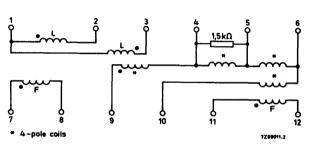


Fig. 2 Connection diagram. L = Line, F = Field.

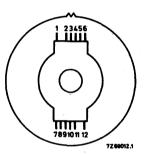


Fig. 3 Terminal location. The pin numbering corresponds to that in Fig. 2.

# **BEAM CORRECTIONS**

With the deflection unit AT1080 and the multipole unit AT1081 mounted on the tube A66-500X or A66-510X, the following corrections may be required:

Maximum required horizontal displacement of the electron beams with respect to the phosphor stripes by the purifying magnet of the AT1081 (note 1)

45 µm

Static convergence deviations must be corrected by the adjustable four-pole and six-pole fields of the AT1081 centred around the tube axis.

Maximum required compensation for static convergence
4-pole device: red opposite to blue (in any direction)

6 mm 3 mm

4-pole device: red opposite to blue (in any direction)
6-pole device: red and blue to green (in any direction)

Notes, see page 4.



# North-South raster shape correction circuitry is not required.

To obtain a symmetrical shape for the horizontal lines at the upper and lower parts of the screen, the unit AT1081 comprises an additional two-pole correction magnet giving a displacement of the beam in the centre of the screen in vertical direction of maximum

5 mm

Maximum centring error in any direction after colour-purity, static convergence, and horizontal centre line correction

5 mm

With respect to dynamic convergence, the display system, consisting of picture tube A66-500X or A66-510X and deflection unit AT1080, is inherently self converging. However, a small systematic correction is required on the vertical axis, and also small corrections should be made to compensate for tolerances and asymmetries in the tube and deflection unit combination. For this purpose two types of dynamic magnetic four-pole fields can be used. One is generated by additional windings on the yoke ring of the deflection unit energized by adjustable sawtooth currents synchronized with scanning. The other type is generated by sawtooth and parabolic currents which are synchronized with scanning and flow through the deflection coils.

Compensation to be provided by these corrections:

<ul> <li>horizontal red-to-blue distance at the end of the</li> </ul>		
horizontal axis (line symmetry)	(note 2)	0 ± 2 mm
<ul> <li>horizontal red-to-blue distance at the top of the</li> </ul>		
vertical axis (field symmetry top)	(note 3)	$3,5 \pm 1,5 \text{ mm}$
- horizontal red-to-blue distance at the bottom of		
the vertical axis (field symmetry bottom)	(note-3)	$3,5 \pm 1,5  \text{mm}$
- vertical red-to-blue distance at the ends of the		
horizontal axis in opposite directions (line balance)	(note 4)	0 ± 1,5 mm
<ul> <li>vertical red-to-blue distance at the ends of the</li> </ul>		
vertical axis (field balance)	(note 5)	0 ± 1,2 mm

Application information available on request.

## Notes

- 1. Purity adjustment in vertical direction is not required.
- This correction is made by feeding a sawtooth current of line frequency through the additional four-pole windings on the deflection unit.
- 3. This correction is made by feeding a rectified sawtooth current of field frequency through the additional four-pole windings on the deflection unit.
- 4. This correction is made by unbalancing the line deflection coils.
- 5. This correction is made by unbalancing the field deflection coils.





# MULTIPOLE UNIT

# **QUICK REFERENCE DATA**

for undeflected beams		
min.	45	μm
min.	8	mm
min.	4	mm
min.	5	mm
	min. min. min.	min. 45 min. 8 min. 4

#### APPLICATION

This unit has been designed for the colour picture tubes A66-500X, A66-510X, A56-500X, A56-510X, A51-500X and A51-510X, with in-line gun arrangement and the deflection units AT1080, AT1083/01 and AT1085. Its purpose is threefold:

- horizontal colour-purity adjustment
- static convergence adjustment
- adjustment of raster symmetry in N and S or adjustment of the horizontal axis for optimum straightness.

# DESCRIPTION

The unit incorporates four ring-shaped permanent magnets, supported by non-magnetic plastic support rings, and a cam-actuated collet, which enables the unit to be clamped to the neck of the picture tube. The magnetic rings are made up of an inner and an outer ring coupled by non-magnetic pinion gears to form an epicyclic train. The support rings carry the pinion gears. The magnetic rings comprise:

- two pairs of 2-pole magnets
- one pair of 4-pole magnets
- one pair of 6-pole magnets

(each pair consisting of an inner and outer ring of identical magnetic configuration). The support rings of both the 2-pole rings are fixed to the collet, those of the 4- and 6-pole rings are rotatable. Rotating the lug on an outer magnetic ring varies the resultant field strength.

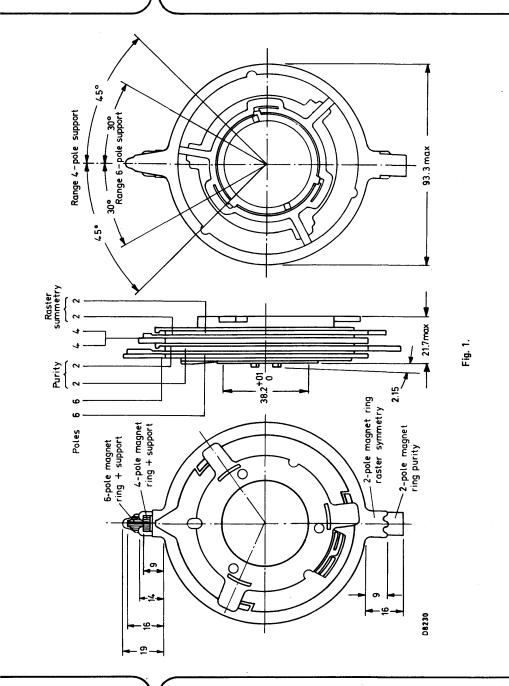
Rotating the lug on a support ring varies the direction of the resultant field.

# **MECHANICAL DATA**

Dimensions (mm)

See Fig. 1 on next page.





2 April 1978



# Mounting

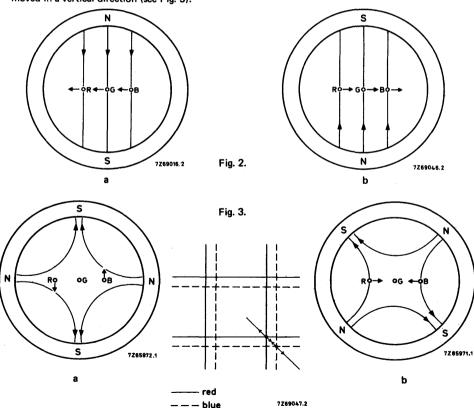
Before mounting the multipole unit, the lug on the rear end of the collet must be rotated anti-clockwise. The unit is slid over the neck of the picture tube and pressed to the deflection unit. Two protrusions on the front of the unit and the corresponding recesses on the back of the deflection unit, will bring the unit into correct position. By rotating the lug on the collet clockwise the unit will be clamped.

#### **ADJUSTMENTS**

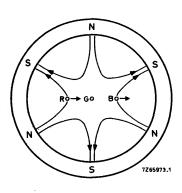
Horizontal colour purity is obtained by varying the field strength of the 2-pole magnet situated between the 4-pole and 6-pole magnets (see Figs 1 and 2).

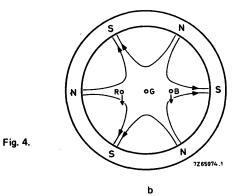
Vertical colour purity adjustment is not required (see data on colour picture tubes).

The static convergence is adjusted by varying the field strength and direction of the 4-pole and 6-pole. The 4-pole field moves the outer electron beams (red and blue) equally in opposite directions (see Fig. 3). The 6-pole field moves the outer electron beams equally in the same direction (see Fig. 4). The centre beam (green) is unaffected. Horizontal axis or raster symmetry is adjusted by varying the field strength of the 2-pole magnet situated at the rear of the unit (see Fig. 1). All three beams are equally moved in a vertical direction (see Fig. 5).



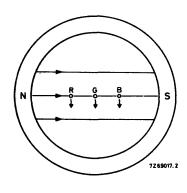






\_\_\_\_ green \_\_\_\_ red - blue 7269048.1

C



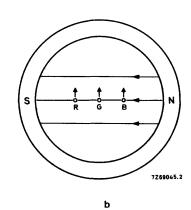


Fig. 5.

# **DEFLECTION UNIT**

with built-in 4-pole coils for symmetrizing of the line and field astigmatism

## **QUICK REFERENCE DATA**

Picture tube, gun arrangement	in line
diagonal	55 cm (22 in)
neck diameter	36,5 mm
Deflection angle	1100
Line deflection current, edge to edge at 25 kV	6,2 A p-p
Inductance of line coils, parallel connected	1,14 mH
Field deflection current, edge to edge at 25 kV	3,4 A p-p
Resistance of field coils, series connected	3,36 Ω
4-pole coils,	
sensitivity for line direction	± max. 25 mm/A
sensitivity for field direction	± max. 18 mm/A
resistance (series connected)	1,4 Ω

#### APPLICATION

This deflection unit has been designed for use with the 110° colour picture tube types A56-500X and A56-510X in CTV receivers in conjunction with:

diode-split line output transformer

AT2076/30 and

line output transistor

BU208A

linearity control unit

AT4042/38

multipole unit

AT1081

# DESCRIPTION

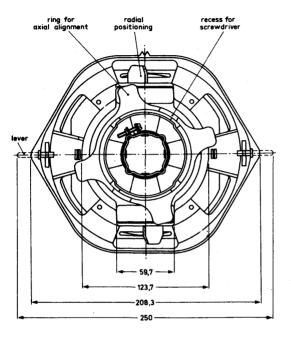
The saddle-shaped line and field deflection coils, and the Ferroxcube yoke ring with 4-pole unit, are supported by a plastic cap. This set is built into a plastic coaxial housing, which is provided with a plastic axial alignment ring. This ring enables the set to be axially adjusted over a distance of 5 mm, after the complete unit has been fastened on the neck of the picture tube with a clamping ring. The screw of the clamping ring is accessible with a screwdriver via a recess in the axial alignment ring. To correct the raster orientation with the complete unit in position on the picture tube neck, the coil assembly can be rotated by means of the protruding parts on the supporting ring, which can be reached by the top and bottom recesses in the coaxial housing. The whole coil assembly is locked in the required position by pushing the levers down until they block.

The unit meets the self-extinguishing requirements of IEC 65 para. 14.4 and UL94, SE1.



# MECHANICAL DATA Outlines

Dimensions in mm



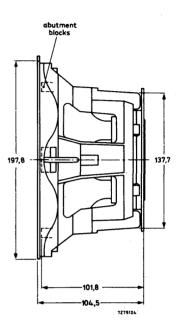


Fig. 1.

The unit is provided with soldering pins for connection.

# Mounting

To obtain easily reproducible and accurate alignment of the picture tube and the deflection unit, the cone of the picture tube has a moulded indexing ridge to centre the deflection unit housing. The deflection unit is brought into correct position by alignment of the protrusion on the housing with the location mark on the cone of the tube. The unit must be pressed against the cone, so that the housing is indexed by the moulded ridge on the cone. The unit is then fixed by tightening the screw in the clamping ring at the rear. The screw should be tightened with a torque of 1,2 to 1,4 Nm.

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# **ELECTRICAL DATA**

Line coils, parallel connected inductance resistance at 25 °C.

Line deflection current, edge to edge at 25 kV

Field coils, series connected inductance resistance at 25 °C

Field deflection current, edge to edge at 25 kV

4-pole coils,

sensitivity for line direction sensitivity for field direction resistance (series connected)

Maximum operating temperature

 $0.9~\Omega \pm 10\%$  6.2~A~p-p  $3.9~mH \pm 10\%$   $3.36~\Omega \pm 7\%$ 3.4~A~p-p

1.14 mH ± 4%

± max. 25 mm/A ± max. 18 mm/A

1.4 Ω

95 °C

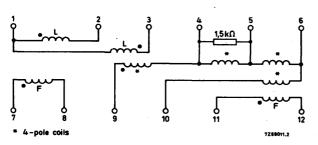


Fig. 2 Connection diagram. L = Line, F = Field.

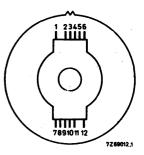


Fig. 3 Terminal location. The pin numbering corresponds to that in Fig. 2.

#### **BEAM CORRECTIONS**

With the deflection unit AT1083/01 and the multipole unit AT1081 mounted on the tube A56-500X or A56-510X, the following corrections may be required:

Maximum required horizontal displacement of the electron beams with respect to the phosphor stripes by the purifying magnet of the AT1081 (note 1)

45 µm

Static convergence deviations must be corrected by the adjustable four-pole and six-pole fields of the AT1081 centred around the tube axis.

Maximum required compensation for static convergence 4-pole device: red opposite to blue (in any direction) 6-pole device: red and blue to green (in any direction)

5,5 mm 2.8 mm

Notes, see page 4.



# North-South raster shape correction circuitry is not required

To obtain a symmetrical shape for the horizontal lines at the upper and lower parts of the screen, the unit AT1081 comprises an additional two-pole correction magnet giving a displacement of the beam in the centre of the screen in vertical direction of maximum

4,5 mm

Maximum centring error in any direction after colour-purity, static convergence, and horizontal centre line correction

4.5 mm

With respect to dynamic convergence, the display system, consisting of picture tube A56-500X or A56-510X and deflection unit AT1083/01 is inherently self converging. However, small corrections should be made to compensate for tolerances and symmetries in the tube and deflection unit combination. For this purpose two types of dynamic magnetic four-pole fields can be used. One generated by additional windings on the yoke ring of the deflection unit energized by adjustable sawtooth currents synchronized with scanning. The other type is generated by sawtooth and parabolic currents which are synchronized with scanning and flow through the deflection coils.

Compensation to be provided by these corrections:

<ul> <li>horizontal red-to-blue distance at the end of the</li> </ul>		
horizontal axis (line symmetry)	(note 2)	0 ± 1,5 mm
<ul> <li>horizontal red-to-blue distance at the ends of the</li> </ul>		1.0
vertical axis (field symmetry)	(note 3)	0 ± 1,5 mm
<ul> <li>vertical red-to-blue distance at the ends of the</li> </ul>		
horizontal axis in opposite directions (line balance)	(note 4)	0 ± 1,0 mm
<ul> <li>vertical red-to-blue distance at the ends of the</li> </ul>		
vertical axis (field balance)	(note 5)	0 ± 1,0 mm

Application information available on request.

#### Notes

- 1. Purity adjustment in vertical direction is not required.
- 2. This correction is made by feeding a sawtooth current of line frequency through the additional four-pole windings on the deflection unit.
- This correction is made by feeding a rectified sawtooth current of field frequency through the additional four-pole windings on the deflection unit.

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- 4. This correction is made by unbalancing the line deflection coils.
- 5. This correction is made by unbalancing the field deflection coils.



# **DEFLECTION UNIT**

• with built-in 4-pole coils for symmetrizing of the line and field astigmatism

#### **OUICK REFERENCE DATA**

Picture tube, gun arrangement	in line
diagonal	51 cm (20 in)
neck diameter	36,5 mm
Deflection angle	1100
Line deflection current, edge to edge at 25 kV	6,2 A p-p
Inductance of line coils, parallel connected	1,14 mH
Field deflection current, edge to edge at 25 kV	3,4 А р-р
Resistance of field coils, series connected	3,36 Ω
4-pole coils,	
sensitivity for line direction	± max. 23 mm/A
sensitivity for field direction	± max. 16 mm/A
resistance (series connected)	1,4 Ω

# APPLICATION

This deflection unit has been designed for use with the 110° colour picture tube types A51-500X and A51-510X in CTV receivers in conjunction with:

diode-split line output transformer AT

AT2076/30 and BU208A

line output transistor

AT4042/38

linearity control unit

AT4004

multipole unit

AT1081

# DESCRIPTION

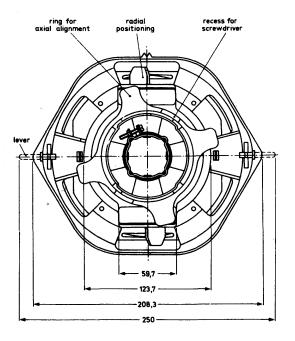
The saddle-shaped line and field deflection coils, and the Ferroxcube yoke ring with 4-pole unit, are supported by a plastic cap. This set is built into a plastic coaxial housing, which is provided with a plastic axial alignment ring. This ring enables the set to be axially adjusted over a distance of 5 mm, after the complete unit has been fastened on the neck of the picture tube with a clamping ring. The screw of the clamping ring is accessible with a screwdriver via a recess in the axial alignment ring. To correct the raster orientation with the complete unit in position on the picture tube neck, the coil assembly can be rotated by means of the protruding parts on the supporting ring, which can be reached by the top and bottom recesses in the coaxial housing. The whole coil assembly is locked in the required position by pushing the levers down until they block.

The unit meets the self-extinguishing requirements of IEC 65 para. 14.4 and UL94, SE1.



# **MECHANICAL DATA** Outlines

# Dimensions in mm



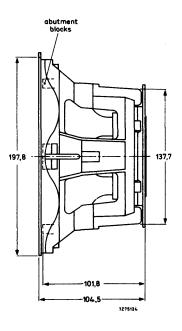


Fig. 1.

The unit is provided with soldering pins for connection.

# Mounting

To obtain easily reproducible and accurate alignment of the picture tube and the deflection unit, the cone of the picture tube has a moulded indexing ridge to centre the deflection unit housing. The deflection unit is brought into correct position by alignment of the protrusion on the housing with the location mark on the cone of the tube. The unit must be pressed against the cone, so that the housing is indexed by the moulded ridge on the cone. The unit is then fixed by tightening the screw in the clamping ring at the rear. The screw should be tightened with a torque of 1,2 to 1,4 Nm.

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#### **ELECTRICAL DATA**

resistance at 25 °C

Line coils, parallel connected inductance

Line deflection current, edge to edge at 25 kV

Field coils, series connected inductance resistance at 25 °C

Field deflection current, edge to edge at 25 kV

4-pole coils, sensitivity for line direction sensitivity for field direction resistance (series connected)

Maximum operating temperature

1,14 mH ± 4% 0,9 Ω ± 10% 6.2 A p-p

3,9 mH ± 10% 3,36 Ω ± 7% 3,4 A p-p

 $\pm$  max. 23 mm/A  $\pm$  max. 16 mm/A 1,4  $\Omega$ 

95 °C

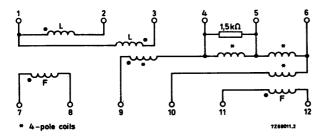


Fig. 2 Connection diagram. L = Line, F = Field.

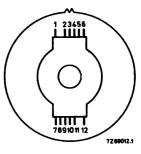


Fig. 3 Terminal location. The pin numbering corresponds to that in Fig. 2.

# **BEAM CORRECTIONS**

With the deflection unit AT1085 and the multipole unit AT1081 mounted on the tube A51-500X or A51-510X the following corrections may be required:

Maximum required horizontal displacement of the electron beams with respect to the phosphor stripes by the purifying magnet of the AT1081 (note 1)

45 µm

Static convergence deviations must be corrected by the adjustable four-pole and six-pole fields of the AT1081 centred around the tube axis.

Maximum required compensation for static convergence 4-pole device: red opposite to blue (in any direction) 6-pole device: red and blue to green (in any direction)

5 mm 2.5 mm

Notes, see page 4.



# North-South raster shape correction circuitry is not required.

To obtain a symmetrical shape for the horizontal lines at the upper and lower parts of the screen, the unit AT1081 comprises an additional two-pole correction magnet giving a displacement of the beam in the centre of the screen in vertical direction of maximum

4 mm

Maximum centring error in any direction after colour-purity, static convergence, and horizontal centre line correction

4 mm

With respect to dynamic convergence, the display system, consisting of picture tube A51-500X or A51-510X and deflection unit AT1085 is inherently self converging. However, a small fixed line parabola correction of 1,3 mm, is required on the horizontal axis and also small corrections should be made to compensate for tolerances and asymmetries in the tube and deflection unit combination. For this purpose two types of dynamic magnetic four-pole fields can be used. One is generated by additional windings on the yoke ring of the deflection unit energized by adjustable sawtooth currents synchronized with scanning. The other type is generated by sawtooth and parabolic currents which

Compensation to be provided by these corrections:

are synchronized with scanning and flow through the deflection coils.

- horizontal red-to-blue distance at the end of the		
horizontal axis (line symmetry)	(note 2)	0 ± 1,5 mm
<ul> <li>horizontal red-to-blue distance at the ends of the</li> </ul>		
vertical axis (field symmetry)	(note 3)	0 ± 1,5 mm
<ul> <li>vertical red-to-blue distance at the ends of the</li> </ul>		
horizontal axis in opposite directions (line balance)	(note 4)	0 ± 1,0 mm
<ul> <li>vertical red-to-blue distance at the ends of the</li> </ul>		
vertical axis (field balance)	(note 5)	0 ± 1,0 mm

Application information available on request.

#### Notes

- 1. Purity adjustment in vertical direction is not required.
- This correction is made by feeding a sawtooth current of line frequency through the additional four-pole windings on the deflection unit.
- This correction is made by feeding a rectified sawtooth current of field frequency through the additional four-pole windings on the deflection unit.

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- 4. This correction is made by unbalancing the line deflection coils.
- 5. This correction is made by unbalancing the field deflection coils.

# DIODE SPLIT LINE OUTPUT TRANSFORMER

# **QUICK REFERENCE DATA**

For transistor line output stages	-
l <sub>eht</sub>	max. 1.5 mA
E.H.T.	25 kV
Ri(eht)	2 ΜΩ
I <sub>p-p</sub> deflection (incl. 6% overscan)	6.5 A
Load inductance (of line deflection coils)	1.12 mH
Supply voltage (VB')	148 V
Supply current (I <sub>average</sub> ) at I <sub>eht</sub> = 1.5 mA	660 mA
Voltages of primary windings*	+64 V <sub>p</sub> , +105 V <sub>p</sub> , +335 V <sub>p</sub> , +520 V <sub>p</sub>
Voltages of auxiliary windings	$-335 \text{ V}_{\text{p}}$ , $-160 \text{ V}_{\text{p}}$ , $+160 \text{ V}_{\text{p}}$ , $+335 \text{ V}_{\text{p}}$ picture tube heater voltage

## APPLICATION

This transformer has been designed to provide the required scanning amplitude for 20AX 110° colour picture tubes with a neck diameter of 36.5 mm in transistor equipped television receivers presenting 625 lines at 50 fields per second (CCIR) or 525 lines at 60 fields per second (USA).

It is intended for use in conjunction with:

- deflection unit AT1080, AT1083/01 or AT1085;
- linearity control unit AT4042/38;
- line output transistor BU208A;
- a screened e.h.t. cable with a length of 1 m (available under catalogue number 3122 137 58254),
   as shown in the circuit diagram of Fig.3.

#### DESCRIPTION

The magnetic circuit of the transformer comprises 2 Ferroxcube U-cores, screwed together. The primary winding of aluminium foil and the secondary windings are situated on one leg of the core. The e.h.t. winding is moulded in flame retarding polyester, meeting the self-extinguishing requirements of IEC65, para. 14.4 and UL492, para. 280-SE1. The transformer is provided with 2 M3 screw-studs for mounting. \*\*External circuit connection is made to connecting pins, positioned as indicated in Fig.1 enabling the unit to be soldered directly into a printed-wiring board (Fig.2)

<sup>\*\*</sup>For mounting on the printed-wiring board a washer of 20 mm in diameter has to be used. Tightening torque on printed-wiring board: 500 + 100 mNm.



<sup>\*</sup> D.C. component on the pulses is VB' (see Fig.3).

**MECHANICAL DATA** 

Dimensions in mm

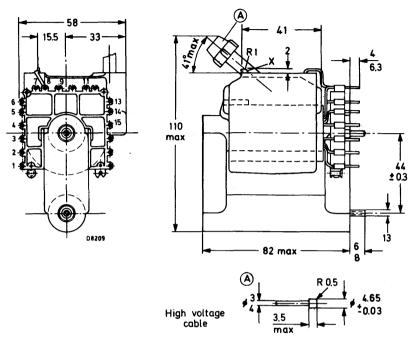


Fig.1

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Mass

500 g approximately

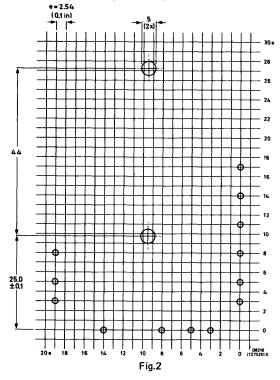
Solderability

in accordance with IEC 68, Test T

#### MOUNTING

The transformer may be mounted on either a printed-wiring board or, under certain conditions, on a metal chassis. Two securing studs (M3) are provided. The fit of the connecting and the mounting pins in a printed-wiring grid with a pitch of 2.54 mm is illustrated in Fig.2.

# RECOMMENDED PIERCING DIAGRAM (solder side)



Hole pattern for mounting on a printed-wiring board. Hole diameter 1.3 ± 0.1 mm.

Whether the transformer is board or chassis mounted, the core must be earthed.

#### Temperature

The operating temperature of the e.h.t. coil should not exceed +85 °C under worst conditions, i.e. taking into account:

- over-voltage on the coils;
- low atmospheric pressure (at high altitudes) implying bad cooling by convection;
- high ambient temperature (up to 45 °C).

To satisfy this requirement it may be necessary to provide an ample cool air flow around the transformer.



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# Distances

The following minimum distances between the transformer and neighbouring conductive flat surfaces must be maintained (it should be noticed that edges of conductive parts must have a greater distance):

From the e.h.t. coil radially, 10 mm axially 10 mm

The transformer, and the leads and components carrying high voltage pulses, should be kept free from metal particles, solder drops etc.

# ELECTRICAL DATA (measured in circuit of Fig.3, mains voltage 220 V)

l <sub>eht</sub>	50	1500	μΑ
e.h.t.	25.0	21.8	kV
R <sub>i(eht)</sub>	2		$\Omega$ M
• •	163	163	V
_	148	141.5	V
=	540	760	mA
	1200	1180	V
	4.1	4.25	Α
	6.5	6.2	Α
overscan	6	7	%
flyback time	11		
V <sub>focus</sub>	630	5.65	kV
$V_1 - 2$	7.6	7.4	V <sub>rms</sub>
V <sub>3</sub>	-335		V (+38 V d.c.)
V <sub>4</sub>	-160		V
V <sub>6</sub>	+160		V
V7**	+335		٧
Vg**	+105		V
V <sub>11</sub> **	+64		V
V   }			
V <sub>14</sub> **	+520		٧
	e.h.t.  Ri(eht)  VB  VB'  I(AV)  VCEM  ICM  Ip-p  overscan  flyback time  Vfocus  V1 - 2  V3  V4  V6  V7**  V9**	e.h.t. 25.0  R <sub>i(eht)</sub> 2  VB 163  VB' 148  I(AV) 540  VCEM 1200  ICM 4.1  Ip-p 6.5  overscan 6  flyback time 11  Vfocus 630   V1 - 2 7.6  V3 -335  V4 -160  V7** +335  V9** +105	e.h.t. 25.0 21.8  R <sub>i(eht)</sub> 2  V <sub>B</sub> 163 163  V <sub>B</sub> ' 148 141.5  I(AV) 540 760  VCEM 1200 1180  ICM 4.1 4.25  I <sub>p-p</sub> 6.5 6.2  overscan 6 7  flyback time 11  Vfocus 630 5.65   V1 - 2 7.6 7.4  V3 -335  V4 -160  V6 +160  V7** +335  V9** +105



<sup>\*</sup> Class-B video stage.

<sup>\*\*</sup>D.C. component on these pulses is VB'.

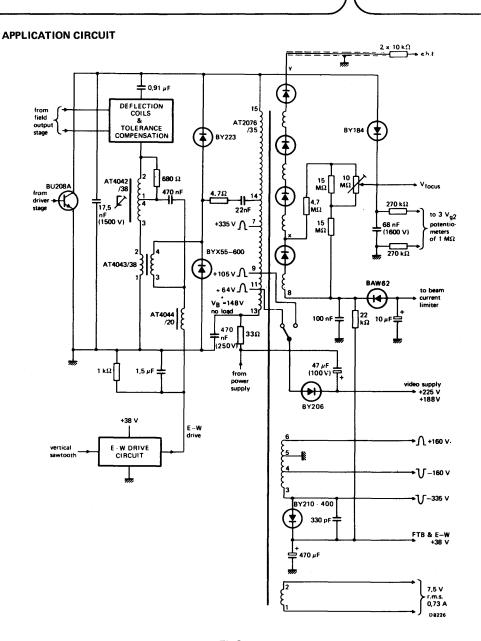


Fig.3



**Mullard** 

June 1978

# **DIODE SPLIT LINE OUTPUT TRANSFORMER**

# QUICK REFERENCE DATA

leht	1.5 mA
E.H.T.	25 kV
R <sub>i(eht)</sub>	2.0 ΜΩ
I <sub>D-D</sub> deflection (incl. 6% overscan)	6.5 A
Load inductance (of line deflection)	1.12 mH
Input voltage (VB')	148 V
current (IB) at leht = 1.5 mA	605 mA
Voltages of auxiliary windings	– 170, + 170 + 330 V <sub>p</sub>
Picture tube heater voltage	6.8 V
Primary taps	+ 510, + 330, + 105, + 64 V*

## **APPLICATION**

This transformer has been designed to provide the required scanning amplitude for 20AX 110° colour picture tubes with a neck diameter of 36.5 mm in transistor equipped television receivers 625 lines at 50 fields per second (CCIR) or 525 lines at 60 fields per second (USA).

It is intended for use in conjunction with:

deflection unit AT1080 or AT1083/01 or AT1085

linearity control unit AT4042/38

line driver transformer AT4043/50

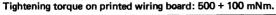
line output transistor BU208A

a screened e.h.t. cable with a length of 1 metre (available under catalogue number 3122 137 58250) as shown in the circuit diagram of Fig.3.

# DESCRIPTION

The magnetic circuit of the transformer comprises two Ferroxcube U-cores, screwed together. The primary winding of aluminium foil and the secondary windings are situated on one leg of the core. The windings are impregnated in flame retardant polyester, meeting the self-extinguishing requirements of IEC65, para. 14.4 and UL492, para. 280-SEI. The transformer is provided with two M3 screws for mounting \*\*. External circuit connection is made to connecting pins, positioned as indicated in Fig.1 enabling the unit to be soldered directly into a printed wiring board (Fig.2).

- \* The voltages have a d.c. component, the average value of which is Vg'.
- \*\*For mounting on a printed wiring board a washer of 20 mm in diameter has to be used.





**MECHANICAL DATA** 

Dimensions in mm

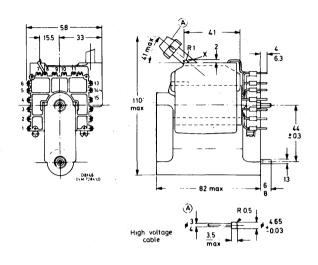


Fig.1

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Weight: 500 g approximately

Solderability: in accordance with IEC, Test T.

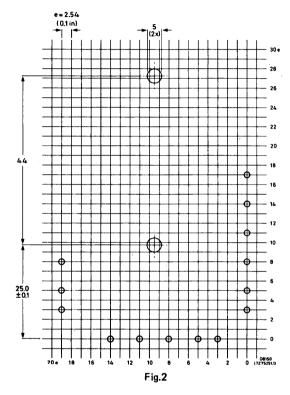
2

## MOUNTING

The transformer may be mounted on either a printed-wiring board or, under certain conditions, on a metal chassis. Two securing studs (M3) are provided.

The fit of the connecting and the mounting pins in a printed-wiring grid with a pitch of 2.54 mm is illustrated in Fig.2.

# RECOMMENDED PIERCING DIAGRAM (solder side)



Grid spacing: 2.54 mm (0.1 in.) Hole diameter 1.3 ± 0.1 mm

Whether the transformer is board or chassis mounted, the core must be earthed.

# Temperature

The operating temperature on the e.h.t. coil should not exceed + 85 °C under worst conditions.

To satisfy this requirement it may be necessary to provide an ample cool air flow around the transformer.



## Distances

The following minimum distances between the transformer and neighbouring conductive flat surfaces must be maintained (edges of conductive parts must have a greater distance):

From the e.h.t. coil, radially 10 mm axially 10 mm

The transformer, and the leads and components carrying high voltage pulses, should be kept free from metal particles, solder drops etc.

# **ELECTRICAL DATA** (measured in circuit of Fig.3)

E.H.T. supply	l <sub>eht</sub>	50	1500	μΑ
	e.h.t.	25.0	21.8	kV
	R <sub>i(eht)</sub>	2		$M\Omega$
Power supply	$V_{B}$	159	159	V
	V <sub>B</sub> ′	148	144	V
	I(AV)	400	605	mA
Output transistor	VCEM	1200	1130	V
	<sup>†</sup> CM	4.0	4.1	Α
Deflection	I <sub>p-p</sub>	6.5	6.2	Α
	overscan	6	8	%
	flyback time	11.5	11.55	μs
Focus voltage	V <sub>focus</sub>	6.15	5.45	kV
Peak voltage between pins 3 - 10	V <sub>3</sub>	-170		V
5 – 4	V <sub>5</sub>	+170		V
6 – 4	V <sub>6</sub>	+330		V
7 – 13*	V <sub>7</sub>	+330		V
9 – 13*	V <sub>9</sub>	+100		V
11 – 13*	V <sub>11</sub>	+50		V
14 — 13*	V <sub>14</sub>	+510		V
Picture tube heater voltage	V <sub>1</sub> – 2 (rms)	6.8	6.4	V (typical load 4.67 W)
Scan voltage after rectification				
from pins $3-10$	V <sub>3</sub> ′	+14	+13.2	V <sub>dc</sub> (load 3.4 W)
pins 4 - 10	V4'	+38	+36	V <sub>dc</sub> (load 24 W)
Video supply from pins $9-10$		+228	+218	V <sub>dc</sub> (load 3.0 W)
pins 11 - 10		+188	+178	V <sub>dc</sub> (load 3.0 W)

<sup>\*</sup> These voltages have a d.c. component the average value of which is Vg'.

# **APPLICATION CIRCUIT**

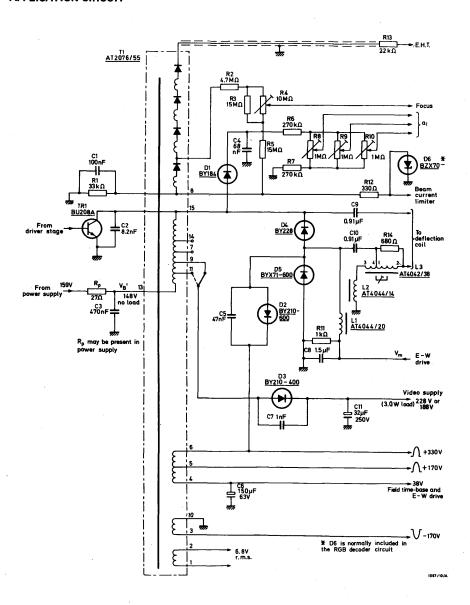


Fig.3



QUICK REFERENCE DATA				
I <sub>eht</sub>	max. 1.5 mA			
E.H.T.	8. 4 kV			
R <sub>i(eht)</sub>	2.0 ΜΩ			
I <sub>p-p</sub> deflection	6.5 A			
Load inductance (of line deflection)	1. 12 mH			
Supply voltage $(V_B')$ current $(I_B)$ at $I_{eht} = 1.5$ mA	148 V 605 mA			
Voltages of auxiliary windings	-320 $V_p$ , -155 $V$ , +155 $V_p$ , +320 $V_p$ picture tube heater voltage			

## APPLICATION

This transformer has been designed to provide the required scanning amplitude for 20AX 110° colour picture tubes with a neck diameter of 36.5 mm in transistor equipped television receivers presenting 625 lines at 50 fields per second (CCIR) or 525 lines at 60 fields per second (USA).

It is intended for use in conjunction with:

- deflection units AT1080, AT1083/01 or AT1085
- linearity control unit AT4042/38
- line output transistor BU208A
- e.h.t. multiplier LP1194/40 or LP1196/40 according to circuit diagram of Fig. 3.

## DESCRIPTION

The magnetic circuit of the transformer comprises a Ferroxcube U- and a Ferroxcube I-core, clamped together with brackets. The primary windings, the secondary windings and the e.h.t. winding are situated on one leg of the core. The windings are impregnated in flame retardant polyester, meeting the self-extinguishing requirements of IEC65, para. 14.4 and UL492, para. 280-SE1. The transformer is provided with four mounting pins and two threaded holes for mounting. External circuit connection is made to connecting pins, positioned as indicated in Fig. 1 enabling the unit to be soldered directly into a printed-wiring board (Fig. 2).

## MECHANICAL DATA

# Dimensions (in mm) and terminals

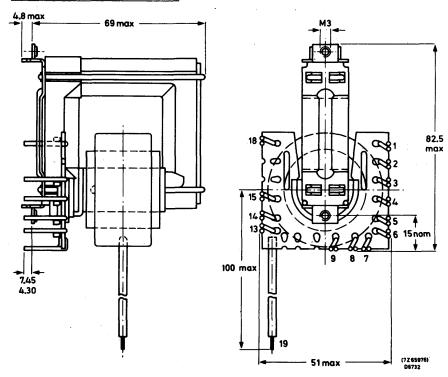


Fig. 1

Weight 240 g approximately

# LINE OUTPUT TRANSFORMER

# MOUNTING

The transformer may be mounted on either a printed-wiring board or, under certain conditions, on a metal chassis. It may be secured with M3 screws.

For mounting on a printed-wiring board the fit of the connecting and the mounting pins in a printed-wiring grid with a pitch of 2.54 mm is illustrated in Fig. 2.

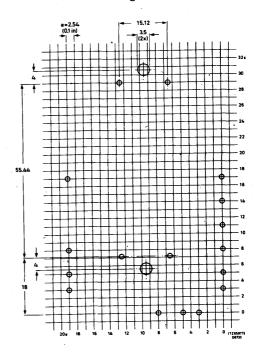


Fig. 2 Hole pattern for mounting on a printed-wiring board (solder side) Grid hole diameter 1. 3 ± 0.1 mm

Whether the transformer is board- or chassis mounted, the core must be earthed.

# Temperature

The operating temperature of the core and the coils should not exceed  $105\ ^{\rm o}{\rm C}$ , under worst conditions.

To satisfy this requirement it may be desired to provide ample cool air circulation around the transformer.

# Distances

The following minimum distances between the transformer and neighbouring conductive flat surfaces (it should be noticed that edges of conductive parts must have a greater distance) must be maintained:

- a. From the e.h.t. winding, radially 15 mm axially 10 mm
- b. From the e.h.t. lead

15 mm

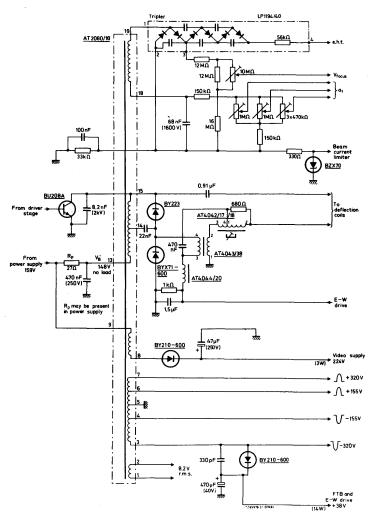
The transformer, and the leads and components carrying high voltage pulses should be kept free from metal particles, solder drops etc.

## ELECTRICAL DATA

I <sub>eht</sub>		0.05	
-ent	mA	0.05	1.5
e.h.t.	kV	24.9	<b>22.</b> 0
R <sub>i(eht)</sub>	МΩ	1	. 8
$v_{\rm B}$	v	159	159
$v_{B'}$	v	148	143
I <sub>average</sub>	mA	400	605
V <sub>CEM</sub>	v	1200	1150
<sup>+I</sup> CM	A	4.0	4.0
I <sub>p-p</sub>	A	6.5	6. 2
flyback ratio (average)	<b>%</b> *	11.6	
Overscan	%		6
Variation	%	≤1.5	
	kV	8.4	7.7
			- "
picture tube heater voltage V <sub>1-2</sub>		8. 2	7.6
Voltages between pin 3 - 5		-320 (+38 V <sub>d</sub> ,	c. )
pin 4 - 5			
pin 6 - 5		+155	
pin 7 - 5		+320	
		+224 (159 + 65	5)
	R <sub>i(eht)</sub> V <sub>B</sub> V <sub>B</sub> V <sub>B</sub> Iaverage  VCEM  +ICM  Ip-p  flyback ratio (average)  Overscan  Variation	Ri(eht)       MΩ         VB       V         VB'       V         Iaverage       mA         VCEM       V         +ICM       A         Ip-p       A         flyback       %         ratio       %         (average)       %         Overscan       %         kV    age V1 - 2 Vp 5 Vp 5 Vp 5	Ri(eht)       MΩ       1         V <sub>B</sub> V       159         V <sub>B</sub> '       V       148         I <sub>average</sub> mA       400         V <sub>CEM</sub> V       1200         +I <sub>CM</sub> A       4.0         I <sub>p-p</sub> A       6.5         flyback ratio (average)       %'       11.6         Overscan       %       11.6         Variation       %       8.4         age       V <sub>1</sub> - 2       V <sub>rms</sub> 8.2         -5       V <sub>p</sub> -320 (+38 V <sub>d</sub> -5       V <sub>p</sub> +155         -5       V <sub>p</sub> +155         -5       V <sub>p</sub> +155         -5       V <sub>p</sub> +320

<sup>\*</sup>Video supply

# APPLICATION CIRCUIT





QUICK REFERENCE DATA				
I <sub>eht</sub>	max. 1.5 mA			
E.H.T.	8.4 kV			
R <sub>i(eht)</sub>	2.0 ΜΩ			
$I_{p-p}$ deflection	6. 4 A			
Load inductance (of line deflection)	1.11 mH			
Supply voltage $(V_B')$ current $(I_B)$ at $I_{eht} = 1.5$ mA	148 V 690 mA			
Voltages of auxiliary windings	-380 $V_p$ , +155 $V$ , -155 $V_p$ , +320 $V_p$ picture tube heater voltage			

## APPLICATION

This transformer has been designed to provide the required scanning amplitude for 20AX 110° colour picture tubes with a neck diameter of 36.5 mm in transistor equipped television receivers presenting 625 lines at 50 fields per second (CCIR) or 525 lines at 60 fields per second (USA).

It is intended for use in conjunction with:

- deflection units AT1080, AT1083/01 or AT1085
- linearity control unit AT4042/38
- line output transistor BU208A
- e.h.t. rectifier/multipliers LP1194/40 or LP1196/40 according to circuit diagram of Fig. 3.

## DESCRIPTION

The magnetic circuit of the transformer comprises a Ferroxcube U- and a Ferroxcube I-core, clamped together with brackets. The primary windings, the secondary windings and the e.h.t. winding are situated on one leg of the core. The windings are impregnated in flame retardant polyester, meeting the self-extinquishing requirements of IEC65, para. 14.4 and UL492, para. 280-SE1. The transformer is provided with four mounting pins and two threaded holes for mounting. External circuit connection is made to connecting pins, positioned as indicated in Fig. 1 enabling the unit to be soldered directly into a printed-wiring board (Fig. 2).

## **MECHANICAL DATA**

# Dimensions (in mm) and terminals

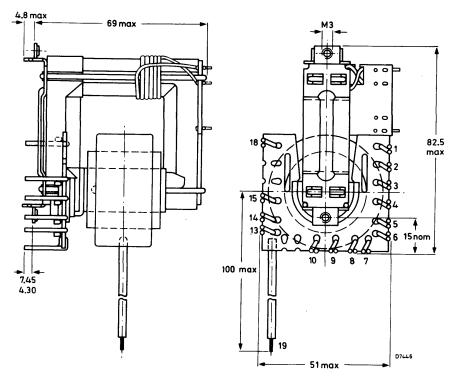


Fig. 1

Weight 240 g approximately

# LINE OUTPUT TRANSFORMER

## MOUNTING

The transformer may be mounted on either a printed-wiring board or, under certain conditions, on a metal chassis. It may be secured with M3 screws.

For mounting on a printed-wiring board the fit of the connecting and the mounting pins in a printed-wiring grid with a pitch of 2.54 mm is illustrated in Fig. 2.

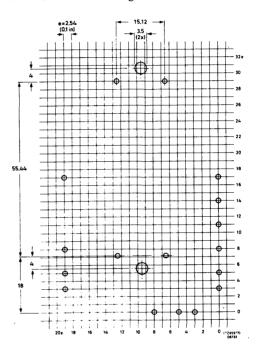


Fig. 2 Hole pattern for mounting on a printed-wiring board (solder side)
Grid hole diameter
1.3 ± 0.1 mm

Whether the transformer is board- or chassis mounted, the core must be earthed.

# Temperature

The operating temperature of the core and the coils should not exceed 105  $^{\rm o}{\rm C},$  under worst conditions.

To satisfy this requirement it may be desired to provide ample cool air circulation around the transformer.

# Distances

The following minimum distances between the transformer and neighbouring conductive flat surfaces (it should be noticed that edges of conductive parts must have a greater distance) must be maintained:

a. From the e.h.t. winding, radially 15 mm axially 10 mm

b. From the e.h.t. lead

15 mm

The transformer, and the leads and components carrying high voltage pulses should be kept free from metal particles, solder drops etc.

# **ELECTRICAL DATA**

E.H.T. supply	I <sub>eht</sub>	mA	0,05		1.5
	e.h.t.	kV	24.9		22.0
	R <sub>i(eht)</sub>	MΩ		2.0	
Power supply	$v_B$	v	159		159
	$V_{\mathbf{B'}}$	v	148		143
	I <sub>average</sub>	mA	400		605
Output transistor	$v_{CEM}$	v	1200		1150
	$^{+I}$ CM	A	4.0		4.0
	-I <sub>CM</sub>	A	2.0		2.0
Deflection	$I_{p-p}$	A	6.4		6.2
	flyback ratio (average)	μs	11.6		
	Overscan	%		6	
	Variation	%		<b>≤1.</b> 5	
Focus voltage		kV	8.4		7.7
Auxiliary windings:					
picture tube heater vo	ltage V <sub>1-2</sub>	$v_{ m rms}$	6.62		6. 18
Voltages between pin	3 - 5	$v_p$	- 350		
pin	4 - 5	$v_p$	+16		
pin	6 - 5	$v_p$	+160		
pin	7 - 5	$v_p$	+350		
pin	10*	v <sub>d.c.</sub>	184 (15	9 + 25 V)	
pin	8*	v <sub>d.c.</sub>	224 (15	9 + 65 V)	

<sup>\*</sup>Video supply

# LINE OUTPUT TRANSFORMER

# APPLICATION CIRCUIT

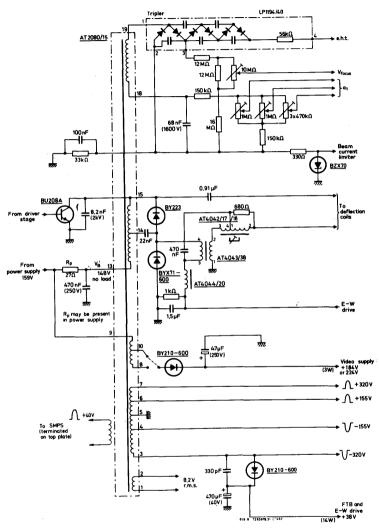


Fig. 3

## APPLICATION

The AT2095 output transformer has been designed for use in non-isolated parallel switched-mode power supplies operating over a range of 16 to 20 kHz.

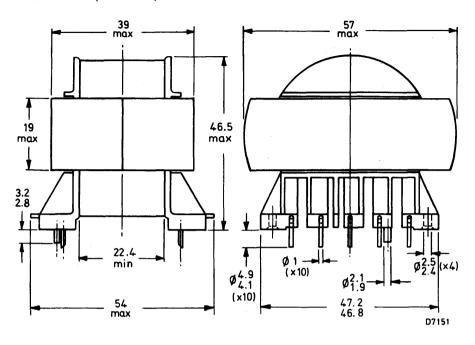
It is used in conjunction with AT4043/03 driver transformer and switched-mode transistor BU126 and control i.c. TDA2640.

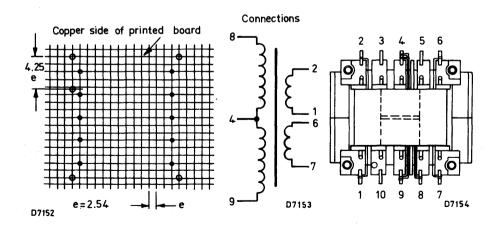
The magnetic circuit of the transformer comprises two Ferroxcube E-cores with an air gap in the centre pole. The unit is provided with pins for mounting on a printed circuit board. (See page 2 for dimensions)

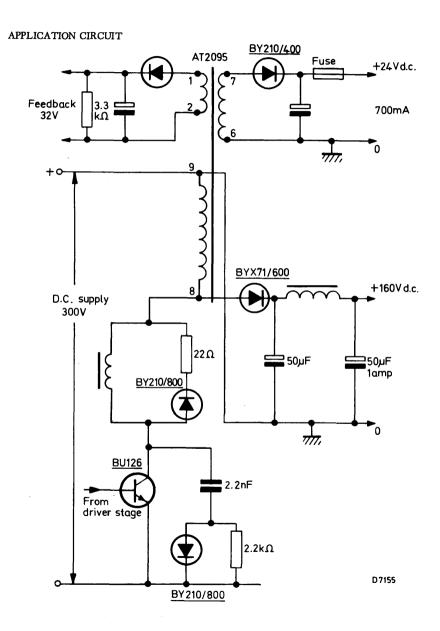
## ELECTRICAL DATA

Inductance	Primary 8-9	5 ± 10%	mH
D.C. Resistance	Primary 8-9	$0.63 \pm 10\%$	Ω
Leakage inductance	1-2 (8-9 Shorted)	≤ 7	μΗ
Turns ratio	Primary 8-9 to Secondary 1-2	5:1 ± 4%	
	Primary 8-9 to Secondary 6-7	6.4:1± 4%	
Absolute maximum operating temperature		115	°C

# DIMENSIONS (millimetres)









## APPLICATION

This unit has been designed for use in colour TV sets equipped with a  $110^{\rm o}$  deflection angle colour picture tube, to adjust the linearity of line deflection. It can be used in combination with the unit AT1080 if parallel connected line coils are used.

#### DESCRIPTION

The unit consists of a coil, mounted on a Ferroxcube rod, two Magnadur magnets and one plastoferrite magnet. One magnet has the shape of a ring and is placed around the Ferroxcube rod above the coils. One has the shape of a half ring and is placed around the Ferroxcube rod under the coils. The third magnet is cylindrical; it is positioned to, and clamped against the Ferroxcube rod opposite the half ring magnet. It is provided with a square hole to facilitate turning to adjust the biasing field and, therefore, the linearity of the line deflection.

# MECHANICAL DATA

# Dimensions in mm

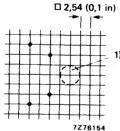
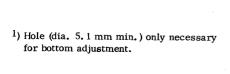
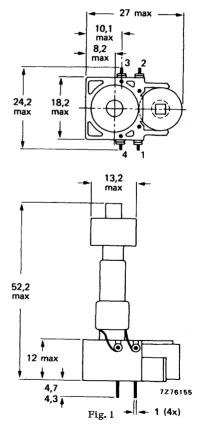


Fig. 2 Hole pattern for mounting on a printed-wiring board. Hole diameter 1. 3 + 0. 1





The correction voltage is pre-adjusted to 23.5 V  $\pm$  2.5% at a saw tooth current of 6.4 A peak-to-peak, frequency 15625 Hz, fly-back ratio 18% (without S-correction), flowing through winding 1-2. The voltage between pins 2 and 3 (pins 1 and 4 interconnected) is then 28.5 V  $\pm$  10%.

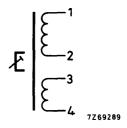


Fig. 3 Circuit diagram

#### MOUNTING

The unit can be mounted on printed-wiring boards by means of its four connection pins (see Fig. 2). To prevent distortion of the magnetic field, no iron part should approach the magnetic parts nearer than 3 mm. The coils should be shunted with a carbon resistor to damp ringing phenomena; the value of resistor depends on applied deflection transformer (typical value  $560~\Omega$  with transformer AT2076/10).

<sup>\*)</sup> Pins 2 and 3 should be interconnected on the printed-wiring board.

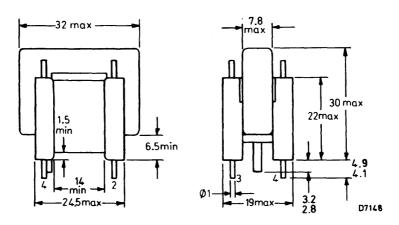
# DRIVER TRANSFORMER for use in switched mode power supplies in television receivers

#### APPLICATION

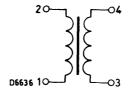
The transformer is used in conjunction with non-isolated transformer AT2095

It uses two Ferroxcube U-cores and it is provided with pins for mounting on a printed wiring board.

#### DIMENSIONS (millimetres)

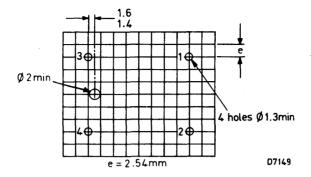


#### CIRCUIT DIAGRAM

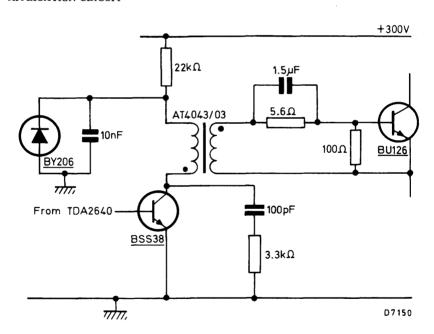


Inductance	Primary (3-4)	≥350	mH
	Secondary (1-2) (with (3-4) Shorted)	< 5	μΗ
Absolute maximum operating temperature		115	°С
Turns ratio	21.	74 ± 4%	

#### RECOMMENDED PIERCING DIAGRAM



#### APPLICATION CIRCUIT



# LINE DRIVER TRANSFORMER

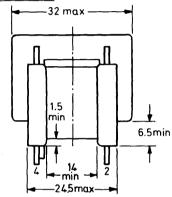
#### APPLICATION

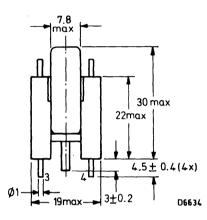
The transformer AT4043/29 has been designed for all-transistor colour television sets. It can be used in the single-transistor (BU208) line-output circuit in conjunction with the line-output transformer AT2063/03 and AT2080/...

#### MECHANICA L DATA

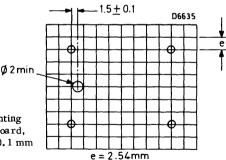
The magnetic circuit of the transformer comprises two ferroxcube U-cores. The unit is provided with pins for mounting on a printed-wiring board.

#### Dimensions (mm)









Hole pattern for mounting on a printed-wiring board, hole diameter 1.3 + 0.1 mm

#### E LECTRICA L DATA

Inductance (primary)

Maximum working temperature

370 mH ± 12%

D6636 10

#### APPLICATION

This transformer has been designed for all-transistor or transistor/thyristor colour television sets. It is intended to be used in conjunction with the deflection unit AT 1080, which is provided with a 4-pole unit for equalisation of the line and field astigmatism (see also data sheet of AT 1080).

#### MECHANICAL DATA

The magnetic circuit of the transformer comprises two Ferroxcube cores, an E-and an I-core. The unit is provided with pins for mounting on a printed-wiring board.

#### Dimensions (mm)

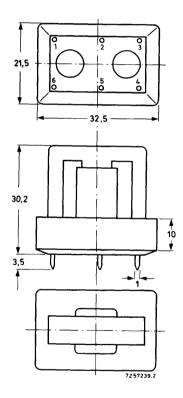


Fig. 1

#### Mounting

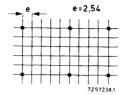


Fig. 2 Hole pattern for mounting on a printed-wiring board, hole diameter 1,3+0,1 mm.

#### ELECTRICAL DATA

Inductance between 1 and 6, 3 and 4 interconnected

Maximum working temperature

 $7,3 \text{ mH} \pm 10\%$ 

105 °C



Fig. 3 Connection diagram

### Application circuit

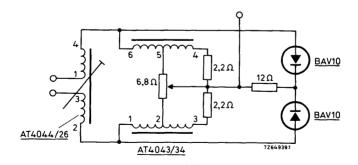


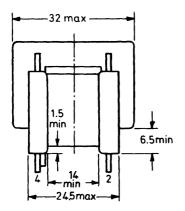
Fig. 4

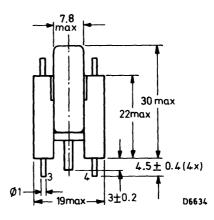
# BRIDGE COIL for colour television receivers

#### APPLICATION

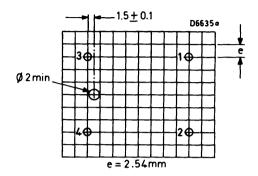
For use as a bridge transformer in the line output transformer circuit of the AT2080/10, in conjunction with the deflection unit AT1080 (see also data sheet AT2080/10). The magnetic circuit of the coil comprises two Ferroxcube U-cores. The unit is provided with pins for mounting on a printed-wiring board.

#### Dimensions (millimetres)





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Hole pattern for mounting on a printed-wiring board. Hole diameter 1.3 min. e = 2.54 mm (0.1 in).

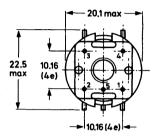
Inductance (primary 1-2)	425 $\mu$ H $\pm$ 10 $\%$	
Resistance (primary 1-2)	< 0.4 Ω	20
Maximum voltage, peak-to-peak	400 V	<b>₹ </b> }
Maximum current, peak-to-peak	6.7 A	315
Maximum current, r.m.s.	1.8 A	)IC
Maximum working temperature	100 °C	D6636 1003

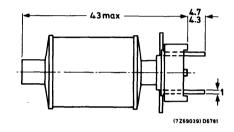
# EAST-WEST LOADING COIL

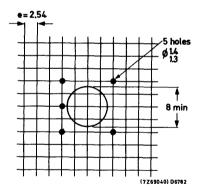
#### GENERAL

Designed for use with line output transformer AT2080 and deflection unit AT1080. The coil is provided with pins for mounting on a printed wiring board, and it can be adjusted by means of a trimming key.

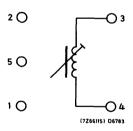
#### DIMENSIONS (millimetres)







Recommended piercing diagram



Circuit diagram

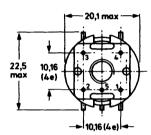
Inductance between 3 and 4 (measured with 5000 pF in parallel	1 to 5.3	mH
Resistance between 3 and 4	2	Ω
MAXIMUM WORKING TEMPERATURE	95	$^{\mathrm{o}}\mathrm{C}$

#### APPLICATION

This coil has been designed for the circuitry of the four-pole unit incorporated in the deflection unit AT1080, for equalization of line and field astigmatism (see also data on AT1080).

#### **MECHANICAL DATA**

Dimensions (mm)



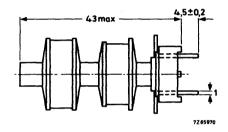


Fig. 1

The coil is provided with pins for mounting on a printed-wiring board. It can be adjusted at the top by means of a trimming key.

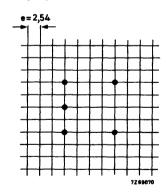
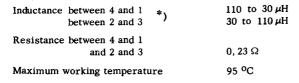


Fig. 2 Hole pattern for mounting on a printed-wiring board, hole dia 1,3+0,1 mm



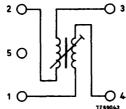


Fig. 3 Connection diagram

Pins 2 and 4 should be interconnected.

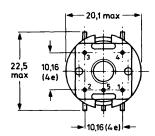
<sup>\*)</sup> measured with 5000 pF in parallel.

#### APPLICATION

This correction coil has been designed for the circuitry of the four-pole unit incorporated in the deflection unit AT1080, for equalization of line astigmatism (see also data on AT1080).

#### MECHANICAL DATA

#### Dimensions (mm)



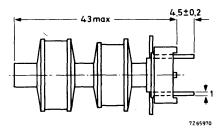


Fig.1

The coil is provided with pins for mounting on a printed-wiring board. It can be adjusted at the top by means of a trimming key.

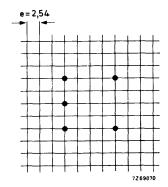


Fig. 2 Hole pattern for mounting on a printed-wiring board, hole dia 1, 3 + 0, 1 mm

Inductance, measured	with 5000 pF in p	arallel
between 3 and 5	*)	33 to 150 µH
between 4 and 5	*)	150 to 33 μH
Resistance at 25 °C		
between 1 and 2		$0,23~\Omega$
between 3 and 4		$0,18~\Omega$
Maximum working ten	nnerature	95 °C

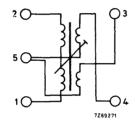
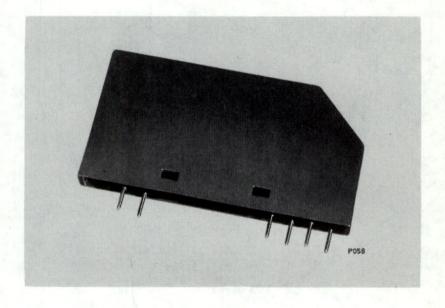


Fig. 3. Connection diagram

<sup>\*)</sup> Supplied with core position for  $L_{3-5} = L_{5-4} = 11.3 \mu H \pm 5\%$ .

QUICK REFERENC	E DATA		
For receivers up to European PAL standard.			
Nominal frequency	4.433619	MHz	
Phase delay time	63.943	μs	
Dimensions	71 x 7.5 x 38	mm	
Self-extinguíshing properties			

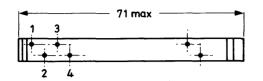


#### APPLICATION

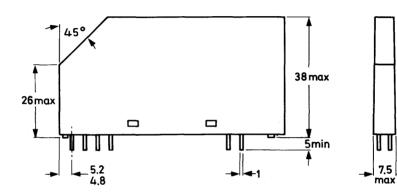
The DL50 is intended for use in decoder circuits of colour television receivers.

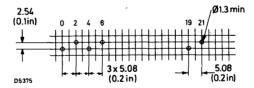
#### DESCRIPTION

A very thin slab of zero TC glass, provided with two transducers, is shock proof mounted in a suitable housing that complies with the self extinguishing and non-dripping properties of the American Underwriters Laboratories rating mentioned in UL94 SE-1. Six pins enable the unit to be soldered directly into a printed wiring board. Input and output coil are not to be included.



D5374





Piercing diagram

Measured at 25°C according to the m	neasuring circuit of Fig. 3.	
Nominal frequency (f nom)	4.433619	MHz
Phase delay time $(\tau)$ between $V_1$ and $V_2$ at $f_{nom}$ (unmodulated)		
sinewave voltage	63.943 ± 0.005	μs
Bandwidth at -3 dB	from $< 3.43$ to $> 5.23$	MHz
Insertion loss at f	8 ± 3	dB
Drift of phase delay with temperature (relative to 25°C)	max. 5, typ. 3	ns
Maximum input voltage at f	15	v
Unwanted reflections at $3\tau$	$\leq$ -22 with respect to $1\tau$ signal	V p-p dB
Other reflections	$\leq$ -30 with respect to $1\tau$ signal	dB
Operating temperature range	<b>-</b> 20 to +70	°C
Nominal terminations at f		
R1, R2 termination resistance	390	Ω
C1 total capacitance	120	pF
L1 inductive reactance	128	Ω
L2 inductive reactance	231	Ω

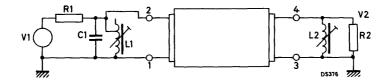


Fig. 3

Recommended adjustment range of the coils	-19 to +36	%
Maximum capacitance of the coils	20	pF



QUICK REFERENCE DATA			
For receivers to European P.	AL/SECAM standard		
Nominal frequency	4.433619 MHz		
Phase delay time	$63.943  \mu s$		
Dimensions	71 x 7, 5 x 37, 5 mm		
Self-extinguishing properties			

#### APPLICATION

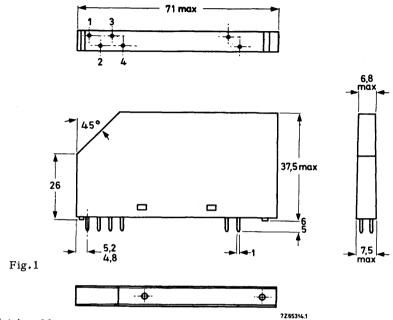
The DL51 is intended for use in decoder circuits of colour television receivers.

#### DESCRIPTION

A very thin slab of zero TC glass provided with two transducers is mounted shock proof in a housing, that complies with the self-extinguishing and non-dripping properties of the American Underwriters' Laboratories rating mentioned in UL94 SE-1 Six pins enable the unit to be soldered directly into a printed-wiring board. Input and output coil are not included.

#### **MECHANICAL DATA**

## Dimensions (mm)



Weight 16 g

# Mounting

The unit can be soldered directly into a printed-wiring board.

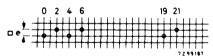


Fig. 2.

Recommended hole pattern for mounting on a printed-wiring board. e = 2,54 mm. The tolerance on the distances of the different holes to the 0-line is  $\pm 0$ ,1 mm. Hole diameter is 1,3  $\pm$  0,1 mm.

Measured according to the circuit of Fig.3 at 25  $^{o}\text{C}$  and  $f_{nom}$  (unless otherwise specified)

Nominal frequency (fnom)

4.433619 MHz

Phase delay time (7) between V<sub>1</sub> and V<sub>2</sub> (unmodulated sinewave voltage)

 $63.943 \pm 0.005 \,\mu s$ 

Bandwidth at -3 dB

from  $\leq 3.43$  to  $\geq 5.23$  MHz

Insertion loss

 $8 \pm 3 \, dB$ 

Drift of phase delay from +10 to +60 °C (relative to 25 °C)

max. 5 ns, typ.  $3 \text{ ns}^{1}$ )

Maximum input voltage (p-p)

15.V

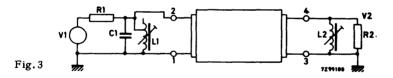
Unwanted reflections,

3 τ other reflections  $\leq$  -22 dB with respect to  $1 \tau$  signal  $\leq$  -35 dB with respect to  $1 \tau$  signal

Operating temperature range

-20 to +70 °C

#### Measuring circuit



#### Nominal terminations

R1, R2	termination resistance	390 Ω
Cl	total capacitance	120 pF
L1	inductive reactance	128 Ω
L2	inductive reactance	231 Ω

Recommended adjustment range of the coils

- 19 to + 36 %

Maximum capacitance of the coils

20 pF

QUICK REFERENCE DATA			
For receivers up to European PAL standard			
Nominal frequency	4,433619 MHz		
Phase delay time	63,943 μs		
Dimensions	37 x 7, 5 x 28, 5 mm		
Self-extinguishing properties			

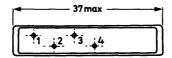
#### **APPLICATION**

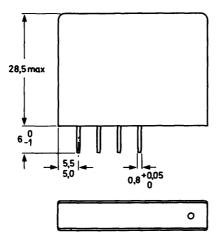
The DL60 is intended for use in decoder circuits of colour television receivers.

#### DESCRIPTION

A very thin slab of zero TC glass provided with two transducers is shock-proof mounted in a housing that satisfies the flame test described in IEC 50 C (secretariat) 11. Four pins enable the unit to be soldered directly onto a printed-wiring board.

#### Outlines





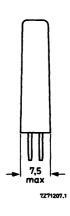


Fig. 1

Mass

7 g

#### Mounting

The unit can be soldered directly onto a printed-wiring board.

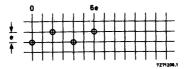


Fig. 2.

Recommended hole pattern for mounting on a printed-wiring board; e=2,54 mm. The tolerance on the distances of the different holes to the 0-line is  $\pm\,0,1$  mm. Hole diameter is 1,0+0,1 mm.

Measured with the circuit of Fig. 3 at 25  $^{\rm oC}$  and  ${\rm f_0}$  (unless otherwise specified)

Nominal frequency (f<sub>0</sub>)

Phase delay time  $(\tau)$ 

Bandwidth at -3 dB

Insertion loss

Drift of phase delay from + 10 to +60 °C (relative to +25 °C)

Maximum input voltage (p-p)

Spurious signals

 $3\tau$  signals other signals

Phase relation  $\varphi_{4-3} - \varphi_{2-1}$ 

Storage temperature range

4,433619 MHz

 $63,943 \pm 0,005 \mu s$ 

from  $\leq 3.43$  to  $\geq 5.23$  MHz

 $9 \pm 3 dB$ 

max. 5 ns, typ. 3 ns

10 V

 $\leq$  -22 dB with respect to 1  $\tau$  signal

≤ -30 dB with respect to 1  $\tau$  signal

180°

-40 to +70 °C

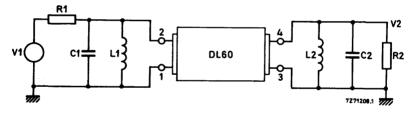


Fig. 3.

#### **Terminations**

 $R1 = R2 = 560 \Omega$ 

C1 = 20 pF

C2 = 30 pF

 $L1 = 10,5 \, \mu H$ 

 $L2 = 9.7 \, \mu H$ 

total capacitance of test jig without delay-line i.e. wiring capacitance, capacitance of coil and extra trimming capacitor.

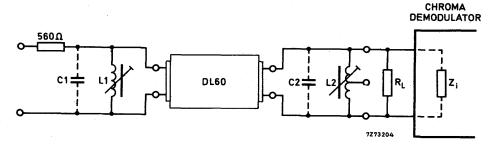


Fig. 4

$$(R_L//Z_i) = 560 \Omega$$

C1, C2 < 30 pF (wiring capacitance and capacitance of the coil)

L1, L2 nominal values depend on values of C1 and C2 to produce the reactances:

$$X1 = \frac{\omega_{0} \text{ L1}}{1 - \omega_{0}^{2} \text{ L1C1}} = 350 \Omega$$

$$X2 = \frac{\omega_{0} \text{ L2}}{1 - \omega_{0}^{2} \text{ L2C2}} = 350 \Omega$$

$$f_{0} = 4,433619 \text{ MHz}$$

Maximum bandwidth is obtained at minimum C1 and C2.

Recommended adjustment range of the coils -19 to +36%.

#### **DEVELOPMENT SAMPLE DATA**

#### **QUICK REFERENCE DATA**

For receivers up to European PAL standard

Nominal frequency

4.433619 MHz

Phase delay time

63.943 µs

Dimensions

 $37 \times 7.5 \times 28.5 \text{ mm}$ 

Self-extinguishing properties

#### APPLICATION

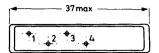
The DL700 is intended for use in decoder circuits of colour television receivers. It is physically interchangeable with the DL50 and DL60  $\,$ 

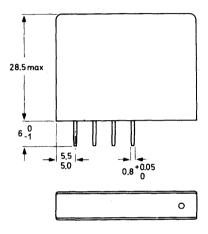
#### DESCRIPTION

A very thin slab of zero TC glass provided with two transducers is shock-proof mounted in a housing that satisfies the flame test described in IEC 50 C. Four pins enable the unit to be soldered directly onto a printed-wiring board.

This information is derived from development samples made available for evaluation, it does not form part of our data is-induction system and does not necessarily imply that the device will go into production







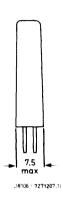


Fig. 1

Mass

7 g

#### Mounting

The unit can be soldered directly onto a printed-wiring board.

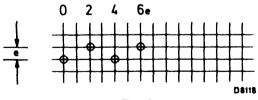


Fig. 2

Recommended hole pattern for mounting on a printed-wiring board: e = 2.54 mm The tolerance on the distances of the different holes to the 0-line is  $\pm$  0.1 mm. Hole diameter is 1.0  $\pm$  0.1 mm

Measured with the circuit of Fig. 3 at 25 °C and  $\boldsymbol{f}_{o}$  (unless otherwise specified)

Nominal frequency (fo)

Phase delay time  $(\tau)$ 

Bandwidth at -3 dB

Insertion loss

Drift of phase delay from +10 to +60 °C (relative to +25 °C)

Maximum input voltage (p-p)

Spurious signals

3 τ signals other signals

Phase relation  $\phi_{4-3}$  -  $\phi_{2-1}$ 

Storage temperature range

4.433619 MHz

63.943 ± 0.005 µs

from  $\leq$  3.43 to  $\geq$  5.23 MHz

 $9 \pm 3 dB$ 

max. 5 ns, typ. 3 ns

10 V

 $\leq$  -22 dB with respect to 1  $\tau$  signal  $\leq$  -30 dB with respect to 1  $\tau$  signal

180°

-40 to +70 °C

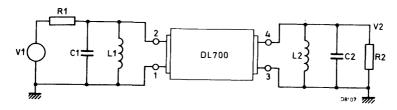


Fig. 3

Terminations

 $R1 = R2 = 390 \Omega$ 

C1 = 20 pF

C2 = 30 pF

 $L1 = 8.64 \mu H$ 

 $L2 = 8.10 \mu H$ 

total capacitance of test jig without delay-line i.e. wiring capacitance, capacitance of coil and extra trimming capacitor.

**DEVELOPMENT SAMPLE DATA** 

Mullard

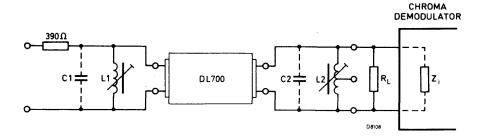


Fig. 4

$$(R_L//Z_i) = 390 \Omega$$

C1, C2 < 30 pF (wiring capacitance and capacitance of the coil)

L1, L2 nominal values depend on values of C1 and C2 to produce the reactances:

$$X1 = \frac{\omega_0 \text{ L1}}{1 - \omega_0^2 \text{ L1C1}} = 278 \,\Omega$$

$$X2 = \frac{\omega_0 L2}{1 - \omega_0^2 L2C2} = 278 \Omega$$

$$f_0 = 4.433619 \text{ MHz}$$

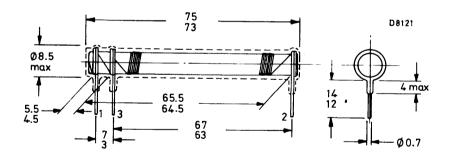
Maximum bandwidth is obtained at minimum C1 and C2.

Recommended adjustment range of the coils -19 to +36%.

The VS series of luminance delay lines are designed for delaying the luminance signals at video frequencies in colour television receivers. Delay times in the range of  $t_d$  = 340 to 600 ns can be supplied to relate to the various set circuit and transmitter encodings. To cover this range the number of turns on the winding is varied and two different ceramic materials are employed. Therefore the characteristics of the delay lines do not vary uniformly with the delay time. The compact outline complies with modern requirements for miniaturisation and modular construction. The cylindrical ceramic body has fired-on silver lacquer tracks and the encapsulation ensures adequate protection against humidity and temperature as well as mechanical damage.

QUICK REFERENCE DATA				
VS340/1			340	ns
VS400/1			400	ns
VS470/1	delay times	typ.	470	ns
VS550/1			550	ns
VS600/1			600	ns
Insertion loss	s	typ.	1	dB
Reflection co	efficient	<	2	%

#### **OUTLINE AND DIMENSIONS (millimetres)**



weight 7 g

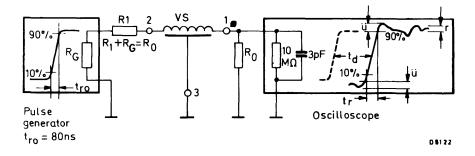
RATINGS Limiting values of operation according to the Absolute Maximum System.

Input voltage	$v_{1-3}$	100	V
Input current	I <sub>1-2</sub>	30	mA
Operating temperature	$T_{amb}$	-20 to +80	٥C
Storage temperature	${\sf T}_{\sf stg}$	-20 to +100	°C

#### CHARACTERISTICS

The characteristics for the delay time  $(t_d)$ , rise time  $(t_r)$ , overshoot  $(\ddot{u})$  and reflection (r) are measured in the test circuit shown below.

#### Test circuit



#### Delay time

The delay time  $t_d$  will be within  $\pm$  10% of the nominal value (max.  $\pm$  50 ns)

#### Reflection

With terminations of the required standard resistance ( $Z=f(t_d)$  the reflection is typically less than 2%. This value applies to all types in the range.

#### Insertion loss

The insertion loss at low frequencies (f <<<8) measured assuming that  $R_1$  =  $R_0$  = Z, increases approximately with the delay time and has a maximum value of 2 dB.

# **LUMINANCE DELAY LINES**

# VS340/I VS470/I VS400/I VS550/I VS600/I

Type No.	Delay time ns	Impedance Z	Bandwidth B		Reflection r	
			typ. MHz	min. MHz	typ. %	max. %
VS340/1	340	1000	8.0	7.0	1.8	4.0
VS400/1	400	1200	7.0	6.0	1.8	4.0
VS470/1	470	1300	7.0	6.0	1.8	4.0
VS550/1	550	820	4.2	3. 8	1.8	4.0
VS600/1	600	910	4.2	3. 8	1.8	4.0

# QUARTZ CRYSTAL UNIT

#### **QUICK REFERENCE DATA**

Nominal frequency	8867, 238 Hz
Mode of vibration	fundamental
Type of holder	RW-10

#### **APPLICATION**

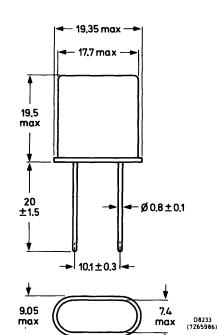
Intended to be used in the sub-carrier oscillator of colour television sets according to the PAL system.

#### DESCRIPTION

The unit consists of a metal-plated AT-cut quartz plate, mounted in a resistance welded metal holder, provided with two connecting leads.

#### **MECHANICAL DATA**

Outline



Dimensions in mm

#### MARKING

The holder is marked as follows

Frequency in kHz

5 digit code number (0312) Date code (year/week)

Note: The last digit of the code number signifies the factory production specification and should not be used for ordering purposes.

**MASS** 

approx. 2 g



Mullard

### **ELECTRICAL DATA\***

Unless otherwise specified the values apply at a temperature of 25 ± 2 °C

Load resonance frequency, load capacitance 20 pF drive level 0.5 mW		8867, 238	kHz
Adjustment tolerance	±max.	$40 \times 10^{-6}$	
Tolerance over the temperature range of +10 to +60 °C, with respect to +25 °C	±max.	30 × 10 <sup>-6</sup>	
Trimability at a load capacitance of 20 pF with a load capacitance variation of 10 pF	min.	950	Hz
Motional capacitance (C <sub>1</sub> )	typ.	21	fF
Parallel capacitance (C <sub>0</sub> )	max. typ.	6 5	pF pF
Resonance resistance in temperature range of +10 to +60 °C	typ. max.	15 60	$\Omega$
Maximum permissible d.c. voltage between terminations		100	٧
Operating temperature range		+10 to +60	οС

<sup>\*</sup>The terminology of IEC document 49 (secretariat) 76 is used.

### **TESTS AND REQUIREMENTS**

Essentially the following tests mentioned in the schedule of IEC publication 122 are carried out along the lines of IEC publication 68.

IEC 122 clause	IEC 68-2 test method	test	procedure	requirements
2.5.17	_	Aging	30 days, +85 °C	$\Delta f/f \pm max.$ 15 ppm
2.5.12	Db	Damp heat accelerated	1 day, +55 °C 100% R.H.	$\Delta f/f \pm$ max. 10 ppm R $_{ins}$ at 50 V d.c. min. 20 M $\Omega$
	Na	Rapid change of temperature	-20/+50 °C 15 cycles 1 h per cycle	∆f/f ± max. 5 ppm
2.5.2	Ea	Shock	40 g, sawtooth 6 directions, 1 blow per direction	$\Delta f/f \pm max. 5 ppm$ $\Delta R/R \pm max. 15%$
2.5.3	Fc	Vibration	10-55-10 Hz, 0.75 mm displacement 2 h, 3 directions*	Δf/f ± max. 5 ppm ΔR/R ± max. 15%
2.5.6	Ub	Flexibility of terminations	1 x 90°, 5 N	no visible damage
2.5.10	Т	Soldering	300 °C, 2 s	Δf/f ± max. 2 ppm good tinning no visible damage

<sup>\*</sup>The batch is divided into three equal parts, each part is tested in one of the three perpendicular directions.

### **ORDERING**

Crystals should be ordered using the full catalogue number e.g. 4322 143 03120.





## QUARTZ CRYSTAL UNIT

### **QUICK REFERENCE DATA**

Nominal frequency	4433,619 kHz
Mode of vibration	fundamental
Type of holder	RW-10

### APPLICATION

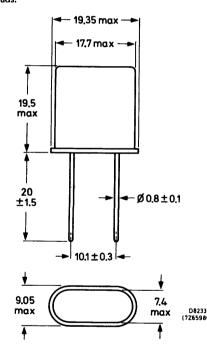
Intended to be used in the sub-carrier oscillator of colour television sets according to the PAL system.

### DESCRIPTION

The unit consists of a metal-plated AT-cut quartz plate, mounted in a resistance welded metal holder, provided with two connecting leads.

### **MECHANICAL DATA**

Outline



Dimensions in mm

### MARKING

The holder is marked as follows

Frequency in kHz

5 digit code number (0110).

Date code (year/week)

Note: The last digit of the code number signifies the factory production specification and should not be used for ordering purposes.

MASS

approx. 2 g





## **ELECTRICAL DATA\***

Unless otherwise specified the values apply at a temperature of 25  $\pm$  2  $^{\rm o}{\rm C}$ 

Load resonance frequency,			
load capacitance 20 pF			
drive level 0.5 mW		4433,619	kHz
Adjustment tolerance	±max.	40 x 10 <sup>-6</sup>	
Tolerance over the temperature range			
of +10 to +60 °C, with respect to +25 °C	±max.	30 x 10 <sup>-6</sup>	
Trimability at a load capacitance of 20 pF			
with a load capacitance variation of 10 pF	min.	600	Hz
•		29	fF
Motional capacitance (C <sub>1</sub> )	typ.	29	I.F.
Parallel capacitance (C <sub>O</sub> )	max.	- 7	рF
•	typ.	6.5	рF
Resonance resistance	typ.	15	Ω
in temperature range of +10 to +60 °C	max.	50	Ω
Maximum permissible d.c. voltage			
between terminations		100	V
Operating temperature range		+10 to +60	оC

**Mullard** 

<sup>\*</sup>The terminology of IEC document 49 (secretariat) 76 is used.

## **TESTS AND REQUIREMENTS**

Essentially the following tests mentioned in the schedule of IEC publication 122 are carried out along the lines of IEC publication 68.

IEC 122 clause	IEC 68-2 test method	test	procedure	requirements
2.5.17	-	Aging	30 days, +85 °C	Δf/f max. 15 ppm
2.5.12	Db	Damp heat accelerated	1 day, +55 °C 100% R.H.	$\Delta f/f \pm max.$ 10 ppm $R_{ins}$ at 50 V d.c. min. 20 M $\Omega$
!	Na	Rapid change of temperature	-20/+50 °C 15 cycles 1 h per cycle	Δf/f ± max. 5 ppm
2.5.2	Ea	Shock	40 g, sawtooth 6 directions, 1 blow per direction	$\Delta f/f \pm \text{ max. 5 ppm}$ $\Delta R/R \pm \text{ max. 15}\%$
2.5.3	Fc	Vibration	10-55-10 Hz, 0.75 mm displacement 2 h, 3 directions*	$\Delta$ f/f max. 5 ppm $\Delta$ R/R ± max. 15%
2.5.6	Ub	Flexibility of terminations	1 x 90°, 5 N	no visible damage
2.5.10	Т	Soldering	300 °C, 2 s	$\Delta f/f \pm max$ . 2 ppm good tinning no visible damage

<sup>\*</sup>The batch is divided into three equal parts, each part is tested in one of the three perpendicular directions.

## **ORDERING**

Crystals should be ordered using the full catalogue number e.g. 4322 152 01100.



# **MODULES**

E

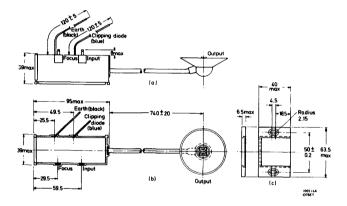


### **DEVELOPMENT SAMPLE DATA**

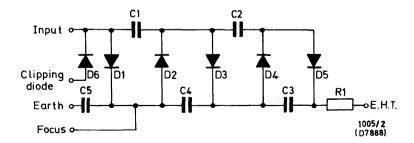
Voltage tripler designed for domestic high performance colour television receivers. The electrical operation is entirely conventional and it can be used in a standard system. However, every aspect of component design and assembly has been re-assessed and only the optimum solutions have been adopted in order that the tripler can be characterised by its intrinsic reliability.

QUICK REFER	RENCE DATA	
V <sub>in</sub> (peak-to-peak)	8. 3	kV
V <sub>out</sub> (e.h.t. supply) (d.c.)	25	kV
V <sub>out</sub> (focus supply) (d.c.)	8. 3	kV

### DIMENSIONS (millimetres)



### CIRCUIT DIAGRAM



This information is derived from development samples made available for evaluation. It does not form part of our data handbook system and does not necessarily imply that the device will go into production

## RATINGS (Limiting values according to the Absolute Maximum System)

V <sub>in</sub> (peak-to-peak)	10.6	kV
$V_{out} (I_{out} = 0)$	30	kV
Focus to e.h.t. voltage	20	kV
I <sub>out</sub> (e.h.t.)	1.7	mA
I <sub>out</sub> (clipping diode)*	4.0	mA
Focus current	0.4	mA
T <sub>amb</sub> (tripler)	70	oC.
T <sub>sto</sub>	-25 to +70	°C

<sup>\*</sup>Clipping diode current is the sum of the clipping diode load plus focus and e.h.t. currents.

## TYPICAL OPERATING CONDITIONS

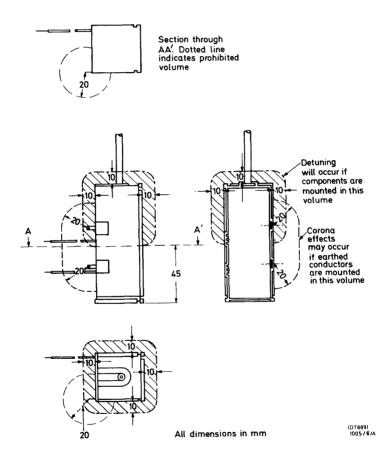
V <sub>in</sub> (peak-to-peak)	8. 3	kV
Vout (e.h.t. supply) (d.c.)	25	kV
Vout (focus supply (d.c.)	8.3	kV
Iout (e.h.t. supply) (d.c.)	1.0	mA
Iout (focus supply) (d.c.)	0. 25	mA
Internal impedance (I <sub>out</sub> = 0.1 to 1.5 mA)	<0.5	$M\Omega$
Surge limiting resistor	47	kΩ
Input capacitance	$9 \pm 1.0$	pF

## VOLTAGE MULTIPLYING MODULE

### MOUNTING PRECAUTIONS

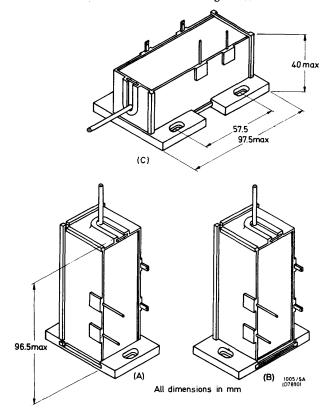
Sufficient clearance must be maintained around the tripler body to prevent corona and detuning as shown below.

Components or other conducting parts shall not be introduced within the volume specified.



### MOUNTING

Is intended to be versatile in either of the two mounting faces.



### PRODUCT SAFETY

For the safety of operators it is recommended that precautions should be taken to prevent personnel from coming within 2 inches of the tripler body or its leads and connections unless the device is not energised and is fully discharged. The module is sold in compliance with a valid BS415 certificate.

## **VOLTAGE MULTIPLYING MODULES**



AVAILABLE FOR CURRENT PRODUCTION; NOT INTENDED FOR NEW DESIGNS

The modules generate e.h.t. and focus voltage from line output pulses to supply colour picture tubes. To be used in conjunction with AT2055 or AT2056 line output transformers.

	QUICK REFERE	NCE DATA	
v <sub>in</sub>	(peak-to-peak)	8.7	kV
$v_{out}$	(e.h.t. supply) (d.c.)	25	kV
$v_{\text{out}}$	(focus supply) (d.c.)	7.7	kV
I <sub>out</sub>	(e.h.t. supply) (average)	1	mA
I <sub>out</sub>	(focus supply) (average)	100	$\mu$ A

### MOUNTING PRECAUTIONS

A separation of at least 15 mm between any part of the main module body or its leads and any metal parts of the receiver is essential to avoid any capacitive discharge current and detuning effects of the line output transformer. Mounting brackets must be confined to the module mounting flanges.

### TYPE NUMBER DESIGNATION

Type No		Numl	per of
Type No.	capacitors	diodes	surge limiting resistors
LP1174/1x	4	5	None
LP1174/3x	5	5	1
LP1174/4x	5	6	1

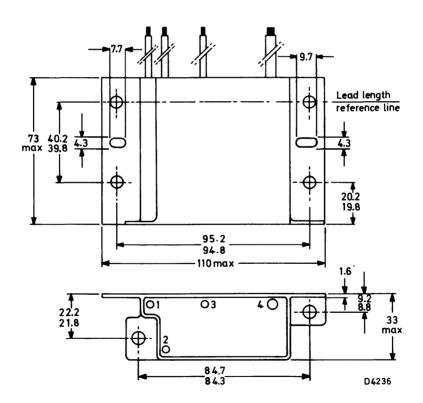
x = variations in lead lengths and terminations.

### CASE AND TERMINATIONS

The modules are encapsulated in a flame retardant plastic case. Connections are by means of flying leads (termination No. 1 can also be supplied with a pin) which are flame retardant to IEC197.

Termination No.	Function	Type of termination
1	Input	Flying lead or pin
2	Earth	Flying lead
3	Focus voltage	Flying lead
4	E.H.T. output	Flying lead

### DIMENSIONS (millimetres)



## VOLTAGE MULTIPLYING MODULES

LP1174 Series

### E LECTRICAL DATA

Typical	operation	conditions

v <sub>in</sub> (peak-to-peak) 1)	8.7	kV
Vout (e.h.t. supply) (d.c.)	25	kV
Vout (focus supply) (d.c.)	7.7	kV
I <sub>out</sub> (e.h.t. supply)	1	mA
I <sub>out</sub> (focus supply)	100	μΑ
E.H.T. regulation (0 to 1.5 mA)	1.3	$M\Omega$

Limiting values - these are the absolute operating limits which must not be exceeded under any conditions.

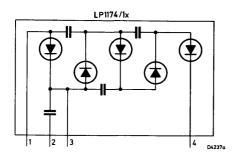
vin (peak-to-peak)	10.5	kV
V <sub>out</sub> (e.h.t. supply) (d.c.) <sup>2</sup> )	31	kV
I <sub>out</sub> (e.h.t. + focus)	2	m A
$T_{amb}$	60	°C

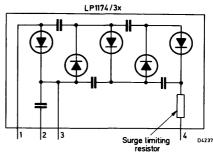
### CIRCUIT DIAGRAMS

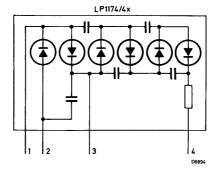
See next page

<sup>1)</sup> Maximum pulse duration 18% of one cycle.

<sup>2)</sup> This does not imply that the voltage rating for the final anode of the picture tube may be exceeded.







OUTCK	REFERENCE	DATA
QUIUN	KELEKENCE	D11111

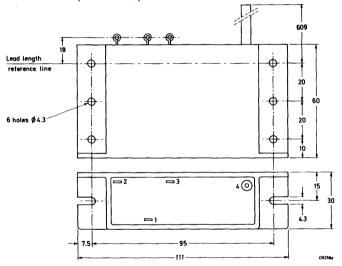
The modules generate e.h.t. and focus voltages from line time base pulses, to supply colour picture tubes.

	Circuit A	Circuit B	
V (peak-to-peak)	8.3	8.6	kV
Vout (e.h.t. supply) (d.c.)	25	25	kV
V (focus supply) (d.c.)	8.3	7.7	kV

### TYPE NUMBERS

LP1194/30 - A five capacitor, five diode module with surge limiting resistor. LP1194/40 - As LP1194/30 with a clipping diode across the input.

### DIMENSIONS (millimetres)



#### Terminations:

- 1 = Input
- 2 = Earth
- 3 = Focus
- 4 = Output

Fig. 1

### MOUNTING PRECAUTIONS

A separation of at least 15mm between any part of the main module body or its lead and any metal parts of the receiver is essential to avoid any capacitive discharge current and detuning effects of the line output transformer. Mounting brackets must be confined to the module mounting flanges.

### E LECTRICAL DATA

Typical operating conditions (where used in typical application circuits A or B)

	Circuit A	Circuit B	
V (peak-to-peak)	8.3	8.6	kV
V <sub>out</sub> (e.h.t. supply) (d.c.)	25	25	kV
I <sub>out</sub> (e.h.t.) (d.c.)	1	1	mA
I (focus) (d.c.)	0.4	0.1	mA
E.H.T. regulation (0 to 1.5mA)	2	2.4	$M\Omega$

## Limiting values

These are absolute operating conditions which must not be exceeded under any condition.

V (peak-to-peak)	10.4	10.8	kV
V <sub>out</sub> (e.h.t.)	31.2	31.5	kV
I (clipping diode)	2.5	2.5	mA
T amb	60	60	$^{\circ}$ C

For other limiting values see figures 2 and 3.

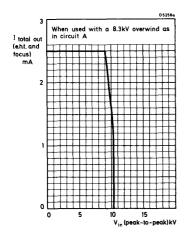


Fig. 2

Fig. 3

### TYPICAL APPLICATION CIRCUITS

### Circuit A

With this arrangement the  $\rm A_1$  diode may be omitted. This configuration is shown for  $110^{\rm O}$  operation.

The beam circuit limiting components may be removed, if not required, and point B connected to earth.

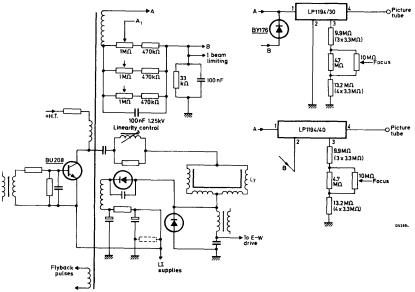


Fig. 4

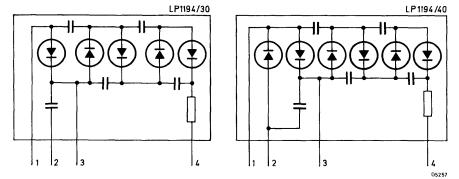


Fig. 5

### Circuit B

For a 7kV overwind, connected to the primary. This configuration is used mainly in  $90^{\circ}$  time bases and the circuit shown is for  $90^{\circ}$  operation.

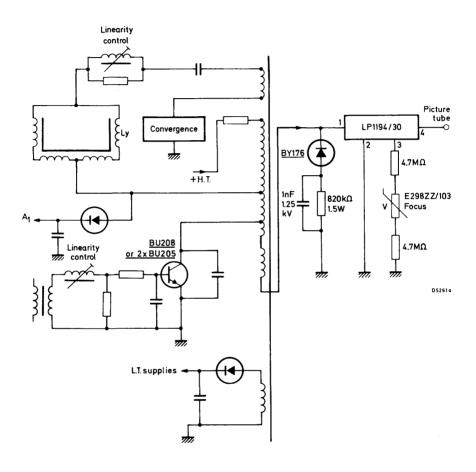


Fig. 6

Both configurations shown in Figs. 4 and 6 may of course be used in either  $90^{\circ}$  or  $110^{\circ}$  operation. The connections between the overwind and tripler will remain the same for both deflection angles.

## VOLTAGE MULTIPLYING MODULES

New range of Mullard voltage multiplying modules conforming to the 'European Standard' outline. This range is intended for new colour television chassis designs where space saving is an important consideration.

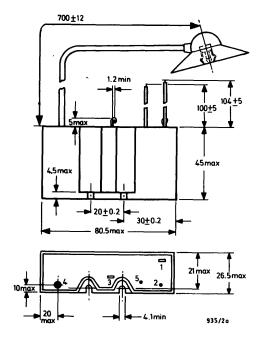
QUICK REFERENC	E DATA	
	LP1196/60	
V <sub>in</sub> (peak to peak)	8. 3	kV
V <sub>out</sub> (e.h.t. supply) (d.c.)	25	kV
Vout (focus supply) (d.c.)	8. 3	kV

## TYPE NUMBERS

LP1196/40 - A five capacitor, six diode module with surge limiting resistor and with the clipping diode across the input.

LP1196/60 - As LP1196/40 with separate lead out for the clipping diode.

### DIMENSIONS (millimetres)



## Terminations

1	=	innıı

<sup>2 =</sup> earth

<sup>3 =</sup> focus

<sup>4 =</sup> output

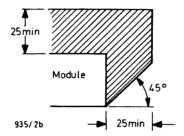
<sup>5 =</sup> clipping diode

### TYPICAL OPERATING CONDITIONS

V <sub>in</sub> (peak-to-peak) (d.c.)	8. 3	kV
Vout (e.h.t. supply) (d.c.)	25	kV
Vout (focus supply) (d.c.)	8. 3	kV
I <sub>out</sub> (e.h.t. supply) (d.c.)	1.0	mA
I <sub>out</sub> (focus supply) (d.c.)	0.25	mA
Internal impedance	<1	$M\Omega$
Surge limiting resistor	60	kΩ
Input capacitance	10	pF

### MOUNTING PRECAUTIONS

A minimum clearance of 25 mm is essential around all surfaces except the base, to prevent breakdown and detuning effects. This is only necessary above an angle of  $45^{\circ}$  from the base of the module.



### HEALTH AND SAFETY

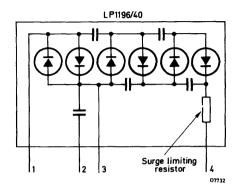
The module is sold in compliance with a valid BS 415 certificate. It is designed to meet the normal conditions of use in a television set with regard to corona, insulation etc. However, the volume of minimum clearance as defined above, under mounting precautions, and within 2" of exposed terminations, must be considered as always being hazardous to personnel unless fully discharged.

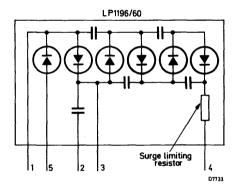
RATINGS (Limiting values of operation according to the Absolute Maximum System).

Input voltage (peak-to-peak)	10.6	kV
•		
Ambient temperature (tripler)	60	°C
Output voltage $(I_{out} = 0)$	30	kV
Output current (e.h.t.)	1.7	mA
Focus current	0.4	mA
Clipping diode current	4.0	mA*
Storage temperature	-25 to +70	°C
Focus to e.h.t. voltage	20	kV

<sup>\*</sup>The clipping diode current is the sum of the clipping diode load current plus the focus and e.h.t. currents.

CIRCUIT DIAGRAMS





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## LOUDSPEAKERS, TELEVISION ASSEMBLIES AND MODULES

Section	Type No.	Section	Type No.	Section	Type No.	Section
Α	AD4890/X	Α	AT1074	С	DL51	D
Α	AD5060/Sq	Α	AT1080	D	DL60	D
A	AD5061/M	Α	AT1081	D	DL700	D
A	AD5061/Sq	Α	AT1083/01	D	ELC1042	В
A	AD5780/M	Α	AT1085	D	ELC1042/05	В
Α [	AD5780/X	Α	AT2048/11	С	ELC1043/05	В
Α	AD5790/M	Α.	AT2076/35	D	ELC1043/06	В -
A	AD5790/X	Α	AT2076/55	D	LP1174*	E
A	AD7060/W	Α	AT2080/10	D	LP1194	E
Α	AD7062/M	Α	AT2080/15	D	LP1196	E
A	AD7063/M	Α	AT2095	D	U321	В
Α	AD7066/W	Α	AT2102/01	С	U321LO	В
Α	AD7080/M	Α	AT2140/10	С	U322	В
A	AD7080/X	Α	AT4034/01	С	U322LO	В
Α	AD7091/M	Α	AT4036	С	V311	В
A	AD7091/X	Α	AT4042/02	С	V314	В
Α	AD8000	Α	AT4042/08	С	V315	В
Α	AD8061/W	Α	AT4042/14	С	V315LO	В
Α	AD8066/W	Α	AT4042/38	D	VS340/01	D
Α	AD8067/W	Α	AT4043/29	D	VS400/01	D
Α	AD10100/W	Ą	AT4043/34	D <sup>-</sup>	VA470/01	D
Α	AD12100/HP	A	AT4043/38	D	VS550/01	D
Α	AD12100/M	Α	AT4043/59	С	VS600/01	D
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Α	ADF700/2600	Α	AT4044/27	D	4322 153 01100	D
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Α	AT1071/01	С	DL50	D		
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<sup>\*</sup>Available for current production; not intended for new designs.

## The following data sheets have been withdrawn:

AT1025 Series	AT2055	AT4041/40	LP1184/2
AT1027	AT2055/02	AT4042/17	LP1185
AT1029	AT2063/00	AT4043/86	LP1186
AT1062/01	AT2063/03	AT4046 Series	LP1400
AT1063/01	AT4040 Series	LP1173	LP1402
AT1068/03	AT4041/08	LP1181	
AT1068/04	AT4041/37	LP1183/2	

# LOUDSPEAKERS, TELEVISION ASSEMBLIES **AND MODULES**



**LOUDSPEAKERS** 

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**ASSEMBLIES** 

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**TELEVISION TUNERS** 

**MONOCHROME TELEVISION** 

**COLOUR TELEVISION ASSEMBLIES** 

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**Mullard Limited** 

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